

EE-2025

Spring-2000

Lecture 8

D-to-A Conversion

11-Feb-99

Information

- Check the Bulletin Board for msgs
- Lab #4 is posted
 - Notes file: **cgnotes.mat** (cgshort.mat)
 - Spectrogram image display info
 - New M-file: **plotspec.m** & spectgr.m
 - FORMAL Lab Report (150 points)
- Problem Set #5
 - **Topic = Sampling & Aliasing**

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READING ASSIGNMENTS

- This Lecture:
 - Chapter 4, pp. 100–111
- Other Reading:
 - Recitation: Chapter 4, pp. 90–100
 - Strobe Demo
 - Next Lecture: Chapter 5 (beginning)

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LECTURE OBJECTIVES

- **FOLDING**: a type of **ALIASING**
- **DIGITAL-to-ANALOG CONVERSION** is
 - Reconstruction from samples
 - **SAMPLING THEOREM** applies
 - Smooth **Interpolation**
- **Mathematical Model of D-to-A**
 - **SUM of SHIFTED PULSES**
 - Linear Interpolation example

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SIGNAL TYPES



■ A-to-D

| Convert $x(t)$ to **numbers** stored in memory

■ D-to-A

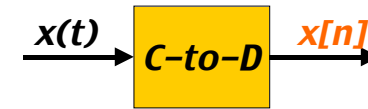
| Convert $y[n]$ back to a “continuous-time” signal, $x(t)$

| $y[n]$ is called a “**discrete-time**” signal

SAMPLING $x(t)$

■ UNIFORM SAMPLING at $t = nT_s$

| IDEAL: $x[n] = x(nT_s)$



Shannon Sampling Theorem

A continuous-time signal $x(t)$ with frequencies no higher than f_{\max} can be reconstructed exactly from its samples $x[n] = x(nT_s)$, if the samples are taken at a rate $f_s = 1/T_s$ that is greater than $2f_{\max}$.

NYQUIST RATE

■ “Nyquist Rate” Sampling

| $f_s =$ TWICE THE HIGHEST FREQUENCY in $x(t)$

| “Sampling above the Nyquist rate”

■ BANDLIMITED SIGNALS

| DEF: $x(t)$ has a HIGHEST FREQUENCY COMPONENT in its SPECTRUM

| NON-BANDLIMITED EXAMPLE

| TRIANGLE WAVE is **NOT** BANDLIMITED

DEMOS from CHAPTER 4

■ CD-ROM DEMOS

■ SAMPLING DEMO

| Different Sampling Rates

| Aliasing of a Sinusoid

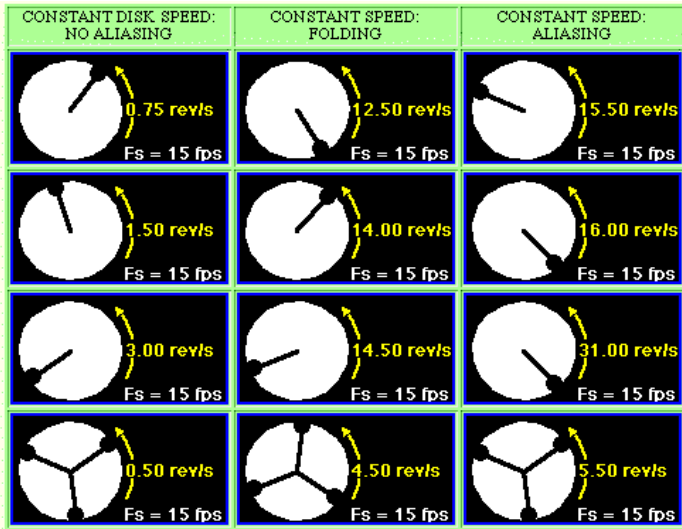
■ STROBE DEMO

| Synthetic vs. Real

| Television **SAMPLES** at 30 fps

■ Sampling & Reconstruction

STROBE DEMO (Synthetic)

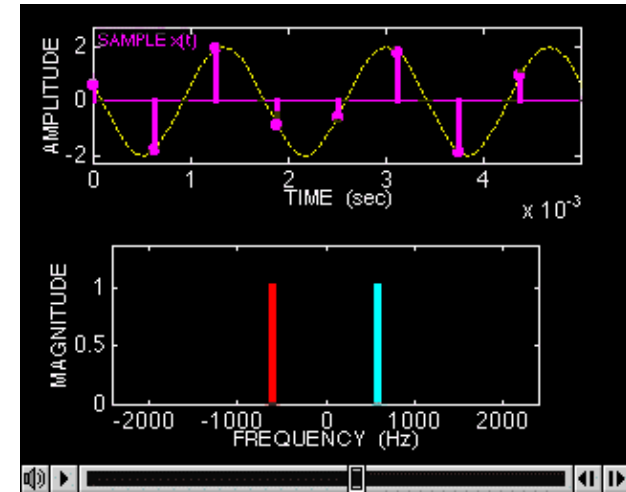


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SAMPLING DEMO (Ch. 4)



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SPECTRUM for $x[n]$

- INCLUDE **ALL** SPECTRUM LINES
 - ALIASES
 - ADD INTEGER MULTIPLES of 2π and -2π
 - FOLDED ALIASES
 - ALIASES of NEGATIVE FREQS
- PLOT versus NORMALIZED FREQUENCY
 - i.e., DIVIDE f_0 by f_s

$$\hat{\omega} = 2\pi \frac{f}{f_s} + 2\pi\ell$$

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EXAMPLE: SPECTRUM

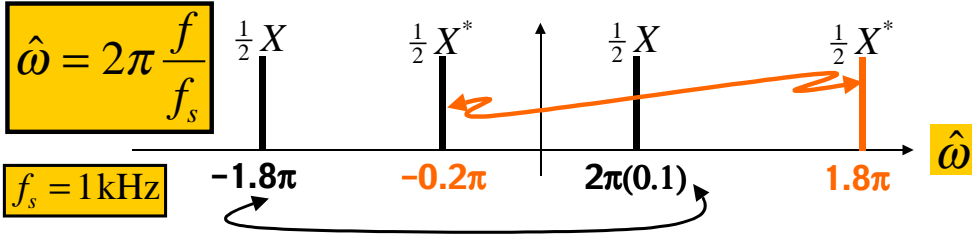
- $x[n] = A\cos(0.2\pi n + \phi)$
- FREQS @ 0.2π and -0.2π
- ALIASES:
 - $\{2.2\pi, 4.2\pi, 6.2\pi, \dots\}$ & $\{-1.8\pi, -3.8\pi, \dots\}$
 - EX: $x[n] = A\cos(4.2\pi n + \phi)$
- ALIASES of **NEGATIVE** FREQ:
 - $\{1.8\pi, 3.8\pi, 5.8\pi, \dots\}$ & $\{-2.2\pi, -4.2\pi, \dots\}$

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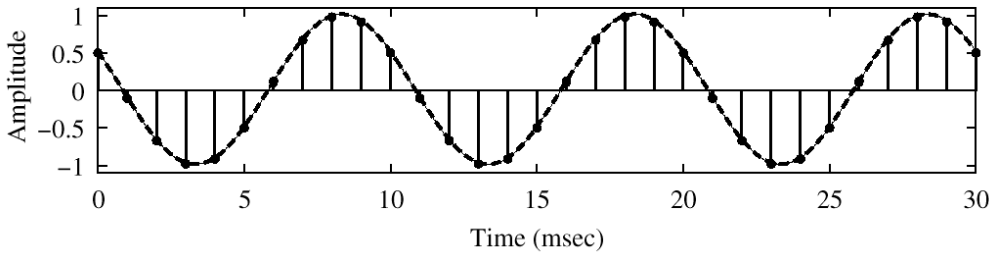
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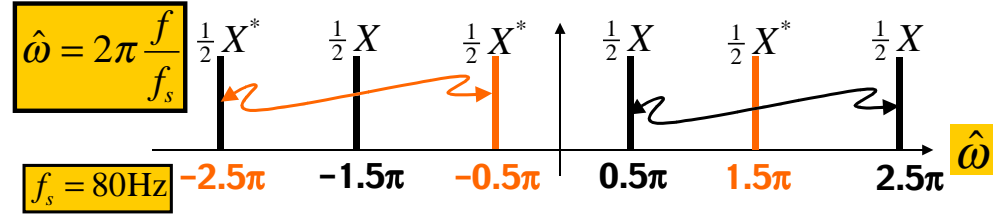
SPECTRUM (MORE LINES)



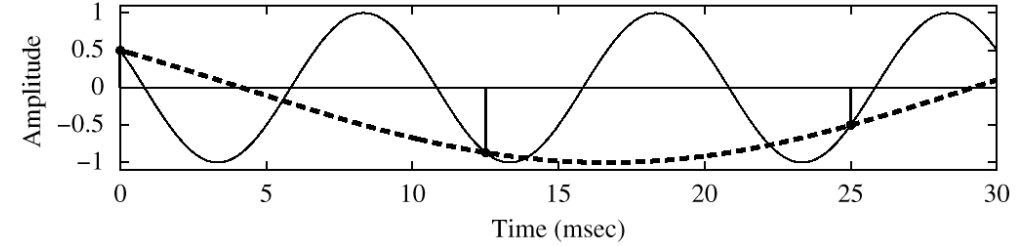
100-Hz Cosine Wave: Sampled with $T_s = 1$ msec (1000 Hz)



SPECTRUM (ALIASING CASE)



100-Hz Cosine Wave: Sampled with $T_s = 12.5$ msec (80 Hz)



FOLDING (a type of ALIASING)

- EXAMPLE: 3 different $x(t)$; same $x[n]$
- 900 Hz “folds” to 100 Hz when $f_s = 1 \text{ kHz}$

$f_s = 1000 \text{ Hz}$

$\cos(2\pi(100)t) \rightarrow \cos[2\pi(0.1)n]$

$\cos(2\pi(1100)t) \rightarrow \cos[2\pi(1.1)n] = \cos[2\pi(0.1)n]$

$\cos(2\pi(900)t) \rightarrow \cos[2\pi(0.9)n]$

$= \cos[2\pi(0.9)n - 2\pi n] = \cos[2\pi(-0.1)n] = \cos[2\pi(0.1)n]$

$\hat{\omega} = 2\pi \frac{100}{1000} = 2(0.1)$

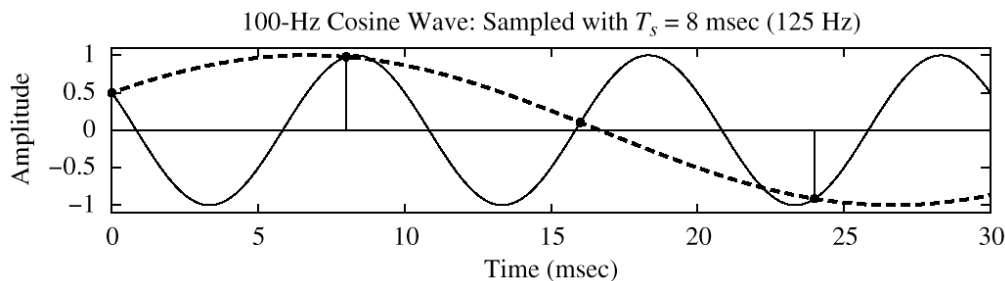
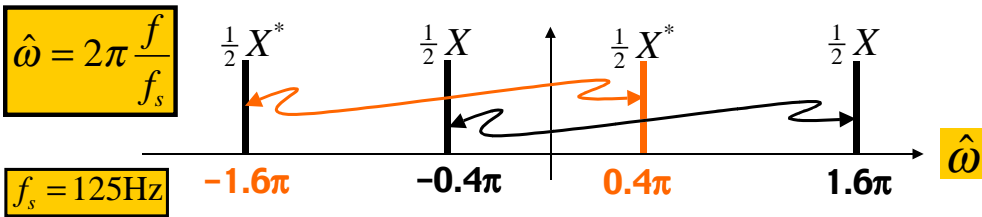
DIGITAL FREQ $\hat{\omega}$ AGAIN

Normalized Radian Frequency

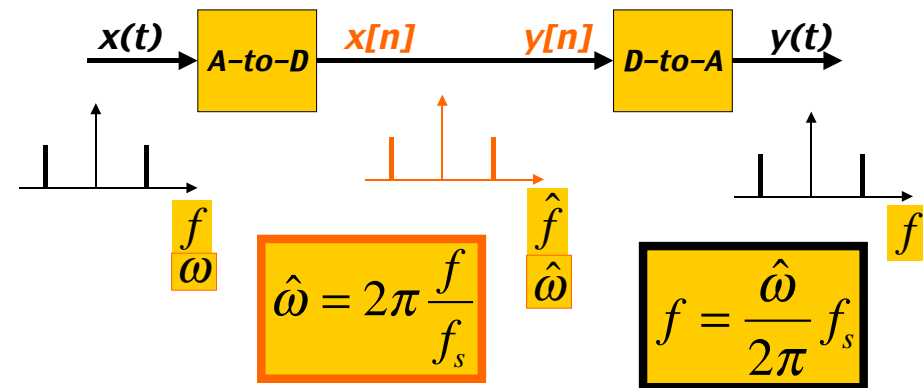
$\hat{\omega} = \omega T_s = \frac{2\pi f}{f_s} + 2\pi\ell$ **ALIASING**

$\hat{\omega} = \omega T_s = -\frac{2\pi f}{f_s} + 2\pi\ell$ **FOLDED ALIAS**

SPECTRUM (FOLDING CASE)



FREQUENCY DOMAINS



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D-to-A Reconstruction



■ Create continuous $y(t)$ from $y[n]$

■ IDEAL

■ If you have formula for $y[n]$

■ Replace n in $y[n]$ with $f_s t$

■ $y[n] = A \cos(0.2\pi n + \phi)$ with $f_s = 8000 \text{ Hz}$

■ $y(t) = A \cos(2\pi(800)t + \phi)$

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D-to-A is AMBIGUOUS !

■ ALIASING

■ Given $y[n]$, which $y(t)$ do we pick ???

■ INFINITE NUMBER of $y(t)$

■ PASSING THRU THE SAMPLES, $y[n]$

■ D-to-A RECONSTRUCTION MUST CHOOSE ONE OUTPUT

■ RECONSTRUCT THE SMOOTHEST ONE

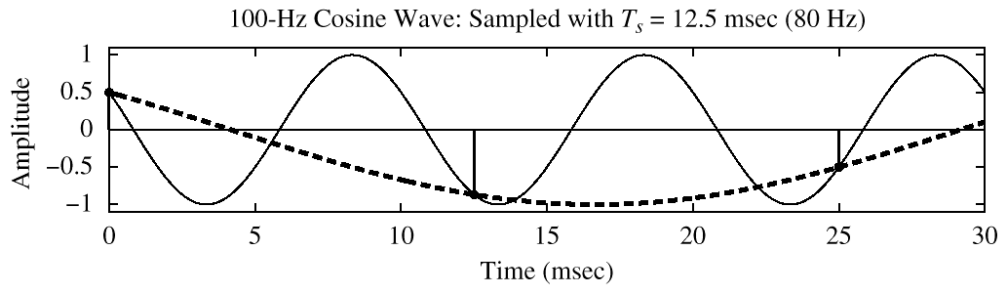
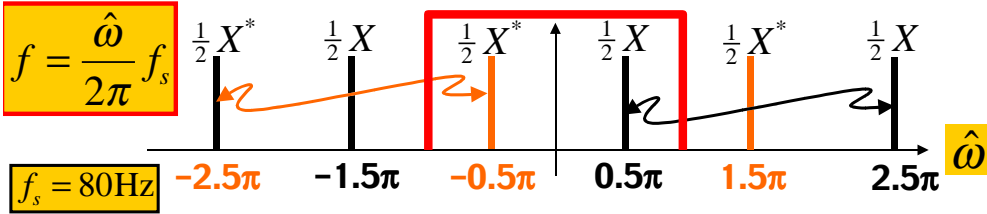
■ THE **LOWEST** FREQ, if $y[n] = \text{sinusoid}$

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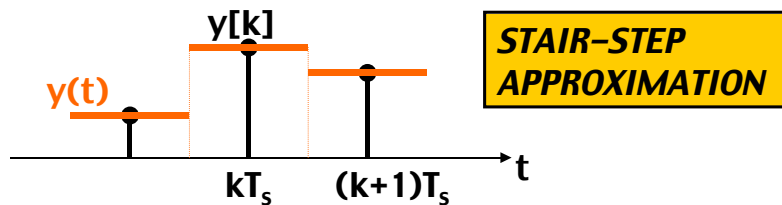
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SPECTRUM (ALIASING CASE)



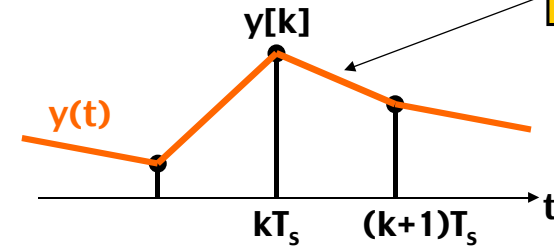
SAMPLE & HOLD DEVICE

- CONVERT $y[n]$ to $y(t)$
 - $y[k]$ should be the value of $y(t)$ at $t = kT_s$
 - Make $y(t)$ equal to $y[k]$ for $kT_s - 0.5T_s < t < kT_s + 0.5T_s$

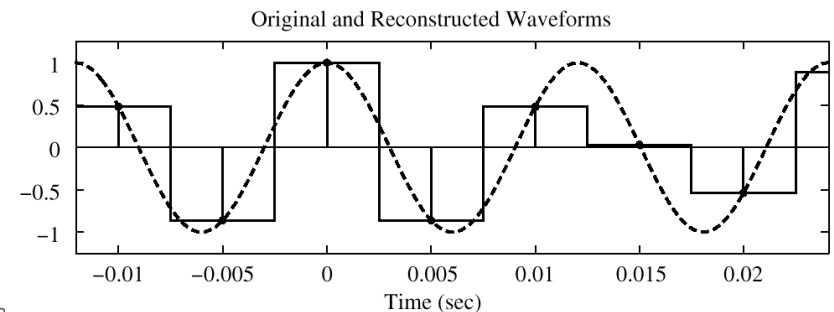
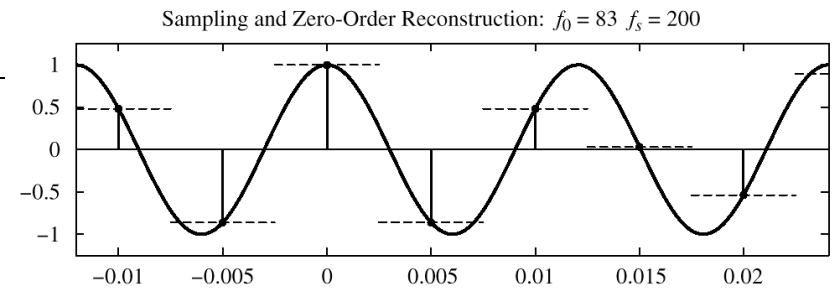


Reconstruction (D-to-A)

- CONVERT STREAM of NUMBERS to $x(t)$
- “CONNECT THE DOTS”
- INTERPOLATION



SQUARE PULSE CASE



MATH MODEL for D-to-A

$$y(t) = \sum_{n=-\infty}^{\infty} y[n]p(t - nT_s)$$

SQUARE PULSE:

$$p(t) = \begin{cases} 1 & -\frac{1}{2}T_s < t \leq \frac{1}{2}T_s \\ 0 & \text{otherwise} \end{cases}$$

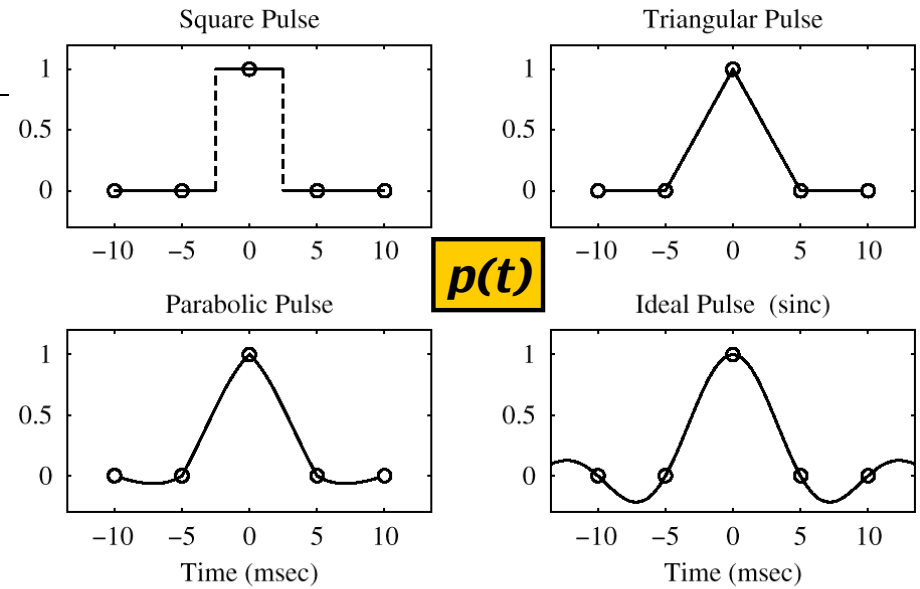


Figure 4.17 Four different pulses for D-to-C conversion. The sampling period is $T_s = 0.005$, i.e., $f_s = 200$ Hz. Note that the duration of each pulse is approximately one or two times T_s .

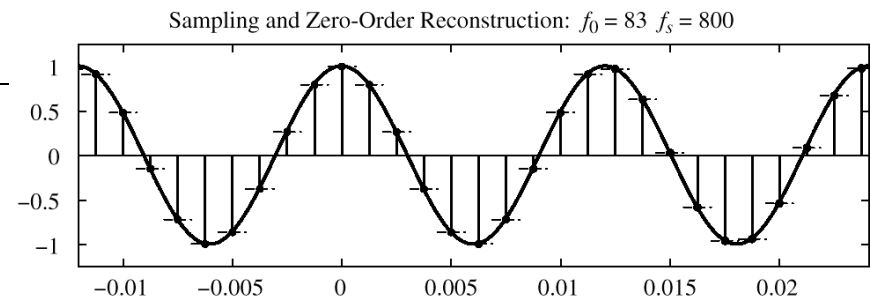
EXPAND the SUMMATION

$$\sum_{n=-\infty}^{\infty} y[n]p(t - nT_s) = \dots + y[0]p(t) + y[1]p(t - T_s) + y[2]p(t - 2T_s) + \dots$$

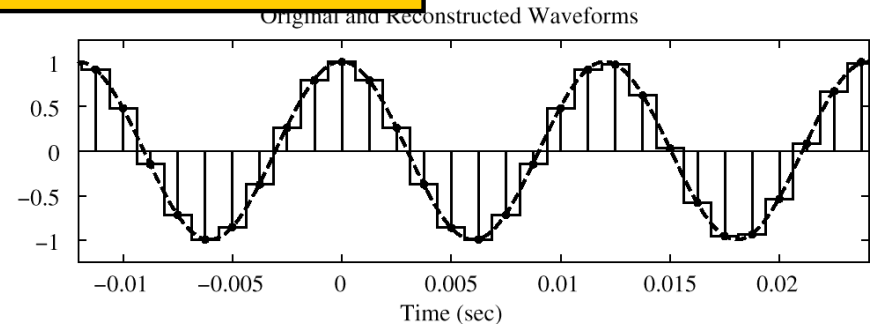
SUM of SHIFTED PULSES $p(t - nT_s)$

- | "WEIGHTED" by $y[n]$
- | CENTERED at $t = nT_s$
- | SPACED by T_s
- | RESTORES "REAL TIME"

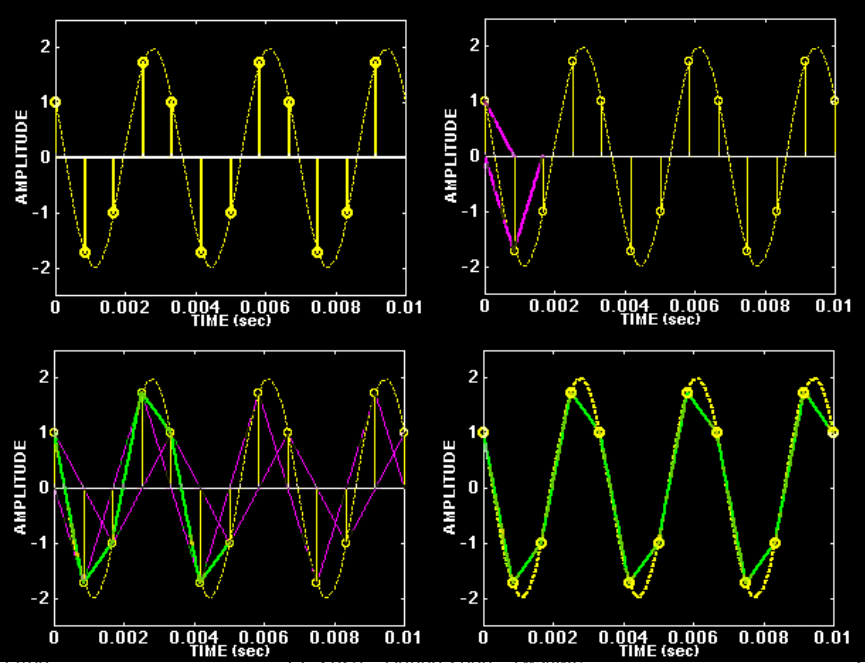
OVER-SAMPLING CASE



EASIER TO RECONSTRUCT

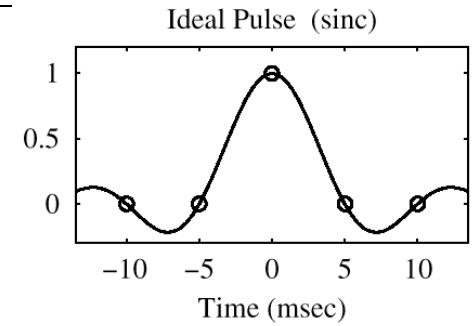


TRIANGULAR PULSE (2X)



OPTIMAL PULSE ?

**CALLLED
"BANDLIMITED
INTERPOLATION"**



$$p(t) = \frac{\sin \frac{\pi}{T_s} t}{\frac{\pi}{T_s} t} \quad \text{for } -\infty < t < \infty$$

$$p(t) = 0 \quad \text{for } t = 0, \pm T_s, \pm 2T_s$$