

Lecture 26

**Review: Digital Filtering,
Frequency Response,
& Sampling
27-April-01**

LAST LAB This Week

- ALL Lab Reports due by Friday (TODAY)
 - 4:30-5pm in VanLeer-475, or directly to TA
 - In PERSON
- Course Evaluations during last week
 - Two for GT: Lecture & Recitation
- HW #13 posted today

Final Exam Info

- Calendar: **Final Exam**
 - Period 15, Friday, May 4, 2:50pm
 - **ID check will be done at Final Exam**
- **CONFLICTS**
 - e.g., 3 exams in one day - see instructor of middle exam
- Reviews will be held on Weds and Thurs
 - 7pm in ECE Auditorium

FINAL EXAM

- FORMULA PAGES ?
 - Students bring **ONE** page **HAND-WRITTEN**
 - Tables 12.1 & 12.2 will be supplied with the exam.
- COVERAGE / EMPHASIS?
 - Fourier Transform
 - Sampling, Filtering & Spectrum
 - Digital Filters: IIR & FIR & $H(z)$, frequency response
 - Hard problems from Quizzes #2, #3.
 - Homework & Old Quizzes & Finals

LECTURE OBJECTIVES

THREE-DOMAIN APPROACH

- EXHIBIT BANDPASS FILTERS

RE-UNIFICATION:

- How does Frequency Response affect $x(t)$ to produce $y(t)$?

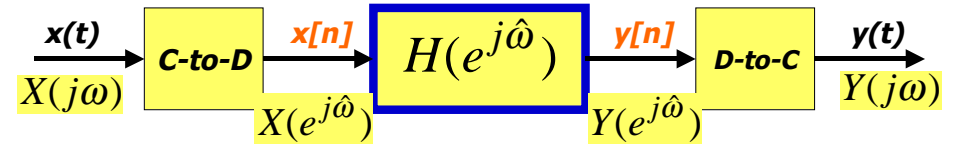


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DT Filtering of CT Signals



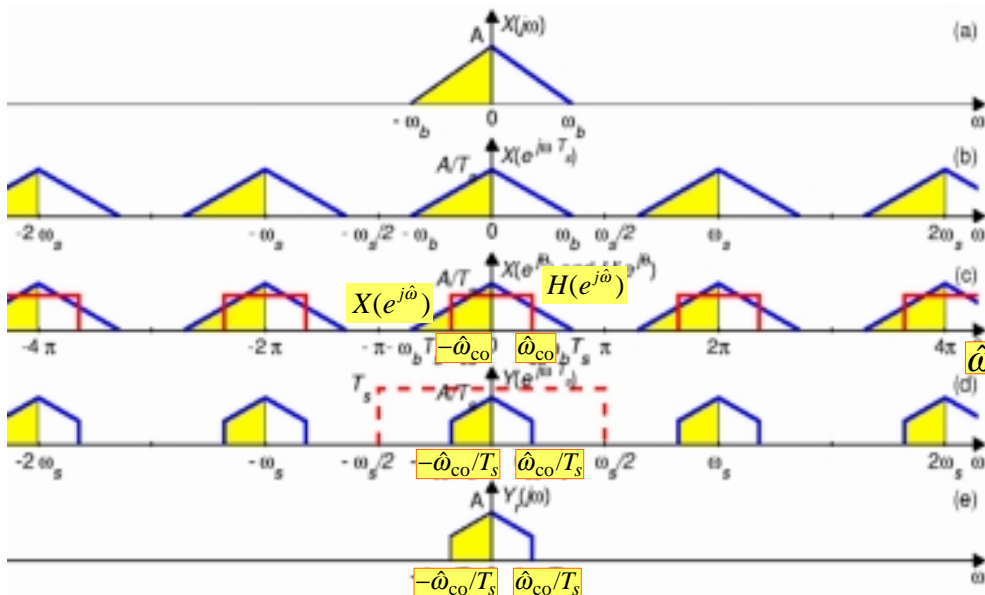
If no aliasing occurs in sampling $x(t)$, then it follows that

$$Y(j\omega) = H_{\text{eff}}(j\omega)X(j\omega)$$

$$H_{\text{eff}}(j\omega) = \begin{cases} H(e^{j\omega T_s}) & |\omega| < \frac{1}{2} \omega_s \\ 0 & |\omega| > \frac{1}{2} \omega_s \end{cases}$$

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Illustration of DT Filtering of a CT Signal



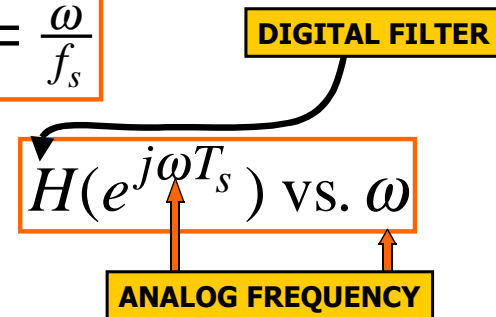
EFFECTIVE Freq. Response

- Assume NO Aliasing, then

- ANALOG FREQ \leftrightarrow DIGITAL FREQ

$$\hat{\omega} = \omega T_s = \frac{\omega}{f_s}$$

- So, we can plot:
- Scaled Freq. Axis

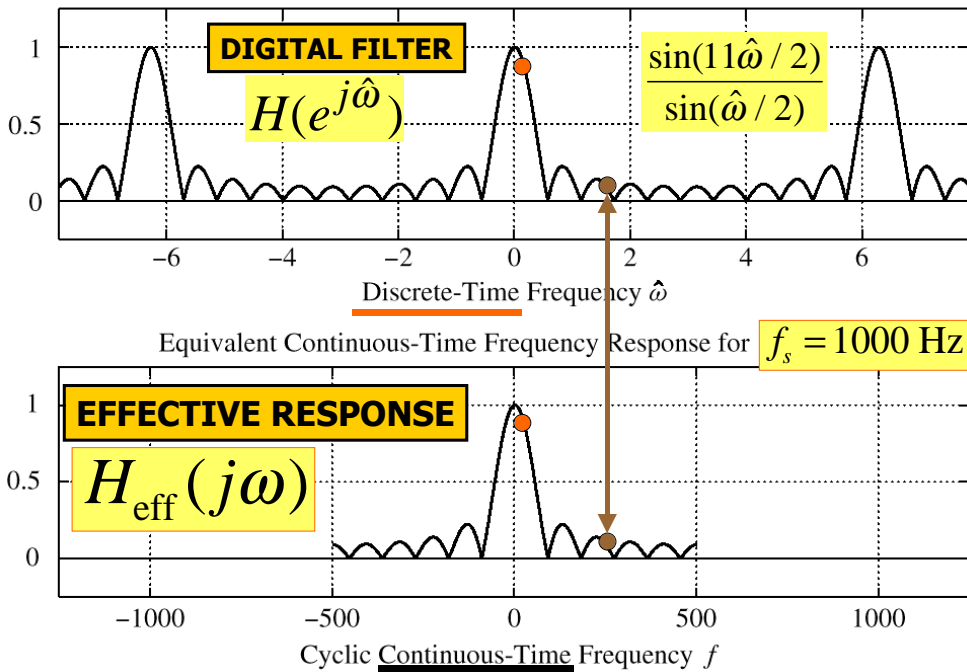


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Magnitude of Frequency Response for 11-Point Running Averager



THREE DOMAINS

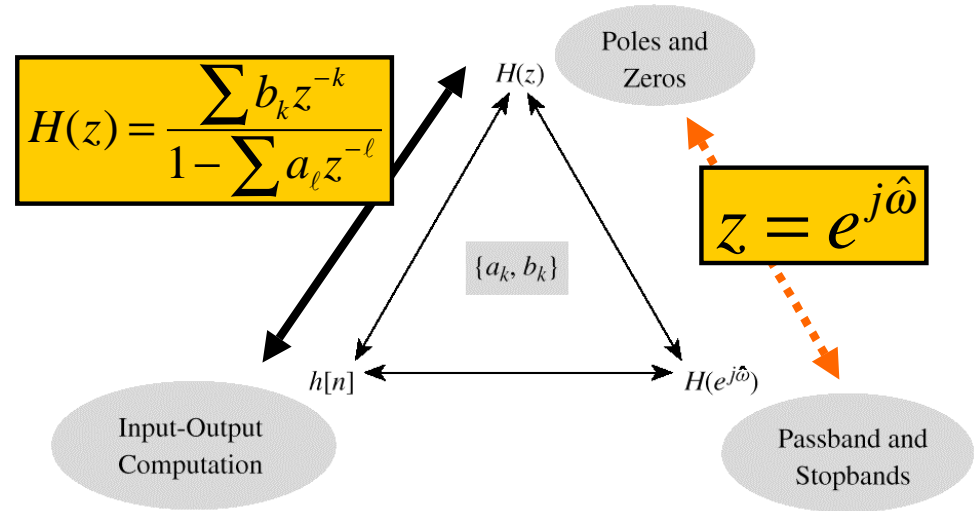
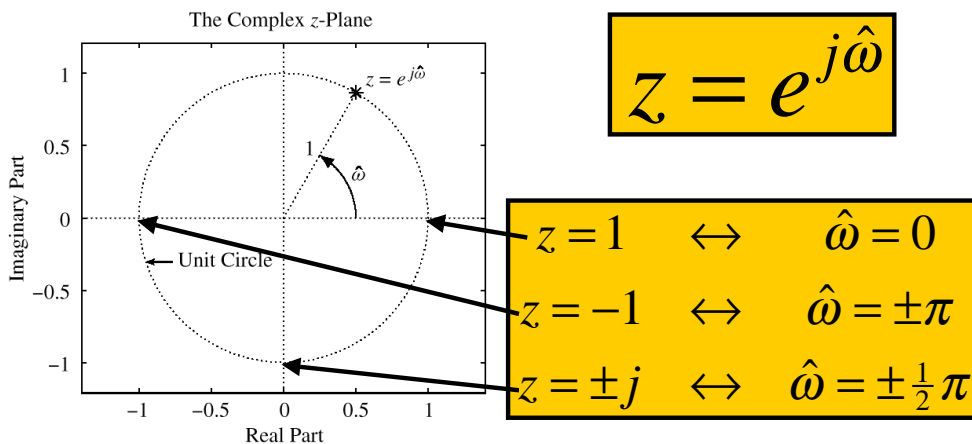


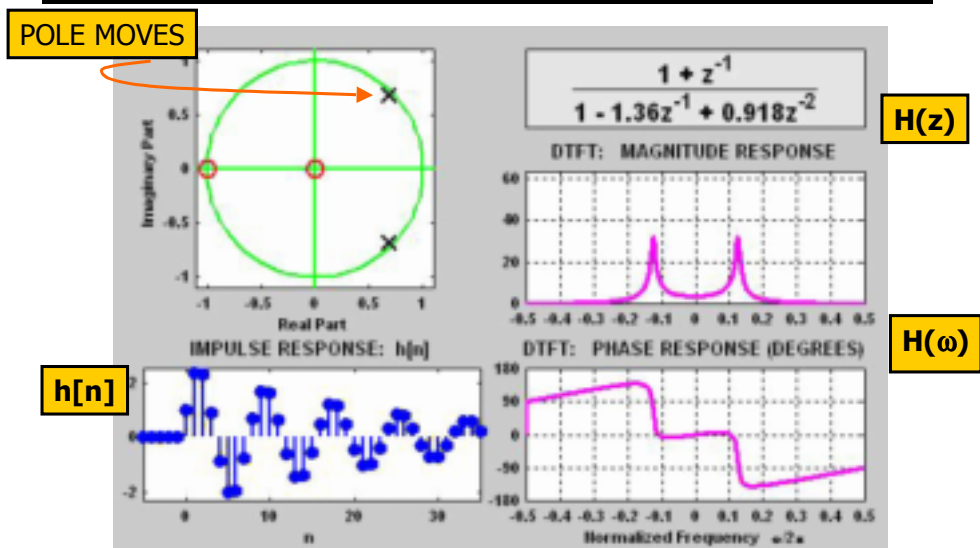
Figure 8.13 Relationship among the n -, z -, and $\hat{\omega}$ -domains. The filter coefficients $\{a_k, b_k\}$ play a central role.

UNIT CIRCLE

MAPPING BETWEEN z and $\hat{\omega}$



3 DOMAINS MOVIE: IIR



DIGITAL FILTER DESIGN

- Find the COEFFICIENTS to satisfy
 - PASSBAND & STOPBAND specifications
- FIR FILTERS
 - High Order: Many ZEROS. e.g., L=100
- IIR FILTERS
 - Poles & Zeros: 8–10 poles for a good filter
 - Implementation tricky with finite-precision

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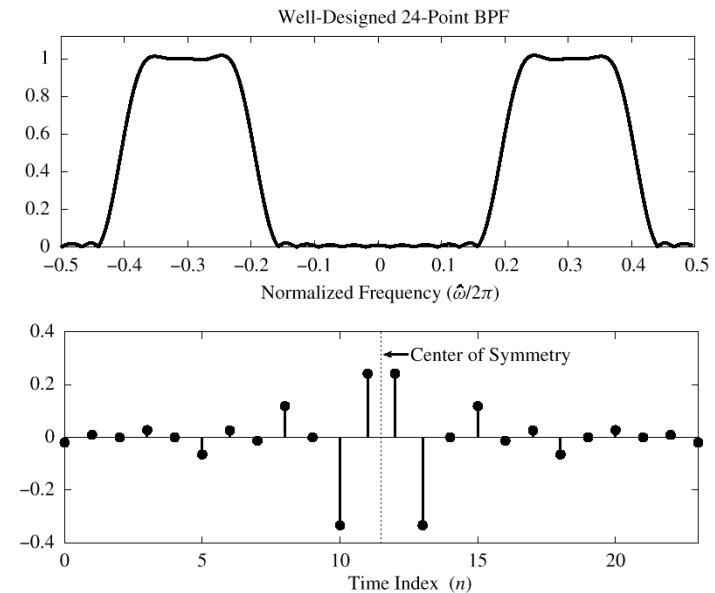
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REALISTIC FIR BANDPASS

$$H(e^{j\hat{\omega}}) = \sum_{n=0}^{23} h[n]e^{-j\hat{\omega}n}$$

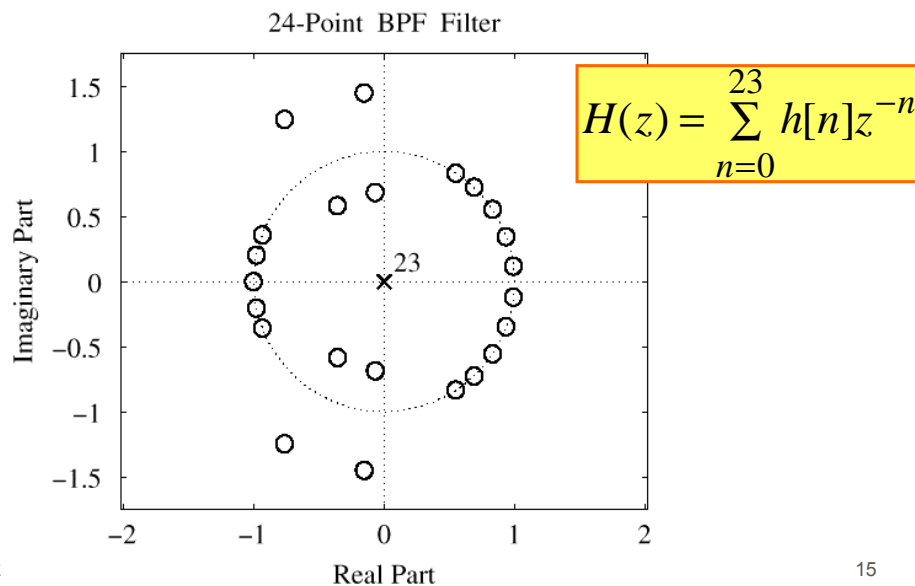
- FIR
- L = 24
- M=23
- 23 zeros

How would you make the gain = 2?



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FIR BPF: 23 ZEROS



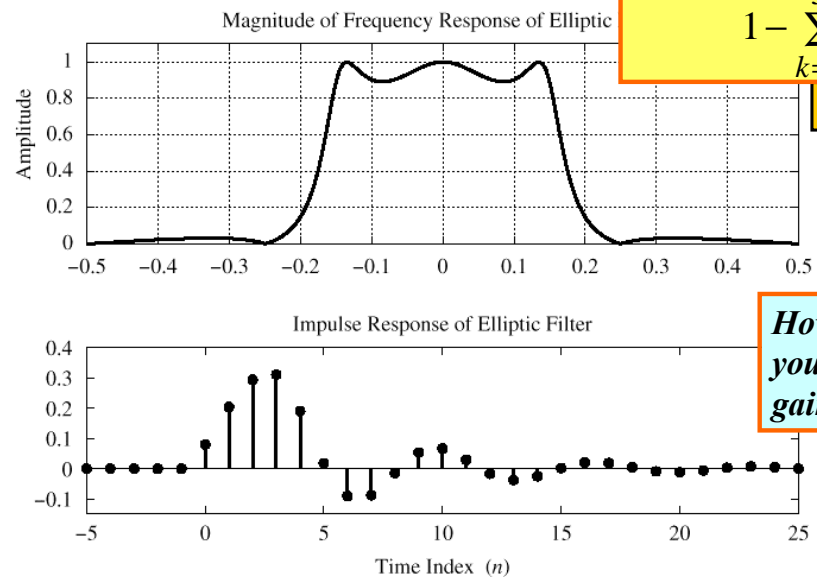
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IIR Elliptic LPF

$$H(z) = \frac{\sum_{k=0}^3 b_k z^{-k}}{1 - \sum_{k=1}^3 a_k z^{-k}}$$

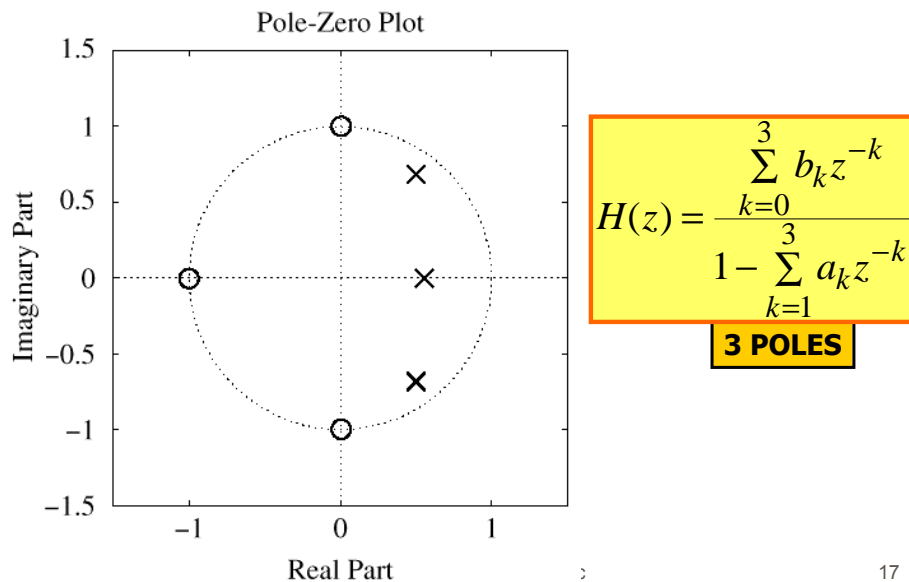
3 POLES



How would you make the gain = 2?

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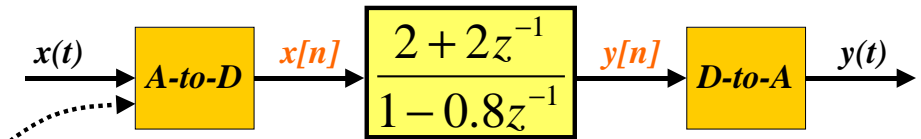
POLES & ZEROS of IIR



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Example

Given:



Find the output, $y(t)$

When $x(t) = \cos(1250\pi t)$

$f_s = 5000 \text{ Hz}$

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After sampling, the INSIDE problem becomes:

Given:

$$H(z) = \frac{2 + 2z^{-1}}{1 - 0.8z^{-1}}$$

Find the output, $y[n]$

When

$$x[n] = \cos(0.25\pi n)$$

Because

$$\omega T_s = 1250\pi / 5000 = 0.25\pi < 0.5\pi$$

NO Aliasing

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D-T SINUSOIDAL RESPONSE

$x[n] = \text{SINUSOID} \Rightarrow y[n]$ is SINUSOID

Get MAGNITUDE & PHASE from $H(z)$

if $x[n] = e^{j\hat{\omega}n}$ then

$$y[n] = H(e^{j\hat{\omega}}) e^{j\hat{\omega}n}$$

where $H(e^{j\hat{\omega}}) = H(z)|_{z=e^{j\hat{\omega}}}$

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INSIDE ANSWER

Given: $H(z) = \frac{2 + 2z^{-1}}{1 - 0.8z^{-1}}$

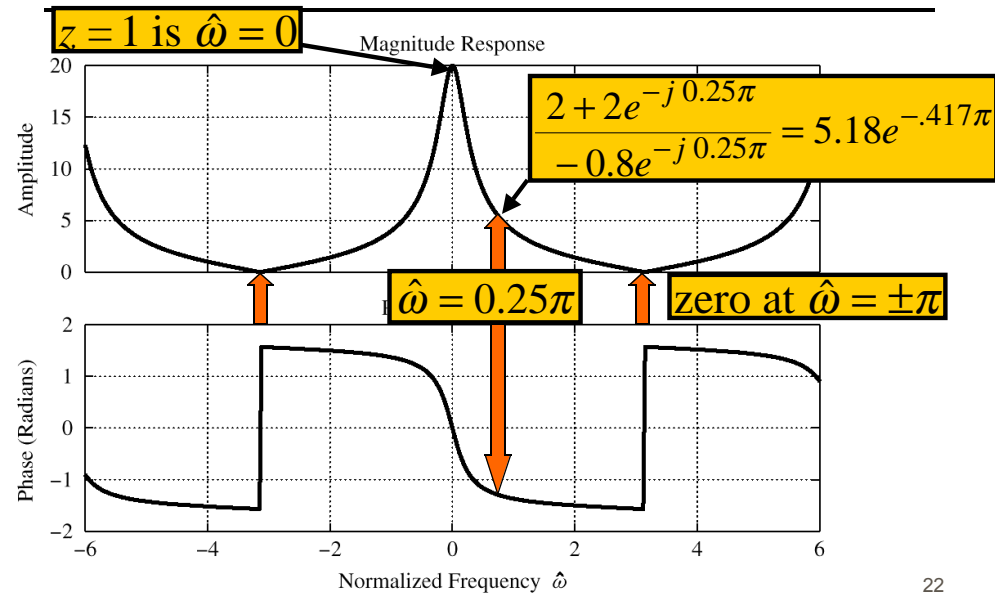
Find output, $y[n]$, when $x[n] = \cos(0.25\pi n)$

Evaluate at $z = e^{j0.25\pi}$

$$H(e^{j0.25\pi}) = \frac{2 + 2(\frac{\sqrt{2}}{2} + j\frac{\sqrt{2}}{2})}{1 - 0.8(\frac{\sqrt{2}}{2} + j\frac{\sqrt{2}}{2})} = 5.18e^{-j0.417\pi}$$

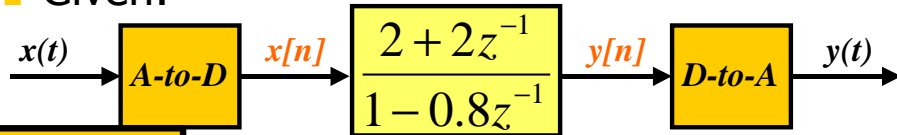
$$y[n] = 5.182 \cos(0.25\pi n - 0.417\pi)$$

Evaluate FREQ. RESPONSE



FINAL ANSWER

Given:



$f_s = 5000 \text{ Hz}$

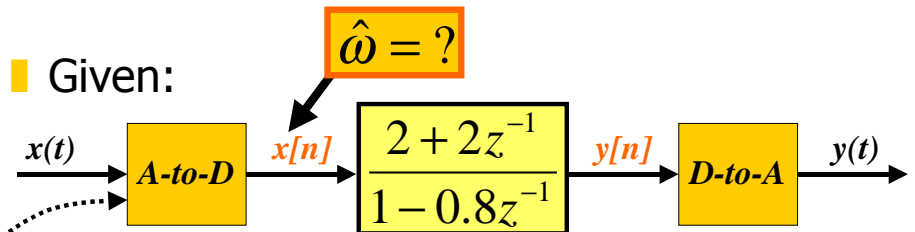
When $x(t) = \cos(1250\pi t)$

The output is

$$y(t) = 5.18 \cos(1250\pi t - 0.417\pi)$$

ANOTHER INPUT FREQ

Given:



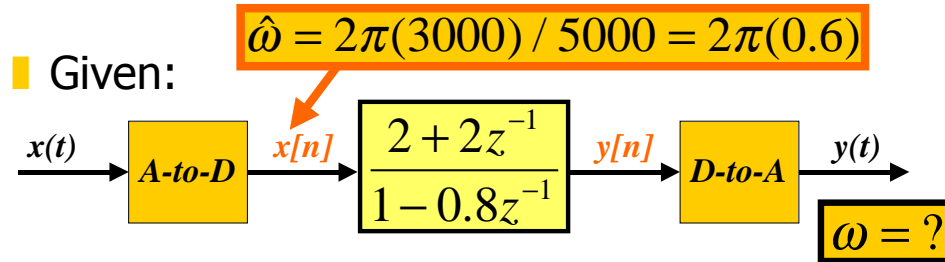
Find the output, $y(t)$

When $x(t) = \cos(2\pi(3000)t)$

$f_s = 5000 \text{ Hz}$

$\hat{\omega} = ?$

2nd PROBLEM



When $x(t) = \cos(2\pi(3000)t)$

$f_s = 5000 \text{ Hz}$ \rightarrow $\hat{\omega} = 1.2\pi$ \rightarrow $y(t) = ?$

EXAMPLE-2

Given: $H(z) = \frac{2 + 2z^{-1}}{1 - 0.8z^{-1}}$

Find the **Impulse Response**, $h[n]$

I.e., find the output, $y[n]$

When $x[n] = \delta[n]$

Z-TRANSFORM TABLES

SHORT TABLE OF z -TRANSFORMS

$x[n] \iff X(z)$

1. $ax_1[n] + bx_2[n] \iff aX_1(z) + bX_2(z)$

2. $x[n - n_0] \iff z^{-n_0} X(z)$

3. $y[n] = x[n] * h[n] \iff Y(z) = H(z)X(z)$

4. $\delta[n] \iff 1$

5. $\delta[n - n_0] \iff z^{-n_0}$

6. $a^n u[n] \iff \frac{1}{1 - az^{-1}}$

Find Inverse z-Transform

Given:

$$H(z) = \frac{2 + 2z^{-1}}{1 - 0.8z^{-1}} = \frac{2}{1 - 0.8z^{-1}} + \frac{2z^{-1}}{1 - 0.8z^{-1}}$$

Find the **Impulse Response**, $h[n]$

Use the DELAY PROPERTY

$$h[n] = 2(0.8)^n u[n] + 2(0.8)^{n-1} u[n-1]$$

EXAMPLE-3

Given:
$$H(z) = \frac{2 + 2z^{-1}}{1 - 0.8z^{-1}}$$

Find the output, $y[n]$

When

$$x[n] = (-1)^n u[n]$$

Find Inverse z-Transform

$$Y(z) = H(z)X(z)$$

$$= \frac{2 + 2z^{-1}}{1 - 0.8z^{-1}} \frac{1}{1 - z^{-1}} = \frac{2}{1 - 0.8z^{-1}}$$

Therefore:

$$y[n] = 2(0.8)^n u[n]$$

THREE INPUTS

Given:
$$H(z) = \frac{5}{1 + 0.8z^{-1}}$$

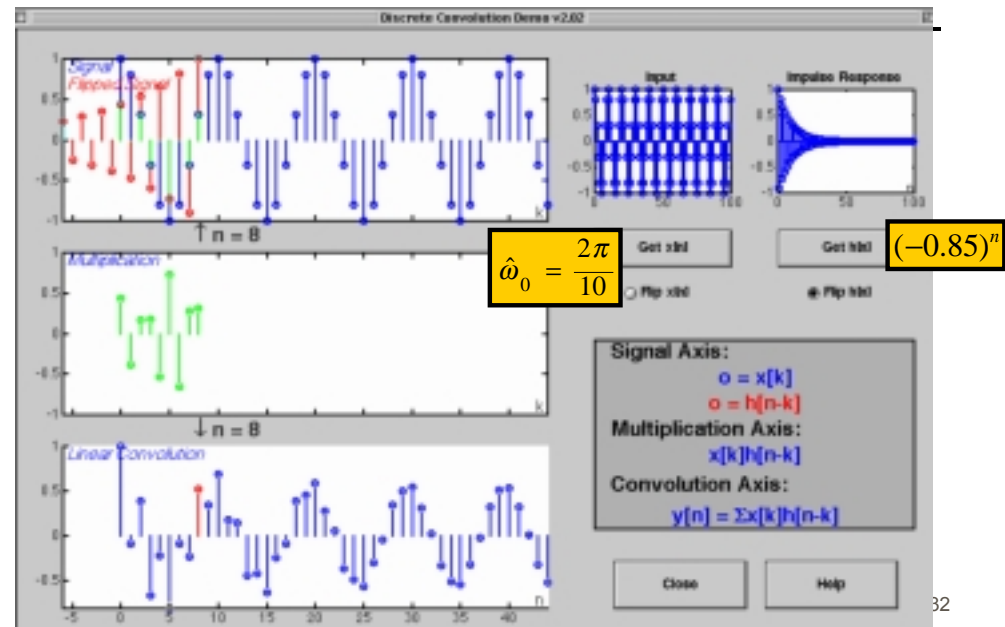
Find the output, $y[n]$

When

$$x[n] = \cos(0.2\pi n)$$

$$x[n] = \cos(0.2\pi n)u[n]$$

Transient & Steady State



Mathematical Elegance

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega$$

Fourier Synthesis
(Inverse Transform)



$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

Fourier Analysis
(Forward Transform)

Time - domain \Leftrightarrow Frequency - domain
 $x(t) \Leftrightarrow X(j\omega)$

IMPORTANT CONCEPTS

- ALL Signals have **Frequency Content**
 - Sum of Sinusoids
 - Complex Exponentials
 - Impulses, Square Pulses
- **FILTERS** alter the **Frequency Content**
 - Image Processing Example: Blur
 - Linear Time-Invariant Processing
- **3 Domains** for Analysis