

EE-2025

Fall-2001

Lecture 23

IIR Filters: Feedback & $H(z)$

26-Nov-01

Info: HW and Labs

- Reading Assignment:
 - Chapter 8 in DSP First
- Prob Set #11 - **due this week**
 - Prob Set #12 next week
- Lab #11 due this week of 26-Nov
 - Lab #12 due next week (3 — 6 Dec)
 - ALL Labs must be turned in by 5pm on 7-Dec-01

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Final Exam Info

- Calendar: **Final Exam(s)**
 - Period 3, Monday, 10-Dec @ 2:50 pm
 - NOON Lecture
 - **Must take exam with your assigned section**
 - Period 11, Thurs, 13-Dec @ 11:30am
 - 11 AM Lecture
- **Report CONFLICTS immediately !!!!**
 - e.g., 3 exams in one day
- Reviews will be held on Sunday & Wednesday
 - 6:00 PM in ECE Auditorium

Grades

- Quiz #3 Results
 - **Median = 78**; average = 74.8
 - Nine 100's
- Check your own grades for accuracy
- Cumulative Averages are now posted
 - Lowest HW is dropped now.
 - Max = **70** (at this moment)
 - Final will be **25** (max)
 - Recitation Points will be **5** (max)

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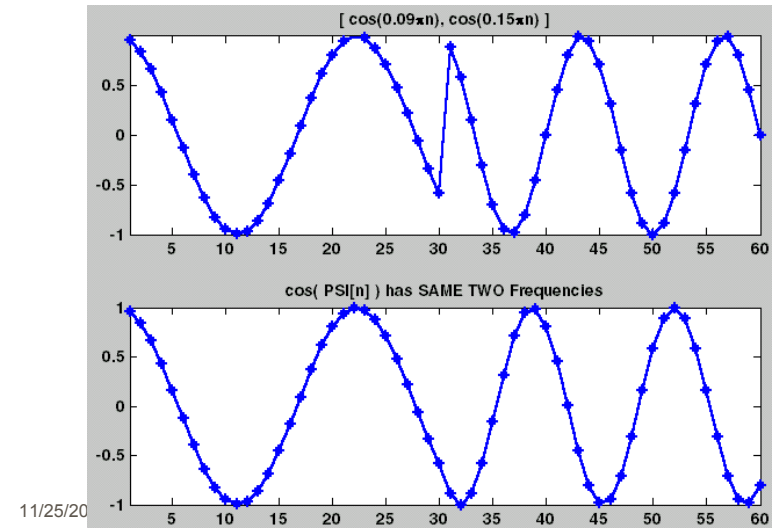
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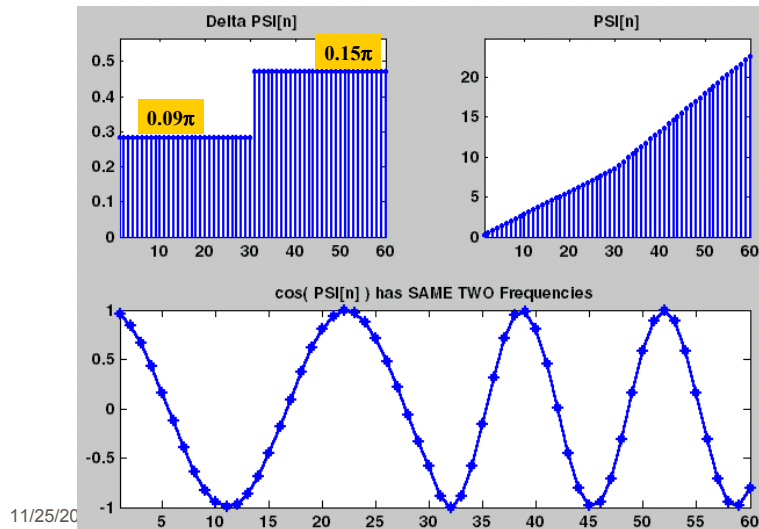
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Labs 11 & 12
FSK synthesis & demodulation
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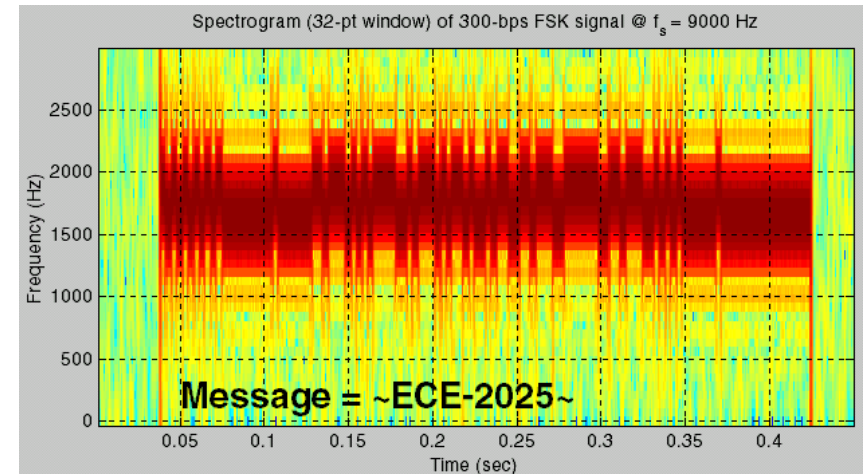
FSK: continuous phase



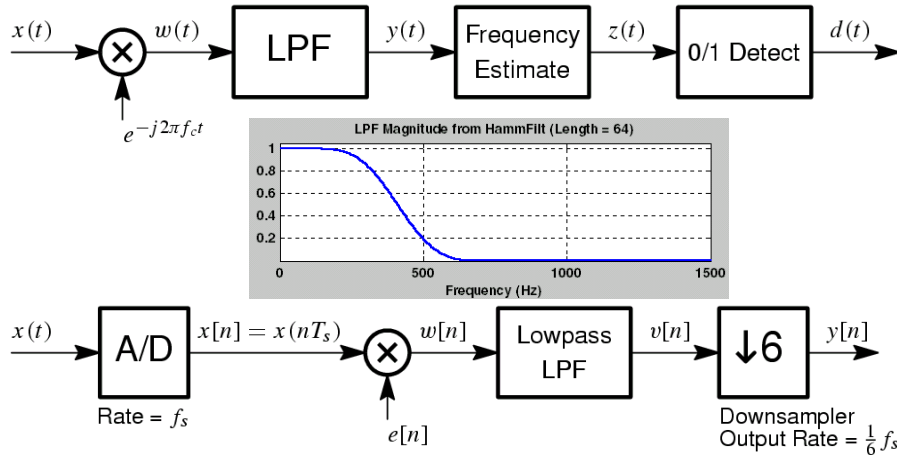
FSK: use cos(psi[n])



FSK Spectrogram



FSK demodulator



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Anti-Aliasing Filter

- Given samples of $w(t)$ at $f_s=9000$ Hz, it is possible to downsample to a lower rate
 - Must obey **SAMPLING THEOREM**
 - NEW $f_s >$ twice the highest frequency
- $w[n]$ is $w(t)$ sampled at $f_s=9000$ Hz
- Want $w(t)$ sampled at $f_s=1500$ Hz
 - Use **DIGITAL LPF** with cutoff at 750 Hz

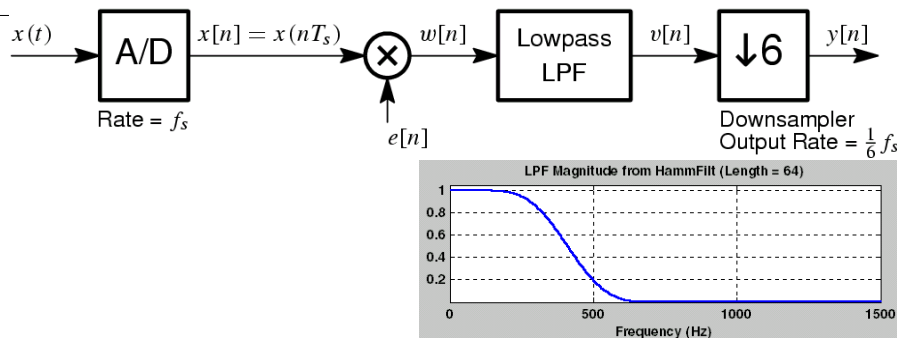
$$\hat{\omega}_c = 2\pi \frac{f_c}{f_s} = 2\pi \frac{750}{9000} = \frac{\pi}{6}$$

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Anti-Aliasing Filter



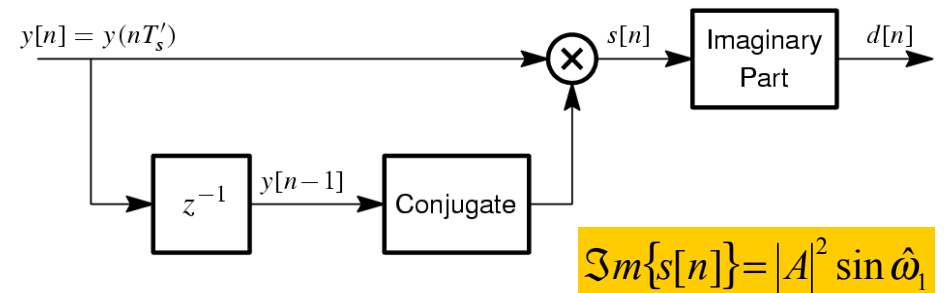
- $y[n]$ is every sixth sample of $v[n]$
 - i.e., samples of $w(t)$ at $f_s=1500$ Hz
- LPF must remove all freqs above 750 Hz

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SLICER produces Frequency



$$s[n] = Ae^{j\hat{\omega}_1 n} (Ae^{j\hat{\omega}_1 (n-1)})^*$$

$$= |A|^2 e^{j\hat{\omega}_1 n} e^{-j\hat{\omega}_1 (n-1)} = |A|^2 e^{j\hat{\omega}_1}$$

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READING ASSIGNMENTS

■ This Lecture:

- Chapter 8, pp. 249-263

■ Other Reading:

- Recitation: Ch. 8, pp. 261-272
 - POLES & ZEROS
- Next Lecture: Chapter 8, pp. 269-282

LECTURE OBJECTIVES

■ INFINITE IMPULSE RESPONSE FILTERS

- Define **IIR** DIGITAL Filters
- Have **FEEDBACK**: use PREVIOUS OUTPUTS

$$y[n] = \sum_{\ell=1}^N a_{\ell} y[n-\ell] + \sum_{k=0}^M b_k x[n-k]$$

- Show how to compute the output $y[n]$
 - FIRST-ORDER CASE (N=1)
 - Z-transform: Impulse Response $h[n] \leftrightarrow H(z)$

Quick Review: L-pt Averager

$$y[n] = \sum_{k=0}^{L-1} \frac{1}{L} x[n-k]$$

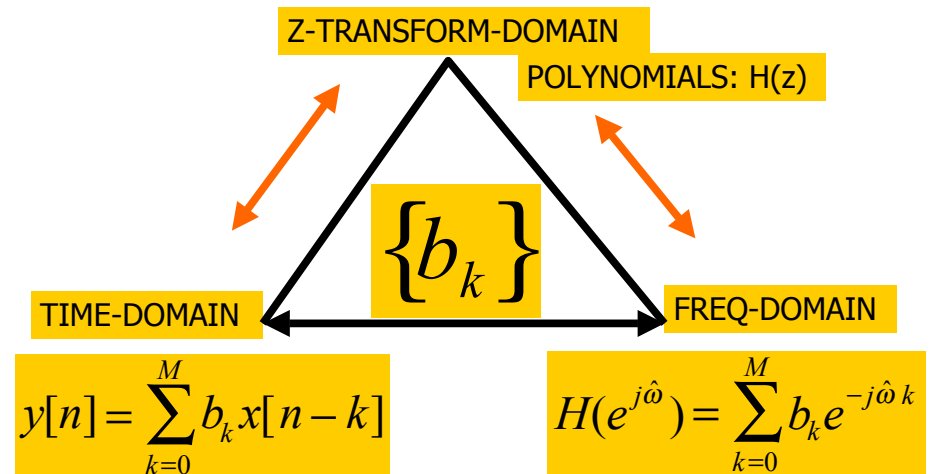
IMPULSE RESPONSE

$$h[n] = \sum_{k=0}^{L-1} \frac{1}{L} \delta[n-k]$$

SYSTEM FUNCTION

$$H(z) = \sum_{n=0}^{L-1} \frac{1}{L} z^{-n}$$

THREE DOMAINS



LOGICAL THREAD

FIND the IMPULSE RESPONSE, $h[n]$

INFINITELY LONG

IIR Filters

$$H(z) = \sum_{n=0}^{\infty} h[n]z^{-n}$$

EXPLOIT THREE DOMAINS:

Show Relationship for IIR:

$$h[n] \leftrightarrow H(z) \leftrightarrow H(e^{j\hat{\omega}})$$

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ONE FEEDBACK TERM

ADD PREVIOUS OUTPUTS

$$y[n] = a_1y[n-1] + b_0x[n] + b_1x[n-1]$$

PREVIOUS
FEEDBACK

FIR PART of the FILTER
FEED-FORWARD

CAUSALITY

NOT USING FUTURE OUTPUTS or INPUTS

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FILTER COEFFICIENTS

ADD PREVIOUS OUTPUTS

$$y[n] = 0.8y[n-1] + 3x[n] - 2x[n-1]$$

FEEDBACK COEFFICIENT

SIGN CHANGE

MATLAB

`yy = filter([3,-2],[1,-0.8],xx)`

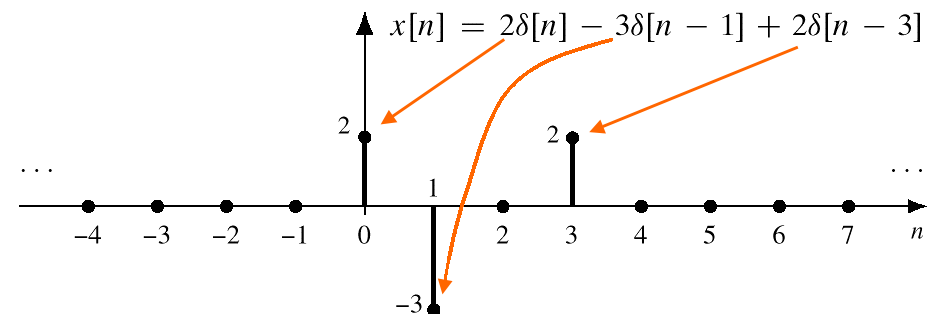
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COMPUTE OUTPUT

$$y[n] = 0.8y[n-1] + 5x[n]$$



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COMPUTE $y[n]$

- FEEDBACK DIFFERENCE EQUATION:

$$y[n] = 0.8y[n-1] + 5x[n]$$

- NEED $y[-1]$ to get started

$$y[0] = 0.8y[-1] + 5x[0]$$

AT REST CONDITION

- $y[n] = 0$, for $n < 0$
- BECAUSE $x[n] = 0$, for $n < 0$

INITIAL REST CONDITIONS

- The input must be assumed to be zero prior to some starting time n_0 , i.e., $x[n] = 0$ for $n < n_0$. We say that such inputs are *suddenly applied*.
- The output is likewise assumed to be zero prior to the starting time of the signal, i.e., $y[n] = 0$ for $n < n_0$. We say that the system is *initially at rest* if its output is zero prior to the application of a suddenly applied input.

COMPUTE $y[0]$

- THIS STARTS THE RECURSION:

With the initial rest assumption, $y[n] = 0$ for $n < 0$,
 $y[0] = 0.8y[-1] + 5(2) = 0.8(0) + 5(2) = 10$

- SAME with MORE FEEDBACK TERMS

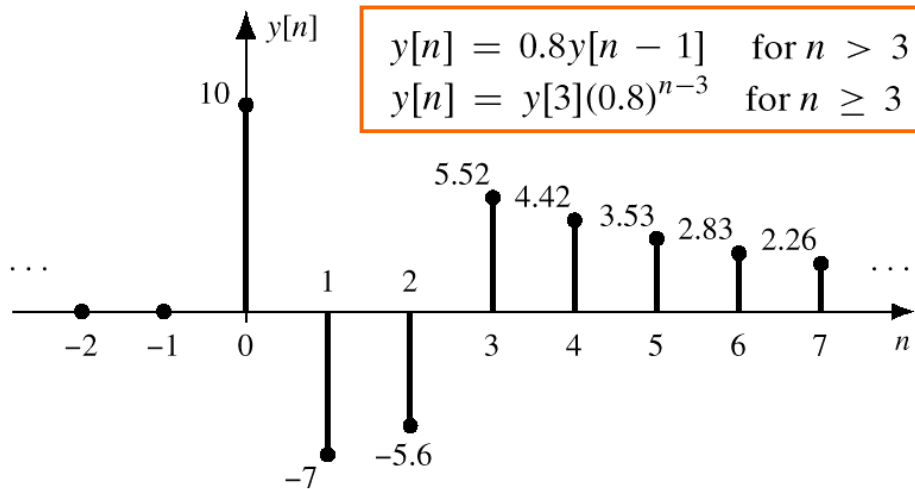
$$y[n] = a_1y[n-1] + a_2y[n-2] + \sum_{k=0}^2 b_kx[n-k]$$

COMPUTE MORE $y[n]$

- CONTINUE THE RECURSION:

$$\begin{aligned}y[1] &= 0.8y[0] + 5x[1] = 0.8(10) + 5(-3) = -7 \\y[2] &= 0.8y[1] + 5x[2] = 0.8(-7) + 5(0) = -5.6 \\y[3] &= 0.8y[2] + 5x[3] = 0.8(-5.6) + 5(2) = 5.52 \\y[4] &= 0.8y[3] + 5x[4] = 0.8(5.52) + 5(0) = 4.416 \\y[5] &= 0.8y[4] + 5x[5] = 0.8(4.416) + 5(0) = 3.5328 \\y[6] &= 0.8y[5] + 5x[6] = 0.8(3.5328) + 5(0) = 2.8262\end{aligned}$$

PLOT $y[n]$



IMPULSE RESPONSE

$$h[n] = a_1 h[n - 1] + b_0 \delta[n]$$

n	$n < 0$	0	1	2	3	4
$\delta[n]$	0	1	0	0	0	0
$h[n - 1]$	0	0	b_0	$b_0(a_1)$	$b_0(a_1)^2$	$b_0(a_1)^3$
$h[n]$	0	b_0	$b_0(a_1)$	$b_0(a_1)^2$	$b_0(a_1)^3$	$b_0(a_1)^4$

From this table it is obvious that the general formula is

$$h[n] = \begin{cases} b_0(a_1)^n & \text{for } n \geq 0 \\ 0 & \text{for } n < 0 \end{cases}$$

$$h[n] = b_0(a_1)^n u[n]$$

$$u[n] = 1, \text{ for } n \geq 0$$

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IMPULSE RESPONSE

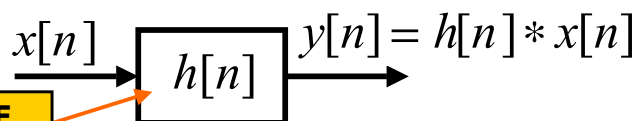
DIFFERENCE EQUATION:

$$y[n] = 0.8y[n - 1] + 3x[n]$$

Find $h[n]$

$$h[n] = 3(0.8)^n u[n]$$

CONVOLUTION in TIME-DOMAIN



IMPULSE RESPONSE

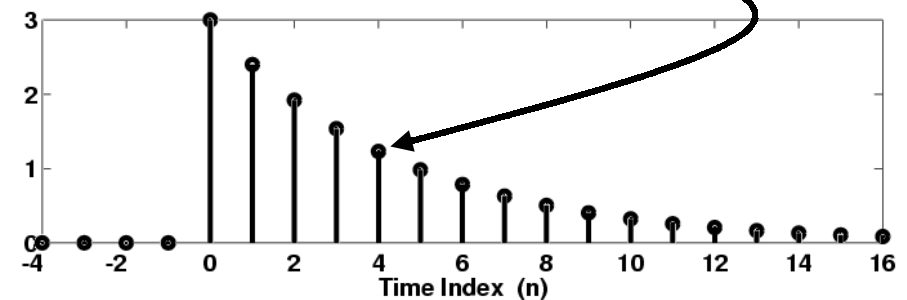
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PLOT IMPULSE RESPONSE

$$h[n] = b_0(a_1)^n u[n] = 3(0.8)^n u[n]$$



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