

**EE-2200**

**Spring-99**

**Lecture 10**

**Linearity & Time-Invariance**

**7-May-99**

## **Info: Web-CT, Lab, HW**

- **Lab Quiz next week**
- **Quiz #2 on 24-May (Monday)**
- **Lab #5: Image Processing**
  - FM signals
  - Also Sampling and Reconstruction
- **Prob Set #4 due TODAY**

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## **READING ASSIGNMENTS**

- **This Lecture:**
  - Chapter 5, pp. 133–152
- **Other Reading:**
  - Recitation: Ch. 5, pp. 127–133, 142–146
    - **CONVOLUTION**
  - Next Lecture: Chapter 6, start

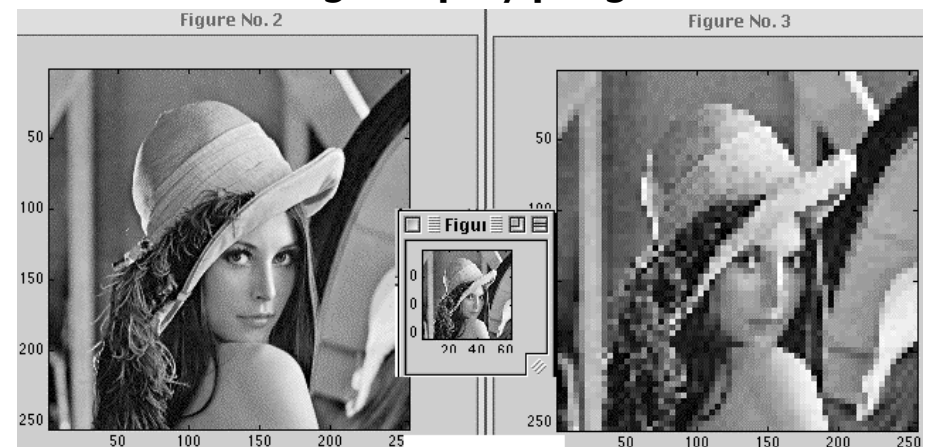
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## **LAB IMAGES (TRUE)**

- Getting **TRUE SIZE** comparisons is hard:  
use an image display program



## Image Display Procedure

- Use 256 by 256 Lenna (512 is OK)
- Make MATLAB Figures in separate windows
- **ALT-PRINT-SCREEN** captures the active window (Win-95)
- Paste into “**Paint**” program
  - Under Win-95 **Accessories**
- Print after arranging images

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## LECTURE OBJECTIVES

- **BLOCK DIAGRAM REPRESENTATION**
  - Components for **Hardware**
  - **Connect** Simple Filters Together to Build More Complicated Systems
- **GENERAL PROPERTIES of FILTERS**
  - LINEARITY
  - TIME-INVARIANCE
  - ==> CONVOLUTION

LTI SYSTEMS

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## DIGITAL FILTERING



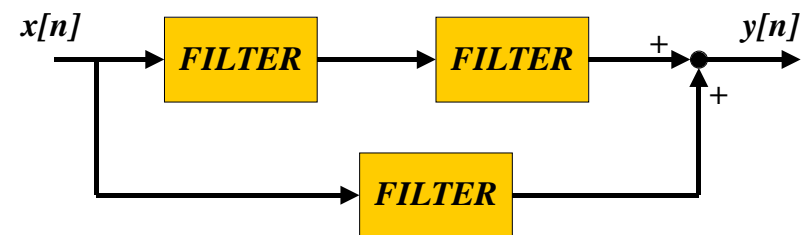
- **CONCENTRATE** on the **FILTER** (DSP)
- **DISCRETE-TIME SIGNALS**
  - FUNCTIONS of  $n$ , the “time index”
  - INPUT  $x[n]$
  - OUTPUT  $y[n]$

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## BUILDING BLOCKS



- **BUILD UP COMPLICATED FILTERS**
  - FROM SIMPLE **MODULES**
  - Ex: FILTER **MODULE** MIGHT BE 3-pt FIR

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# GENERAL FIR FILTER

## FILTER COEFFICIENTS $\{b_k\}$

### DEFINE THE FILTER

$$y[n] = \sum_{k=0}^M b_k x[n - k]$$

### For example, $\{b_k\} = \{3, -1, 2, 1\}$

$$\begin{aligned} y[n] &= \sum_{k=0}^3 b_k x[n - k] \\ &= 3x[n] - x[n - 1] + 2x[n - 2] + x[n - 3] \end{aligned}$$

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# MATLAB for FIR FILTER

$$yy = \text{filter}(bb, 1, xx)$$

### VECTOR **bb** contains Filter Coefficients

### DSP-First: $yy = \text{firfilt}(bb, xx)$

## FILTER COEFFICIENTS $\{b_k\}$

**Conv2 ( )  
for images**

$$y[n] = \sum_{k=0}^M b_k x[n - k]$$

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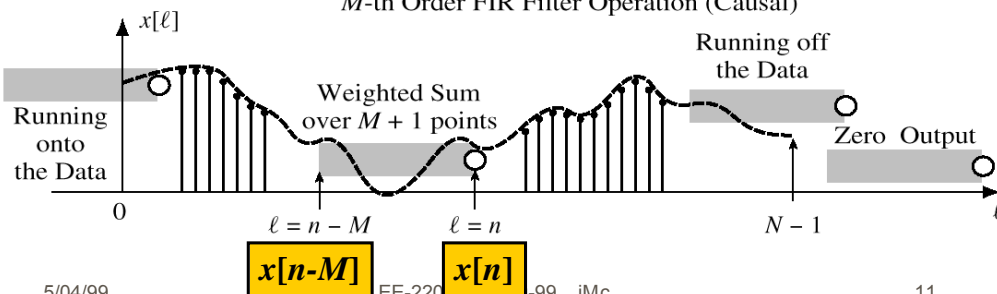
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# GENERAL FIR FILTER

## SLIDE a Length-L WINDOW over $x[n]$

$$y[n] = \sum_{k=0}^M b_k x[n - k]$$

M-th Order FIR Filter Operation (Causal)



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# FILTERING EXAMPLE

## 7-point AVERAGER

$$y_7[n] = \frac{1}{7} \left( \sum_{k=0}^6 x[n - k] \right)$$

### Removes cosine

### By making its amplitude (A) smaller

## 3-point AVERAGER

$$y_3[n] = \frac{1}{3} \left( \sum_{k=0}^2 x[n - k] \right)$$

### Changes A slightly

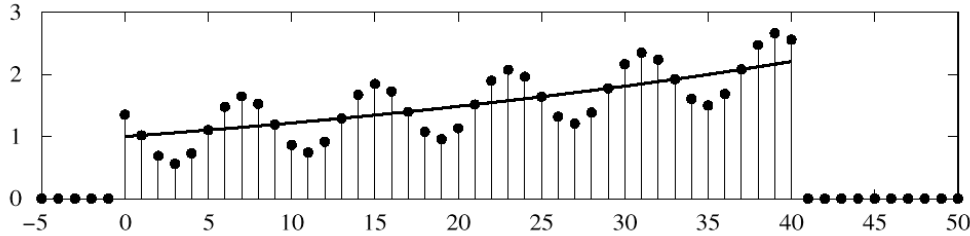
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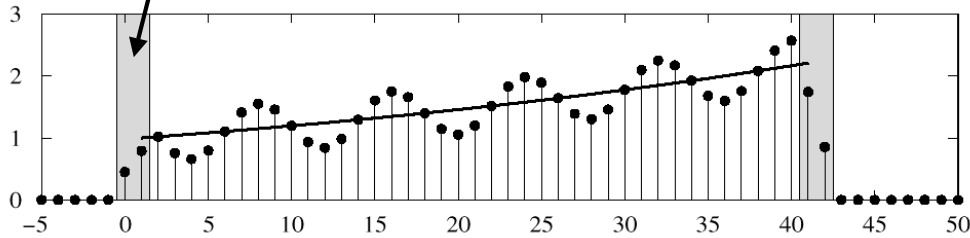
# 3-pt AVG EXAMPLE

Input Signal:  $x[n] = (1.02)^n + \cos(2\pi n/8 + \pi/4)$  for  $0 \leq n \leq 40$



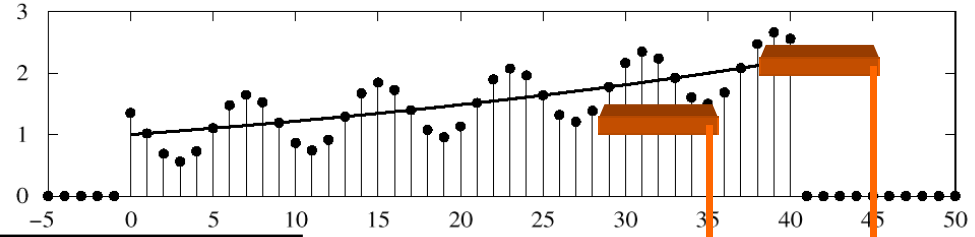
USE PAST VALUES

Output of 3-Point Running-Average Filter



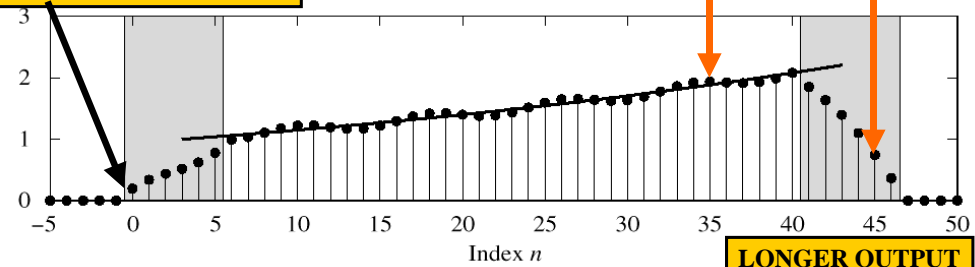
# 7-pt FIR EXAMPLE (AVG)

Input Signal:  $x[n] = (1.02)^n + \cos(2\pi n/8 + \pi/4)$  for  $0 \leq n \leq 40$



CAUSAL: Use Previous

Output of 7-Point Running-Average Filter

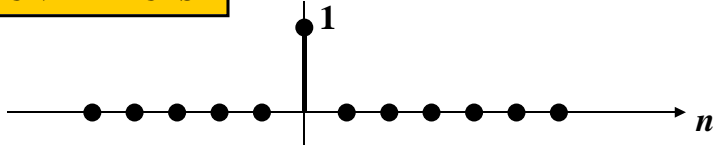


# SPECIAL INPUT SIGNALS

- $x[n] = \text{SINUSOID}$  FREQUENCY RESPONSE
- $x[n]$  has only one NON-ZERO VALUE

$$\delta[n] = \begin{cases} 1 & n = 0 \\ 0 & n \neq 0 \end{cases}$$

UNIT-IMPULSE



# UNIT IMPULSE SIGNAL $\delta[n]$

|               |     |    |    |   |   |   |   |   |   |   |     |
|---------------|-----|----|----|---|---|---|---|---|---|---|-----|
| $n$           | ... | -2 | -1 | 0 | 1 | 2 | 3 | 4 | 5 | 6 | ... |
| $\delta[n]$   | 0   | 0  | 0  | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0   |
| $\delta[n-3]$ | 0   | 0  | 0  | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0   |

n=3

$\delta[n]$  is NON-ZERO  
When its argument  
is equal to ZERO

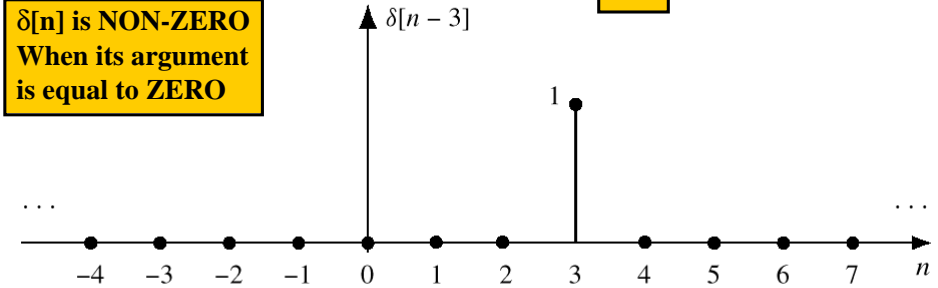
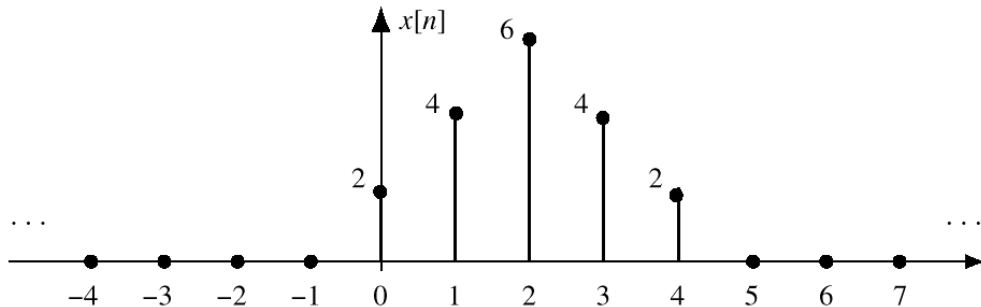


Figure 5.7 Shifted impulse sequence,  $\delta[n-3]$ .

# MATH FORMULA for $x[n]$

Use **SHIFTED IMPULSES** to write  $x[n]$

$$x[n] = 2\delta[n] + 4\delta[n - 1] + 6\delta[n - 2] + 4\delta[n - 3] + 2\delta[n - 4]$$



# SUM of **SHIFTED IMPULSES**

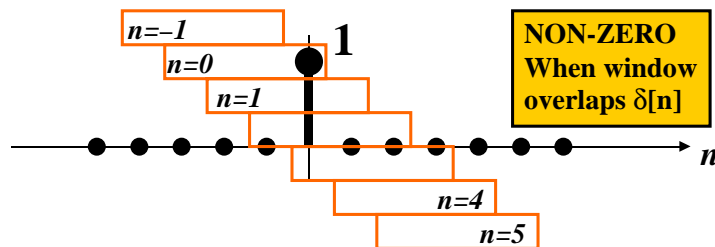
| $n$              | ... | -2 | -1 | 0 | 1 | 2 | 3 | 4 | 5 | 6 | ... |
|------------------|-----|----|----|---|---|---|---|---|---|---|-----|
| $2\delta[n]$     | 0   | 0  | 0  | 2 | 0 | 0 | 0 | 0 | 0 | 0 | 0   |
| $4\delta[n - 1]$ | 0   | 0  | 0  | 0 | 4 | 0 | 0 | 0 | 0 | 0 | 0   |
| $6\delta[n - 2]$ | 0   | 0  | 0  | 0 | 0 | 6 | 0 | 0 | 0 | 0 | 0   |
| $4\delta[n - 3]$ | 0   | 0  | 0  | 0 | 0 | 0 | 4 | 0 | 0 | 0 | 0   |
| $2\delta[n - 4]$ | 0   | 0  | 0  | 0 | 0 | 0 | 0 | 2 | 0 | 0 | 0   |
| $x[n]$           | 0   | 0  | 0  | 2 | 4 | 6 | 4 | 2 | 0 | 0 | 0   |

$$x[n] = \sum_k x[k]\delta[n - k] \leftarrow \text{This formula ALWAYS works}$$

$$= \dots + x[-1]\delta[n + 1] + x[0]\delta[n] + x[1]\delta[n - 1] + \dots \quad (5.3.6)$$

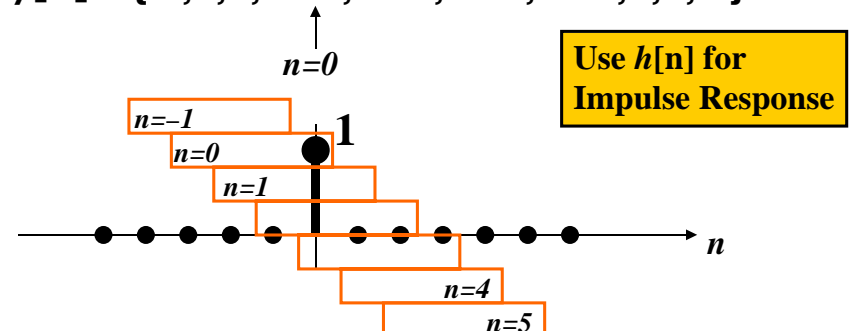
# 4-pt AVERAGER

- CAUSAL SYSTEM: USE PAST VALUES
- INPUT = UNIT IMPULSE SIGNAL =  $\delta[n]$
- OUTPUT is called **"IMPULSE RESPONSE"**



# 4-pt Avg Impulse Response

- $y[n] = 0.25(x[n] + x[n-1] + x[n-2] + x[n-3])$
- "READS OUT" the FILTER COEFFICIENTS**
- $y[n] = \{\dots, 0, 0, 0.25, 0.25, 0.25, 0.25, 0, 0, \dots\}$



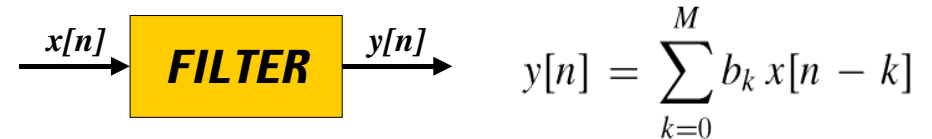
# FIR IMPULSE RESPONSE

- Convolution = Filter Definition
- Filter Coeffs = Impulse Response

| $n$                | $n < 0$ | 0     | 1     | 2     | 3     | ... | $M$   | $M + 1$ | $n > M + 1$ |
|--------------------|---------|-------|-------|-------|-------|-----|-------|---------|-------------|
| $x[n] = \delta[n]$ | 0       | 1     | 0     | 0     | 0     | 0   | 0     | 0       | 0           |
| $y[n] = h[n]$      | 0       | $b_0$ | $b_1$ | $b_2$ | $b_3$ | ... | $b_M$ | 0       | 0           |

$$y[n] = \sum_{k=0}^M h[k] x[n - k]$$

# HARDWARE STRUCTURES

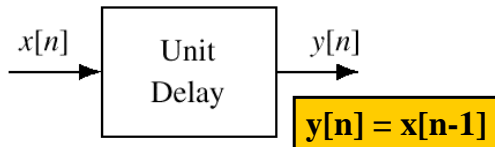
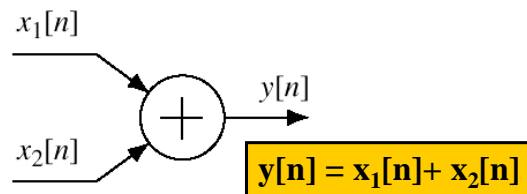
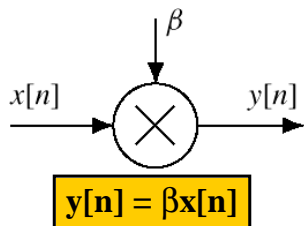


- INTERNAL STRUCTURE of “FILTER”
  - WHAT COMPONENTS ARE NEEDED?
  - HOW DO WE “HOOK” THEM TOGETHER?
- SIGNAL FLOW GRAPH NOTATION

# HARDWARE ATOMS

- Add, Multiply & Store

$$y[n] = \sum_{k=0}^M b_k x[n - k]$$



# FIR STRUCTURE

- Direct Form

$$y[n] = \sum_{k=0}^M b_k x[n - k]$$

SIGNAL FLOW GRAPH

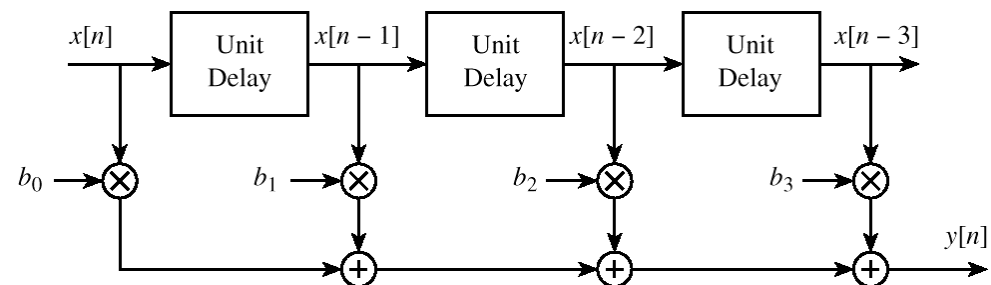


Figure 5.13 Block-diagram structure for the  $M$ th order FIR filter.

# ALTERNATE STRUCTURE

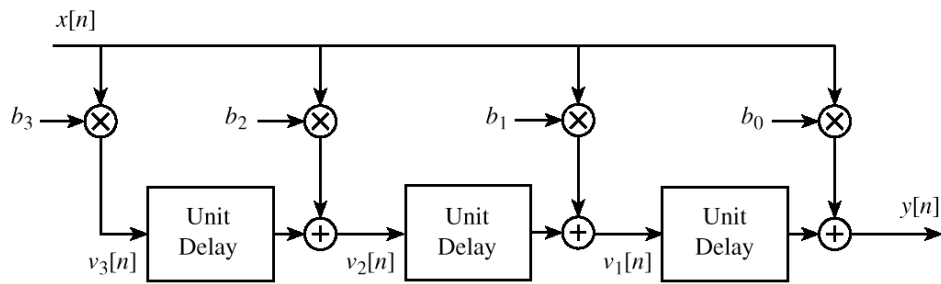


Figure 5.14 Transposed form block diagram structure for the  $M$ th order FIR filter.

$$y[n] = b_0x[n] + v_1[n - 1]$$

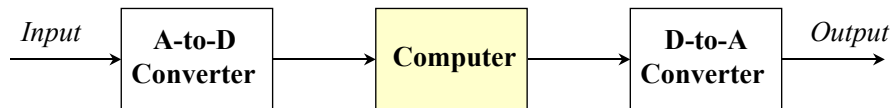
$$v_1[n] = b_1x[n] + v_2[n - 1]$$

$$v_2[n] = b_2x[n] + v_3[n - 1]$$

$$v_3[n] = b_3x[n]$$

**DIFFERENT COMPUTATION**

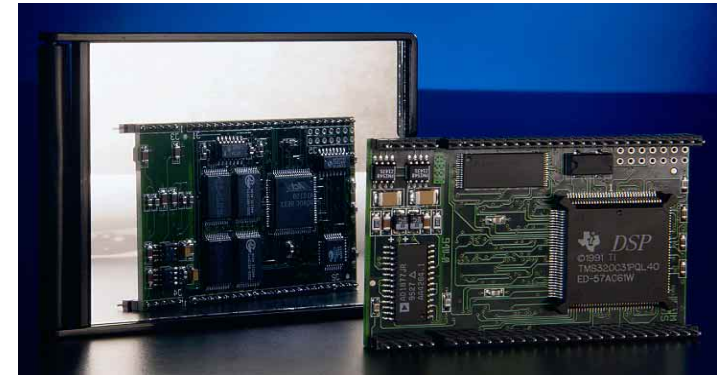
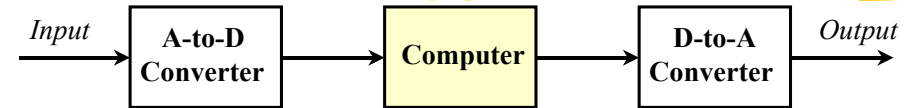
# One View of DSP, c. 1976



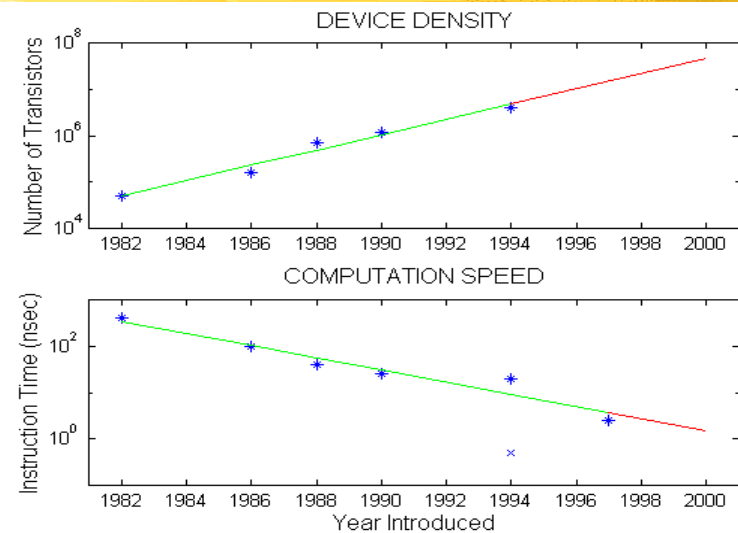
■ “That discipline which has allowed us to replace a circuit previously composed of a capacitor and a resistor with two anti-aliasing filters, an A-to-D and a D-to-A converter, and a general purpose computer (or array processor) so long as the signal we are interested in does not vary too quickly.”

Thomas P. Barnwell, III

# The Essence of DSP



# Moore's Law for TI DSPs



# SYSTEM PROPERTIES



## TIME-INVARIANCE

## LINEARITY

## CAUSALITY

- “No output prior to input”

# TIME-INVARIANCE

## IDEA:

- “Time-Shifting the input will cause the **same** time-shift in the output”

## EQUIVALENTLY,

- We can prove that
  - The time origin ( $n=0$ ) is arbitrary

# TESTING Time-Invariance

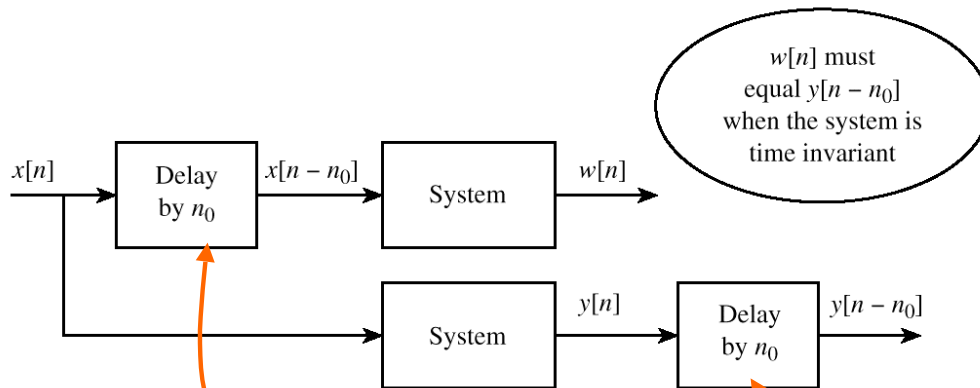


Figure 5.16 Testing time-invariance property by checking the interchange of operations.

# LINEAR SYSTEM

## LINEARITY = Two Properties

## SCALING

- “Doubling  $x[n]$  will double  $y[n]$ ”

## SUPERPOSITION:

- “Adding two inputs gives an output that is the sum of the individual outputs”



# TESTING LINEARITY

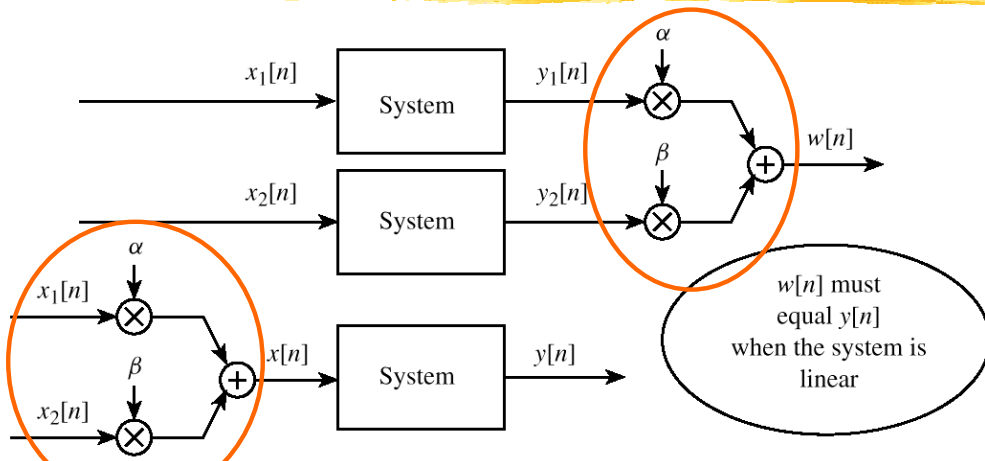


Figure 5.17 Testing linearity by checking the interchange of operations.

# LTI SYSTEMS

- LTI: **L**inear & **T**ime-**I**nvariant
- COMPLETELY CHARACTERIZED by:
  - **CONVOLUTION**:  $y[n] = x[n]*h[n]$
  - **IMPULSE RESPONSE**  $h[n]$ 
    - The “rule” can be re-written as convolution
- FIR Example:  $h[n]$  is same as  $b_k$

# LTI: Convolution

- Output = Convolution of  $x[n]$  &  $h[n]$ 
  - NOTATION:  $y[n] = x[n]*h[n]$
  - Here is the FIR case:

$$y[n] = \sum_{k=0}^M h[k] x[n - k]$$

FINITE LIMITS

Same as  $b_k$

FINITE LIMITS

# CONVOLUTION Example

| n          | 0 | 1  | 2  | 3  | 4  | 5  | 6 | 7 | 8 |
|------------|---|----|----|----|----|----|---|---|---|
| x[n]       | 2 | 4  | 6  | 4  | 2  |    |   |   |   |
| h[n]       | 3 | -1 | 2  | 1  |    |    |   |   |   |
| h[0]x[n-0] | 6 | 12 | 18 | 12 | 6  |    |   |   |   |
| h[1]x[n-1] |   | -2 | -4 | -6 | -4 | -2 |   |   |   |
| h[2]x[n-2] |   |    | 4  | 8  | 12 | 8  | 4 |   |   |
| h[3]x[n-3] |   |    |    | 2  | 4  | 6  | 4 | 2 |   |
| y[n]       | 6 | 10 | 18 | 16 | 18 | 12 | 8 | 2 |   |

# CASCADE SYSTEMS

- Does the order of  $S_1$  &  $S_2$  matter?
  - NO, LTI SYSTEMS can be rearranged !!!
  - WHAT ARE THE FILTER COEFFS?  $\{b_k\}$

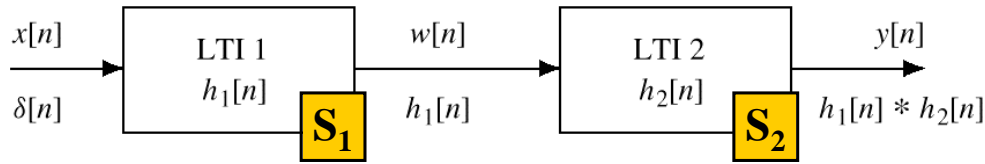


Figure 5.19 A Cascade of Two LTI Systems.

# CASCADE EQUIVALENT

- Find “overall”  $h[n]$  for a cascade ?

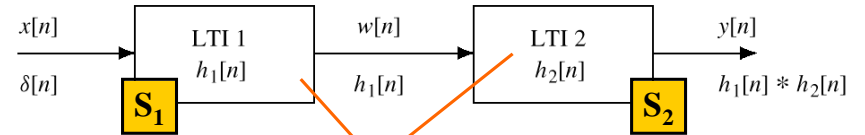


Figure 5.19 A Cascade of Two LTI Systems.

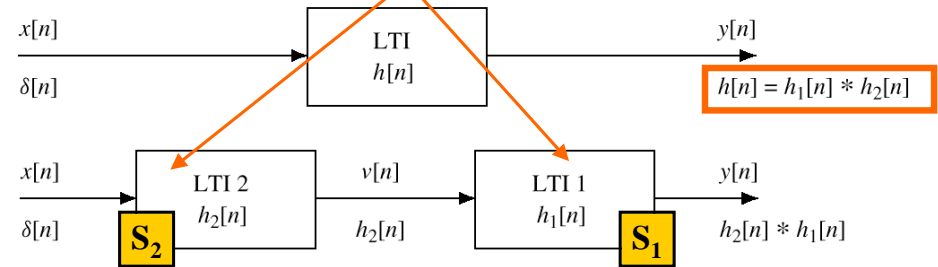


Figure 5.20 Switching the order of cascaded LTI systems.