

GEORGIA INSTITUTE OF TECHNOLOGY  
School of Electrical and Computer Engineering

EE3230  
Quiz No. 3  
March 4, 1998

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Name: \_\_\_\_\_

1. The exam is closed book. You may use one 8.5" by 11" sheet of notes (both sides). You are permitted to use a calculator. **I have given you a sheet of Fourier transform formulas as the last page. Tear it off and use it!**
2. Do all work in the space provided. If you need more room, use the *back* of the *previous* page and indicate that you have done so.
3. Indicate your answer clearly by circling it or drawing a box around it.
4. Think carefully about the the problem before you begin to write. Move on to the next problem if you cannot come up with a plan for the solution.
5. If you want to receive partial credit, you should clearly indicate your reasoning and method of attack on the problem.

Problem	Points	Score
1	40	
2	30	
3	30	
TOTAL	100	

**Problem W98.Q2.1 (40 %)**

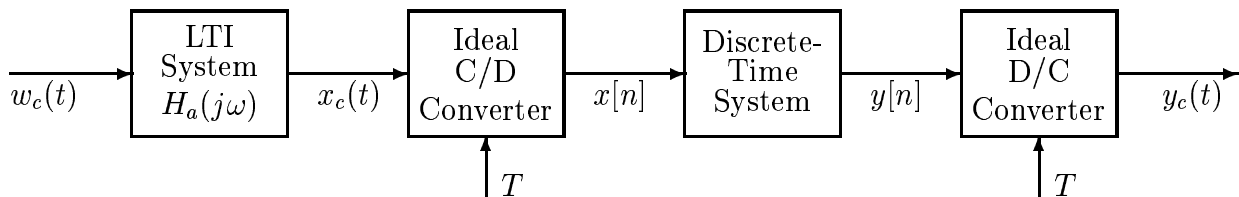
A linear time-invariant system with input  $x(t)$  and output  $y(t)$  has system function

$$H(s) = \frac{s - 2}{s + 2}$$

- (a) If it is known that the system is causal, determine the impulse response  $h(t)$ .
- (b) Is the system stable? *yes* *no*  
How do you know?
- (c) Suppose that the input to the system is  $x(t) = u(-t)$ . Determine the corresponding output  $y(t)$ .
- (d) Determine the Laplace transform  $X(s)$  of the input  $x(t)$  so that  $y(t) = \delta(t - 5)$ .  
If it is known that  $\int_{-\infty}^{\infty} |x(t)| dt < \infty$ , what is the region of convergence of  $X(s)$ ?

**Problem W98.Q2.2: (30 %)**

All parts of this problem are concerned with the following system.



$$x[n] = x_c(nT)$$

$$y_c(t) = \sum_{n=-\infty}^{\infty} y[n] \frac{\sin \frac{\pi}{T}(t - nT)}{\frac{\pi}{T}(t - nT)}$$

- (a) If  $T = 1/1000$ , what condition on  $H_a(j\omega)$  will guarantee that  $y_c(t) = x_c(t)$  if the discrete-time system is defined by the difference equation  $y[n] = x[n]$ ?

- (b) If  $T = 1/1000$  and the system  $H_a(j\omega)$  is as chosen in part (a), determine  $y_c(t)$  in terms of  $x_c(t)$  if the discrete-time system is defined by the difference equation  $y[n] = x[n - 100]$ .

- (c) If the filter  $H_a(j\omega)$  satisfies the condition determined in part (a) and the discrete-time system now is defined by  $y[n] = x[n] + x[n - 1]$ , what is the frequency response  $H_c(j\omega)$  of the overall system from  $w_c(t)$  to  $y_c(t)$ ; i.e., what is  $H_c(j\omega)$  in the equation

$$Y_c(j\omega) = H_c(j\omega)W_c(j\omega)$$

Be as specific as you can.

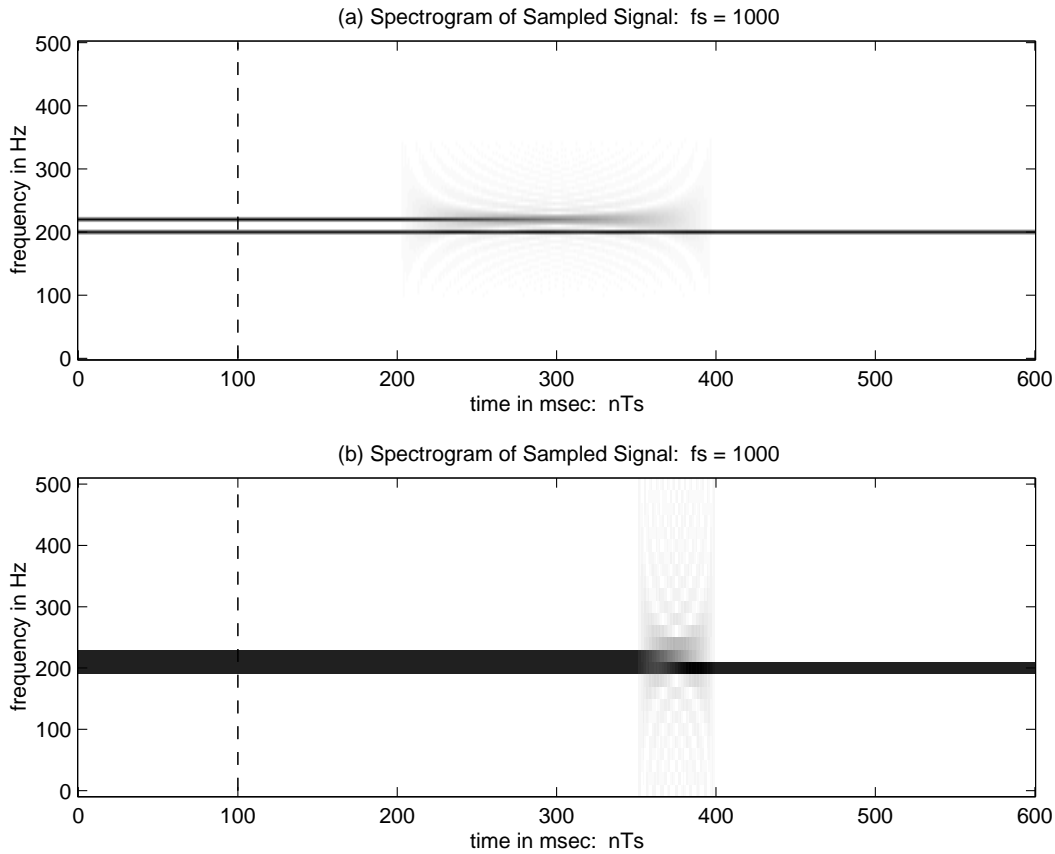
**Problem W98.Q2.3: (30 %)**Figure 1: Spectrum slices for two different values of  $N$ .

Figure 1 shows the spectrogram defined as

$$X[k, n] = \sum_{m=n}^{n+N-1} x[m] e^{-j(2\pi k/N)m} \quad k = 0, 1, \dots, N/2 \text{ and } n = 0, 1, \dots, L \quad (1)$$

In the figure, different values of  $N$  are used in parts (a) and (b). Also, the  $k$  axis is labelled in terms of continuous-time frequency and the  $n$  axis is labelled in terms of continuous-time. The sampling rate of the signal is  $f_s = 1000$  Hz.

- (i) From the figure and the given sampling rate, estimate the value of  $N$  for both spectrograms.

**Problem W98.Q2.3 (continued): (30 %)**

The dotted vertical lines in Figure 1 correspond to the particular value of  $n = 100$  samples in Equation (1). The plots in Figure 2 show  $|X[k, 100]|$  as a function of  $k$  for  $0 \leq k \leq N/2$  for both case (a) and case (b).

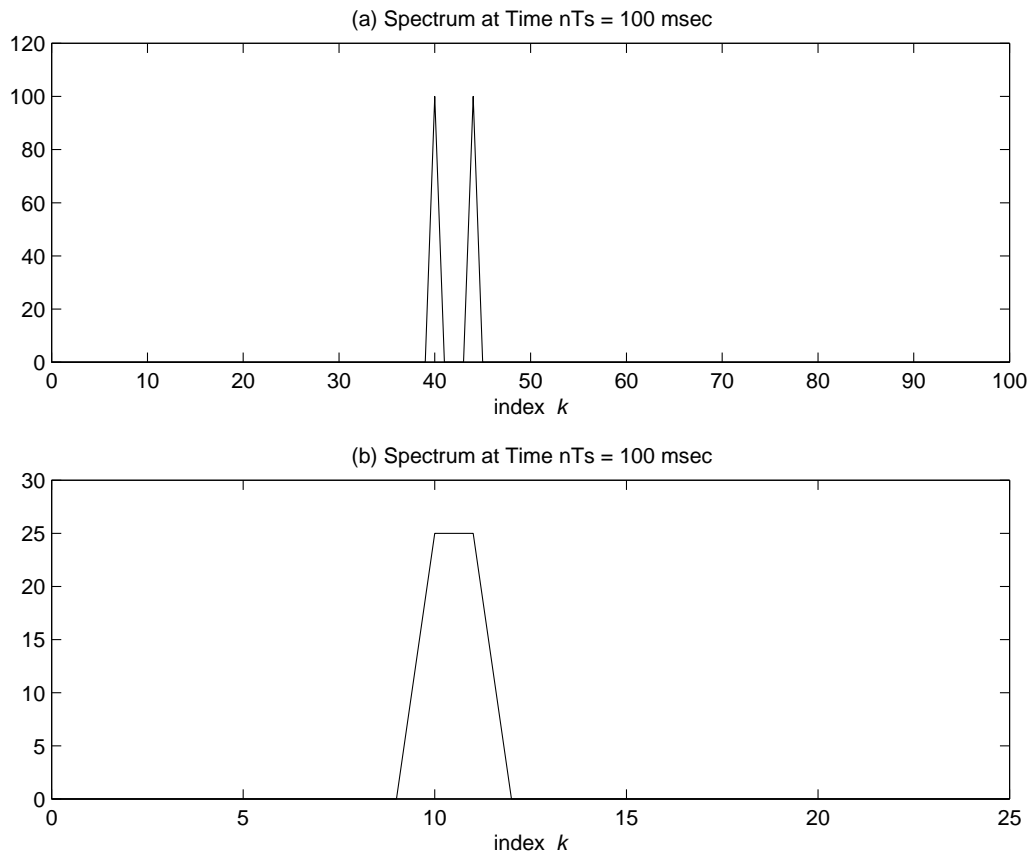


Figure 2: Spectrum slices for two different values of  $N$ .

- (ii) Label the  $k$  axis in both of the above plots. In particular, label the far right point of the plot and label the exact index of the locations of the peaks.
- (iii) At what time does the signal  $x[n]$  change from having 2 frequencies to having only one frequency? *Answer:*  $n = \underline{\hspace{2cm}}$   
 Give an equation that describes  $x[n]$  as completely as you can.

## FORMULAS FOR SECOND EXAM

### Fourier Transform Pairs

$$\begin{aligned}
 x(t) = e^{-at}u(t) &\xleftrightarrow{\mathcal{F}} X(j\omega) = \frac{1}{a + j\omega} \\
 x(t) = \begin{cases} 1 & |t| < T_p/2 \\ 0 & |t| > T_p/2 \end{cases} &\xleftrightarrow{\mathcal{F}} X(j\omega) = \frac{2 \sin(\omega T_p/2)}{\omega} \\
 x(t) = \frac{\sin \omega_b t}{\pi t} &\xleftrightarrow{\mathcal{F}} X(j\omega) = \begin{cases} 1 & |\omega| < \omega_b \\ 0 & |\omega| > \omega_b \end{cases} \\
 x(t) = \delta(t) &\xleftrightarrow{\mathcal{F}} X(j\omega) = 1 \\
 x(t) = u(t) &\xleftrightarrow{\mathcal{F}} X(j\omega) = \pi\delta(\omega) + \frac{1}{j\omega} \\
 x(t) = u_1(t) &\xleftrightarrow{\mathcal{F}} X(j\omega) = j\omega \\
 x(t) = 1 &\xleftrightarrow{\mathcal{F}} X(j\omega) = 2\pi\delta(\omega) \\
 x(t) = e^{j\omega_0 t} &\xleftrightarrow{\mathcal{F}} X(j\omega) = 2\pi\delta(\omega - \omega_0) \\
 x(t) = \cos \omega_0 t &\xleftrightarrow{\mathcal{F}} X(j\omega) = \pi\delta(\omega - \omega_0) + \pi\delta(\omega + \omega_0) \\
 x(t) = \sin \omega_0 t &\xleftrightarrow{\mathcal{F}} X(j\omega) = -j\pi\delta(\omega - \omega_0) + j\pi\delta(\omega + \omega_0) \\
 \left. \begin{aligned} x(t) &= \sum_{k=-\infty}^{\infty} a_k e^{j\omega_0 k t} \\ a_k &= \frac{1}{T} \int_0^T x(t) e^{-jk\omega_0 t} dt \end{aligned} \right\} &\xleftrightarrow{\mathcal{F}} X(j\omega) = \sum_{k=-\infty}^{\infty} 2\pi a_k \delta(\omega - k\omega_0)
 \end{aligned}$$

### Fourier Transform Theorems

$$\begin{aligned}
 ax_1(t) + bx_2(t) &\xleftrightarrow{\mathcal{F}} aX_1(j\omega) + bX_2(j\omega) \\
 x(t - t_0) &\xleftrightarrow{\mathcal{F}} e^{-j\omega t_0} X(j\omega) \\
 e^{j\omega_0 t} x(t) &\xleftrightarrow{\mathcal{F}} X(j(\omega - \omega_0)) \\
 \frac{dx(t)}{dt} &\xleftrightarrow{\mathcal{F}} j\omega X(j\omega) \\
 -jtx(t) &\xleftrightarrow{\mathcal{F}} \frac{dX(j\omega)}{d\omega} \\
 y(t) = x(t) * h(t) &\xleftrightarrow{\mathcal{F}} Y(j\omega) = H(j\omega)X(j\omega) \\
 x(t)p(t) &\xleftrightarrow{\mathcal{F}} \frac{1}{2\pi} X(j\omega) * P(j\omega) \\
 x(at) &\xleftrightarrow{\mathcal{F}} \frac{1}{|a|} X\left(\frac{j\omega}{a}\right) \\
 X(jt) &\xleftrightarrow{\mathcal{F}} 2\pi x(-\omega)
 \end{aligned}$$