EE-2025 Spring 2001

Lecture 17

Frequency Response of Continuous-Time Systems

23-March-2001

Info: Web-CT, Lab, HW

- Calendar:
  - Quiz #3 is 6-April
- CHECK YOUR GRADES !!!
  - Web-CT is the OFFICIAL gradebook
- Prob Set #9 is due next week
- Prob Set #10 available this weekend
- Lab Quiz next week (27-Mar thru 29-Mar)
  - Lab #8 is due next week
  - Lab #10 to be done next week

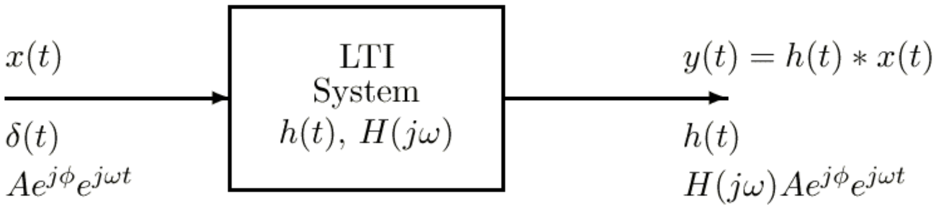
READING ASSIGNMENTS

- This Lecture:
  - Chapter 11, pp. 1100-1118 (all)
  - In NEW Chapters
- Other Reading:
  - Next Lecture & Recitation:
  - Chapter 12, 1200-1230

LECTURE OBJECTIVES

- Review of convolution
  - **THE** operation for **LTI** Systems
- Complex exponential input signals
  - Frequency Response
  - Cosine signals
    - Real part of complex exponential
- Fourier Series thru $H(j\omega)$
  - These are Analog Filters

LTI Systems



- Convolution defines LTI system

$$y(t) = h(t) * x(t) = \int_{-\infty}^{\infty} h(\tau)x(t-\tau)d\tau$$

- Response to a complex exponential gives frequency response *H(jω)*

Complex Exponential Input

$$x(t) = Ae^{j\phi}e^{j\omega t} \mapsto y(t) = H(j\omega)Ae^{j\phi}e^{j\omega t}$$

$$y(t) = \int_{-\infty}^{\infty} h(\tau)Ae^{j\phi}e^{j\omega(t-\tau)}d\tau$$

$$y(t) = \left(\int_{-\infty}^{\infty} h(\tau)e^{-j\omega\tau}d\tau\right)Ae^{j\phi}e^{j\omega t}$$

$$H(j\omega) = \int_{-\infty}^{\infty} h(\tau)e^{-j\omega\tau}d\tau$$

Frequency Response

When does *H(jω)* Exist?

- When is $|H(j\omega)| < \infty$?

$$|H(j\omega)| = \left|\int_{-\infty}^{\infty} h(\tau)e^{-j\omega\tau}d\tau\right| \le \int_{-\infty}^{\infty} |h(\tau)||e^{-j\omega\tau}|d\tau$$

$$|H(j\omega)| \le \int_{-\infty}^{\infty} |h(\tau)|d\tau < \infty$$

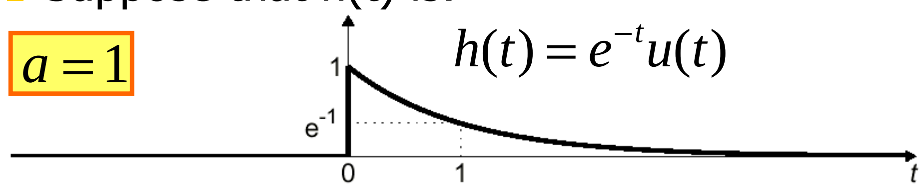
- Thus the frequency sytem is a stable system.

$$h(t) = e^{-at}u(t) \Leftrightarrow H(j\omega) = \frac{1}{a + j\omega}$$

- Suppose that h(t) is:

$a = 1$

$$h(t) = e^{-t}u(t)$$



$$H(j\omega) = \int_{-\infty}^{\infty} e^{-a\tau}u(\tau)e^{-j\omega\tau}d\tau = \int_{0}^{\infty} e^{-(a+j\omega)\tau}d\tau$$

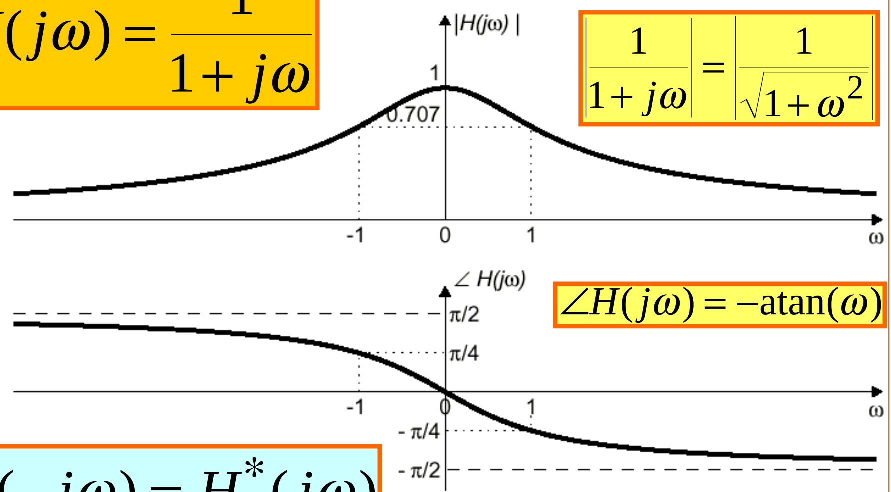
$a > 0$

$$H(j\omega) = \left.\frac{e^{-(a+j\omega)\tau}}{-(a+j\omega)}\right|_0^{\infty} = \left.\frac{e^{-a\tau}e^{-j\omega\tau}}{-(a+j\omega)}\right|_0^{\infty} = \frac{1}{a+j\omega}$$

Magnitude and Phase Plots

$$H(j\omega) = \frac{1}{1+j\omega}$$

$$\left|\frac{1}{1+j\omega}\right| = \frac{1}{\sqrt{1+\omega^2}}$$



$$\angle H(j\omega) = -\text{atan}(\omega)$$

$$H(-j\omega) = H^*(j\omega)$$

Freq Response of Integrator?

- Impulse Response
  - h(t) = u(t)
- NOT a Stable System
  - Frequency response *H(jω)* does NOT exist
- Leaky Integrator (a is small)
  - Cannot build a perfect Integral

$$a \to 0$$

$$h(t) = e^{-at}u(t) \Leftrightarrow H(j\omega) = \frac{1}{a+j\omega} \to \frac{1}{j\omega}?$$

Ideal Delay: $y(t) = x(t-t_d)$

$$H(j\omega) = \int_{-\infty}^{\infty} \delta(\tau - t_d) e^{-j\omega\tau} d\tau = e^{-j\omega t_d}$$

$$H(j\omega) = e^{-j\omega t_d}$$

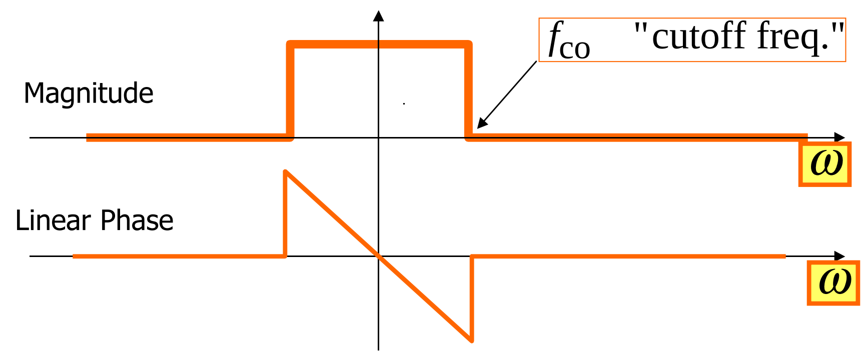
$$x(t) = e^{j\omega t} \mapsto$$

$$H(j\omega)$$

$$y(t) = e^{j\omega(t-t_d)} = e^{-j\omega t_d} e^{j\omega t}$$

Ideal Lowpass Filter

$$H_{LP}(j\omega) = \begin{cases} e^{-j\omega t_d} & |\omega| < \omega_{co} \\ 0 & |\omega| > \omega_{co} \end{cases}$$



Example: Ideal Low Pass

$$H_{LP}(j\omega) = \begin{cases} e^{-j3\omega} & |\omega| < 2 \\ 0 & |\omega| > 2 \end{cases}$$

$$x(t) = 10e^{j\pi/3}e^{j5t} \mapsto \quad y(t) = H(j5)10e^{j\pi/3}e^{j5t}$$

$$y(t) = \left(e^{-j15}\right)10e^{j\pi/3}e^{j5t} = 10e^{j\pi/3}e^{j5(t-3)}$$

Cosine Input

$$x(t) = A\cos(\omega_0 t + \phi) = \frac{A}{2}e^{j\phi}e^{j\omega_0 t} + \frac{A}{2}e^{-j\phi}e^{-j\omega_0 t}$$

$$y(t) = H(j\omega_0)\frac{A}{2}e^{j\phi}e^{j\omega_0 t} + H(-j\omega_0)\frac{A}{2}e^{-j\phi}e^{-j\omega_0 t}$$

Since $H(-j\omega_0) = H^*(j\omega_0)$

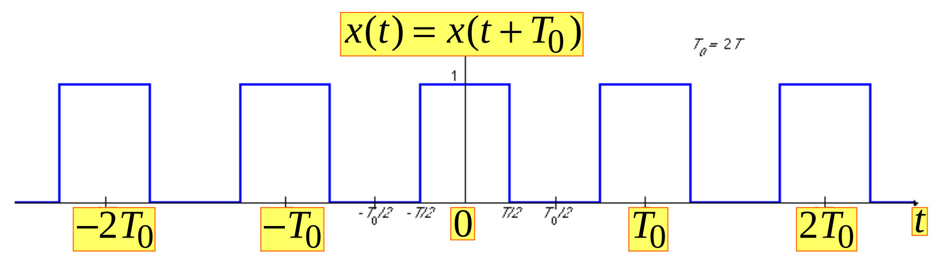
$$y(t) = A|H(j\omega_0)|\cos(\omega_0 t + \phi + \angle H(j\omega_0))$$

Review Fourier Series

- ANALYSIS
  - Get representation from the signal
  - Works for PERIODIC Signals
- Fourier Series
  - INTEGRAL over one period

$$a_k = \frac{1}{T_0}\int_0^{T_0} x(t)e^{-j\omega_0 kt}\,dt$$

General Periodic Signals



$$x(t) = \sum_{k=-\infty}^{\infty} a_k e^{j\omega_0 kt}$$

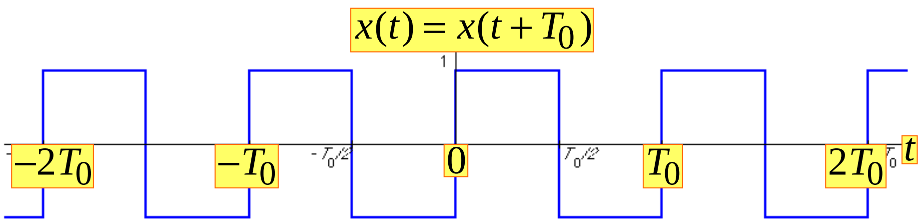
Fourier Synthesis

Fundamental Freq. $\omega_0 = 2\pi / T_0 = 2\pi f_0$

$$a_k = \frac{1}{T_0}\int_0^{T_0} x(t)e^{-j\omega_0 kt}\,dt$$

Fourier Analysis

Square Wave Signal

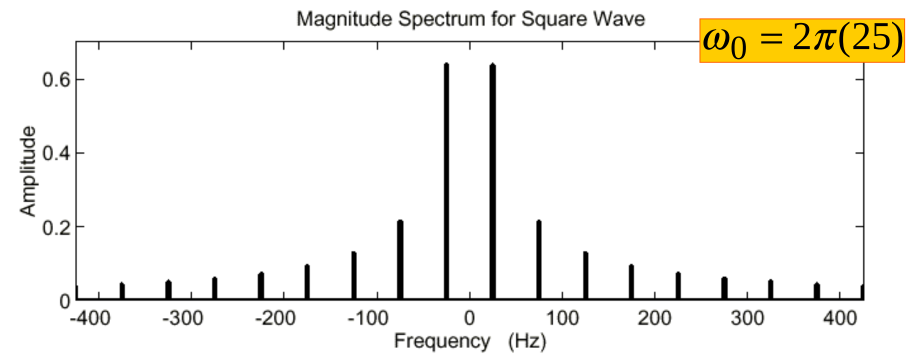


$$a_k = \frac{1}{T_0}\int_0^{T_0/2}(1)e^{-j\omega_0 kt}dt + \frac{1}{T_0}\int_{T_0/2}^{T_0}(-1)e^{-j\omega_0 kt}dt$$

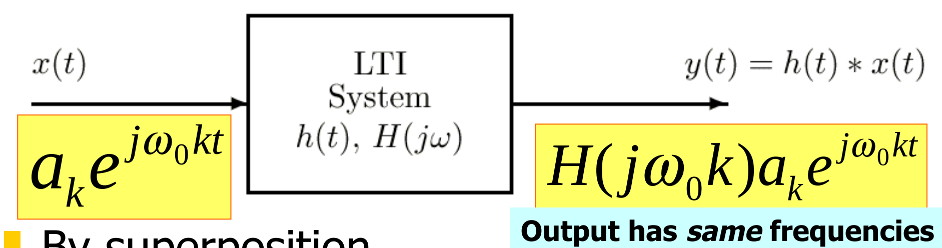
$$a_k = \left.\frac{e^{-j\omega_0 kt}}{-j\omega_0 kT_0}\right|_0^{T_0/2} - \left.\frac{e^{-j\omega_0 kt}}{-j\omega_0 kT_0}\right|_{T_0/2}^{T_0} = \frac{1-e^{-j\pi k}}{j\pi k}$$

Spectrum from Fourier Series

$$a_k = \frac{1-e^{-j\pi k}}{j\pi k} = \begin{cases}\dfrac{2}{j\pi k} & k=\pm1,\pm3,\ldots \\ 0 & k=0,\pm2,\pm4,\ldots\end{cases}$$



LTI Systems with Periodic Inputs

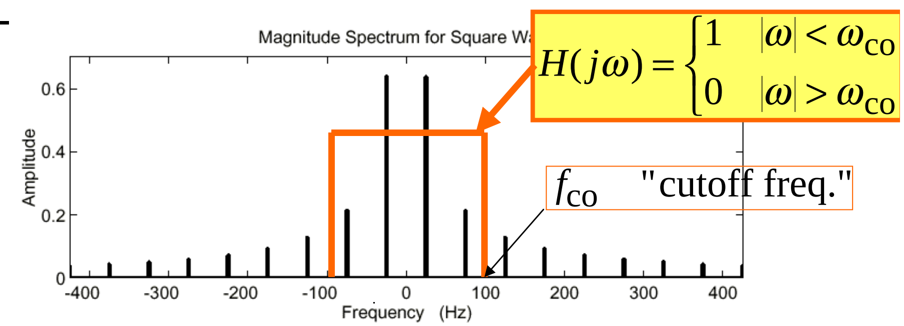


- By superposition,

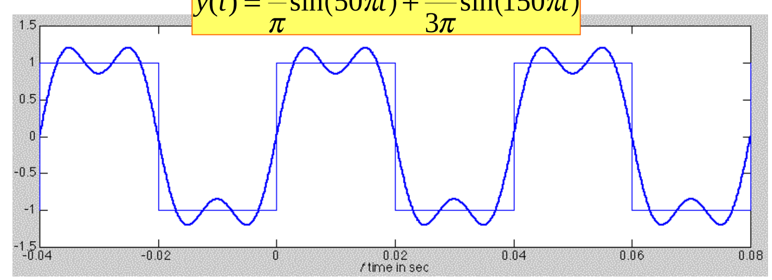
$$y(t) = \sum_{k=-\infty}^{\infty} a_k H(j\omega_0 k)e^{j\omega_0 kt} = \sum_{k=-\infty}^{\infty} b_k e^{j\omega_0 kt}$$

$$b_k = a_k H(j\omega_0 k)$$

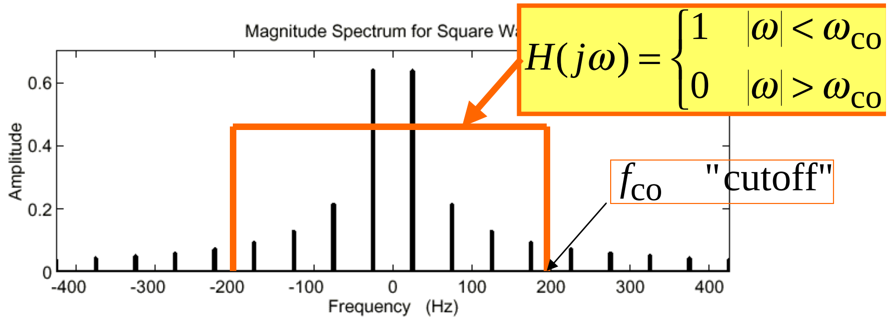
Ideal Lowpass Filter



$$y(t) = \frac{4}{\pi}\sin(50\pi t) + \frac{4}{3\pi}\sin(150\pi t)$$

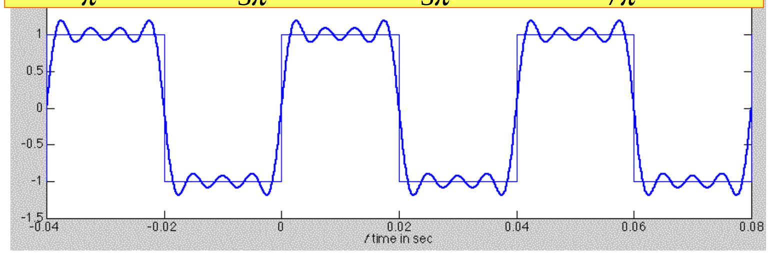


Ideal Lowpass Filter

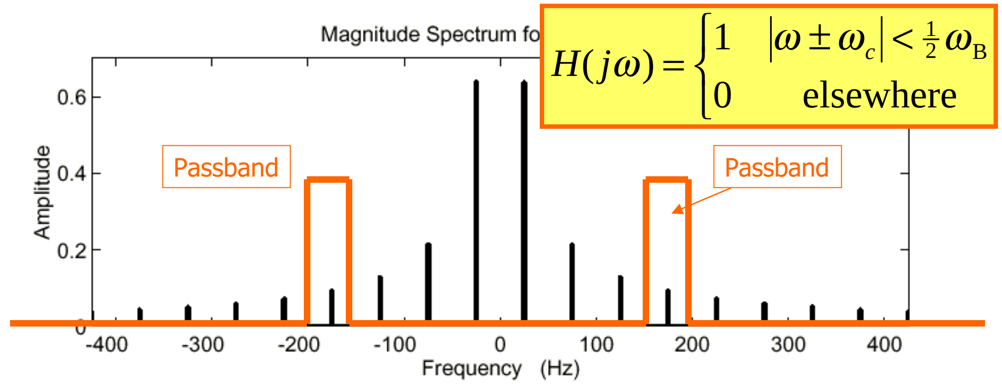


$$H(j\omega) = \begin{cases} 1 & |\omega| < \omega_{co} \\ 0 & |\omega| > \omega_{co} \end{cases}$$

$$y(t) = \frac{4}{\pi}\sin(50\pi t) + \frac{4}{3\pi}\sin(150\pi t) + \frac{4}{5\pi}\sin(250\pi t) + \frac{4}{7\pi}\sin(350\pi t)$$



Ideal Bandpass Filter



$$H(j\omega) = \begin{cases} 1 & |\omega \pm \omega_c| < \frac{1}{2}\omega_B \\ 0 & \text{elsewhere} \end{cases}$$

What is the ouput signal ?

$$y(t) = \frac{2}{j7\pi} e^{j2\pi(175)t} - \frac{2}{j7\pi} e^{-j2\pi(175)t} = \frac{4}{7\pi}\cos\left(2\pi(175)t - \tfrac{1}{2}\pi\right)$$