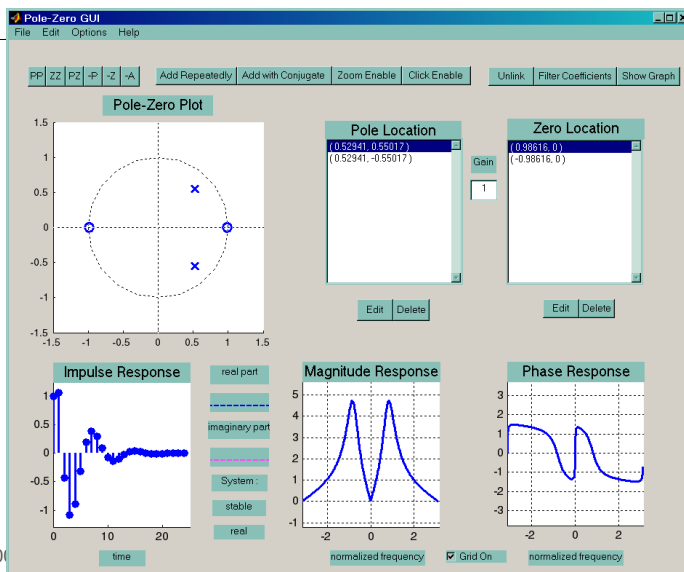


**Lecture 17**  
**H(z) & Frequency Response**  
**for IIR Systems**  
**21-Mar-03**

**Info: Web-CT, Lab, HW**

- Quiz #3 is 11-April (Friday)
- Prob Set #10 is due next week
- Lab #9 is DTMF (Touch-Tone) Lab
- Quiz #2: Resolve any grade changes by Friday (28-Mar)

**PeZ Demo: Pole-Zero Placing**



**ZZZZZ-Transform**



teaching the 'Z-TRANSFORM'...

**LECTURE**

## READING ASSIGNMENTS

- This Lecture:
  - Chapter 8, Sects. 8-4 8-5 & 8-6
- Other Reading:
  - Recitation: Chapter 8, all
    - POLE-ZERO PLOTS

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## LECTURE OBJECTIVES

- SYSTEM FUNCTION:  $H(z)$
- $H(z)$  has **POLES** and ZEROS
- FREQUENCY RESPONSE of IIR
  - Get  $H(z)$  first

$$H(e^{j\hat{\omega}}) = H(z) \Big|_{z=e^{j\hat{\omega}}}$$

- THREE-DOMAIN APPROACH

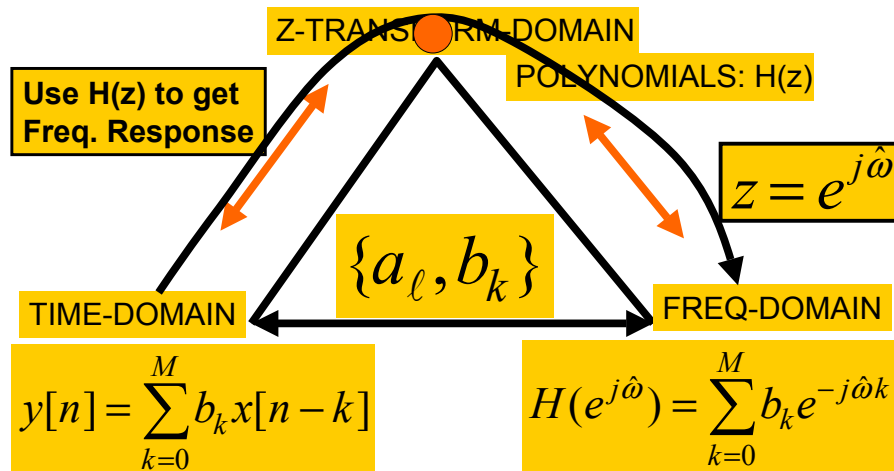
$$h[n] \leftrightarrow H(z) \leftrightarrow H(e^{j\hat{\omega}})$$

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## THREE DOMAINS



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## $H(z) = z$ -Transform{ $h[n]$ }

- FIRST-ORDER IIR FILTER:

$$y[n] = a_1 y[n-1] + b_0 x[n]$$

$$h[n] = b_0 (a_1)^n u[n]$$

$$H(z) = \frac{b_0}{1 - a_1 z^{-1}}$$

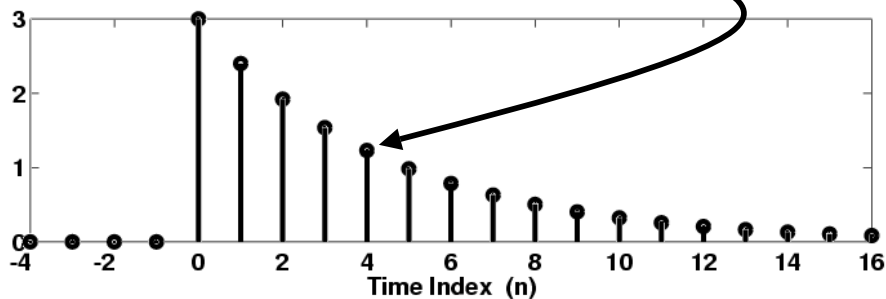
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## PLOT IMPULSE RESPONSE

$$h[n] = b_0(a_1)^n u[n] = 3(0.8)^n u[n]$$



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## First-Order Transform Pair

$$h[n] = ba^n u[n] \leftrightarrow H(z) = \frac{b}{1 - az^{-1}}$$

- GEOMETRIC SEQUENCE:

$$\begin{aligned} H(z) &= b_0 \sum_{n=0}^{\infty} a_1^n z^{-n} = b_0 \sum_{n=0}^{\infty} (a_1 z^{-1})^n \\ &= \frac{b_0}{1 - a_1 z^{-1}} \quad \text{if } |z| > |a_1| \end{aligned}$$

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## DELAY PROPERTY of X(z)

- DELAY in TIME  $\leftrightarrow$  Multiply X(z) by  $z^{-1}$

$$x[n] \leftrightarrow X(z)$$

$$x[n-1] \leftrightarrow z^{-1} X(z)$$

Proof:

$$\begin{aligned} \sum_{n=-\infty}^{\infty} x[n-1] z^{-n} &= \sum_{\ell=-\infty}^{\infty} x[\ell] z^{-(\ell+1)} \\ &= z^{-1} \sum_{\ell=-\infty}^{\infty} x[\ell] z^{-\ell} = z^{-1} X(z) \end{aligned}$$

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## Z-Transform of IIR Filter

- DERIVE the SYSTEM FUNCTION H(z)
  - Use **DELAY PROPERTY**

$$y[n] = a_1 y[n-1] + b_0 x[n] + b_1 x[n-1]$$

$$Y(z) = a_1 z^{-1} Y(z) + b_0 X(z) + b_1 z^{-1} X(z)$$

**EASIER with DELAY PROPERTY**

Time delay of  $n_0$  samples multiplies the z-transform by  $z^{-n_0}$

$$x[n - n_0] \leftrightarrow z^{-n_0} X(z)$$

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## SYSTEM FUNCTION of IIR

- NOTE the FILTER COEFFICIENTS

$$Y(z) - a_1 z^{-1} Y(z) = b_0 X(z) + b_1 z^{-1} X(z)$$

$$(1 - a_1 z^{-1}) Y(z) = (b_0 + b_1 z^{-1}) X(z)$$

$$H(z) = \frac{Y(z)}{X(z)} = \frac{b_0 + b_1 z^{-1}}{1 - a_1 z^{-1}} = \frac{B(z)}{A(z)}$$

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## SYSTEM FUNCTION

- DIFFERENCE EQUATION:

$$y[n] = 0.8y[n-1] + 3x[n] - 2x[n-1]$$

- READ** the FILTER COEFFS:

**H(z)**

$$Y(z) = \left( \frac{3 - 2z^{-1}}{1 - 0.8z^{-1}} \right) X(z)$$

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## CONVOLUTION PROPERTY

- MULTIPLICATION** of z-TRANSFORMS

$$X(z) \rightarrow \boxed{H(z)} \rightarrow Y(z) = H(z)X(z)$$

- CONVOLUTION** in TIME-DOMAIN

$$x[n] \rightarrow \boxed{h[n]} \rightarrow y[n] = h[n] * x[n]$$

**IMPULSE RESPONSE**

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## POLES & ZEROS

- ROOTS of Numerator & Denominator

$$H(z) = \frac{b_0 + b_1 z^{-1}}{1 - a_1 z^{-1}} \rightarrow H(z) = \frac{b_0 z + b_1}{z - a_1}$$

$$b_0 z + b_1 = 0 \Rightarrow z = -\frac{b_1}{b_0}$$

**ZERO:**  
H(z)=0

$$z - a_1 = 0 \Rightarrow z = a_1$$

**POLE:** H(z) → inf

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## EXAMPLE: Poles & Zeros

- VALUE of  $H(z)$  at POLES is **INFINITE**

$$H(z) = \frac{2 + 2z^{-1}}{1 - 0.8z^{-1}}$$

$$H(-1) = \frac{2 + 2(-1)}{1 - 0.8(-1)} = 0$$

ZERO at  $z = -1$

$$H\left(\frac{4}{5}\right) = \frac{2 + 2\left(\frac{4}{5}\right)}{1 - 0.8\left(\frac{4}{5}\right)} = \frac{\frac{7}{5}}{0} \rightarrow \infty$$

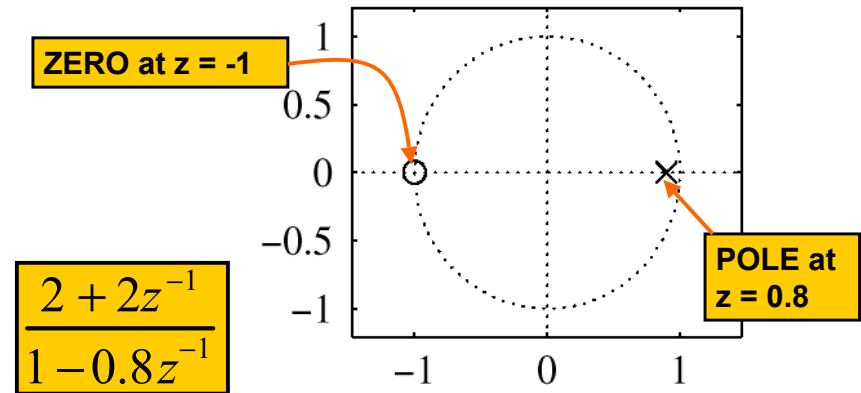
POLE at  $z = 0.8$

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## POLE-ZERO PLOT



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## FREQUENCY RESPONSE

- SYSTEM FUNCTION:  $H(z)$
- $H(z)$  has **DENOMINATOR**
- FREQUENCY RESPONSE of IIR
  - We have  $H(z)$

$$H(e^{j\hat{\omega}}) = H(z) \Big|_{z=e^{j\hat{\omega}}}$$

- THREE-DOMAIN APPROACH

$$h[n] \leftrightarrow H(z) \leftrightarrow H(e^{j\hat{\omega}})$$

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## FREQUENCY RESPONSE

- EVALUATE on the UNIT CIRCLE

$$H(e^{j\hat{\omega}}) = H(z) \Big|_{z=e^{j\hat{\omega}}}$$

$$H(z) = \frac{b_0 + b_1z^{-1}}{1 - a_1z^{-1}}$$

$$H(e^{j\hat{\omega}}) = H(z) \Big|_{z=e^{j\hat{\omega}}} = \frac{b_0 + b_1e^{-j\hat{\omega}}}{1 - a_1e^{-j\hat{\omega}}}$$

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# FREQ. RESPONSE FORMULA

$$H(z) = \frac{1}{1 - 0.8z^{-1}} \rightarrow H(e^{j\hat{\omega}}) = \frac{1}{1 - 0.8e^{-j\hat{\omega}}}$$

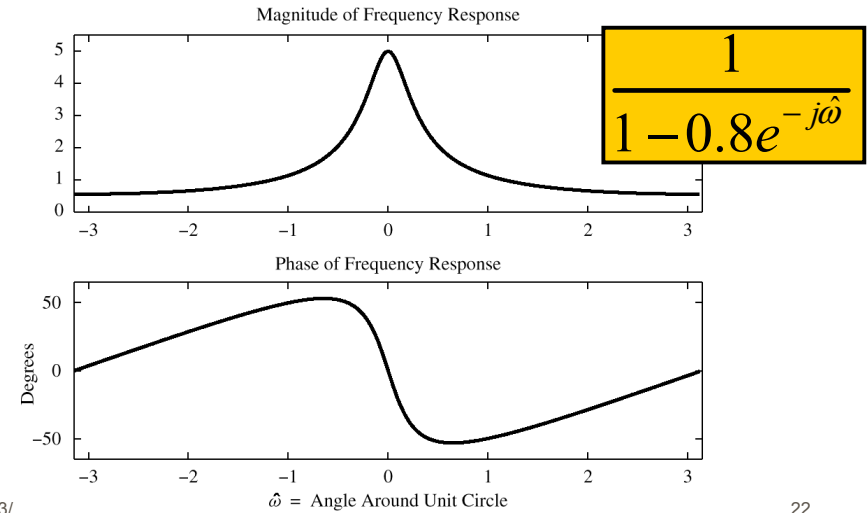
$$|H(e^{j\hat{\omega}})|^2 = \left| \frac{1}{1 - 0.8e^{-j\hat{\omega}}} \right|^2 = \frac{1}{1 - 0.8e^{-j\hat{\omega}}} \cdot \frac{1}{1 - 0.8e^{j\hat{\omega}}}$$

$$\frac{1}{1 + 0.64 - 0.8e^{-j\hat{\omega}} - 0.8e^{j\hat{\omega}}} = \frac{1}{1.64 - 1.6\cos\hat{\omega}}$$

$$@\hat{\omega} = 0, |H(e^{j\hat{\omega}})|^2 = \frac{1}{0.04} = 25 \quad @\hat{\omega} = \pi?$$

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# FREQ. RESPONSE from H(z)

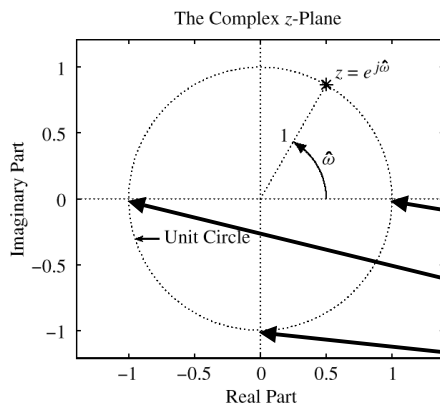


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# UNIT CIRCLE

## MAPPING BETWEEN $z$ and $\hat{\omega}$



$$z = e^{j\hat{\omega}}$$

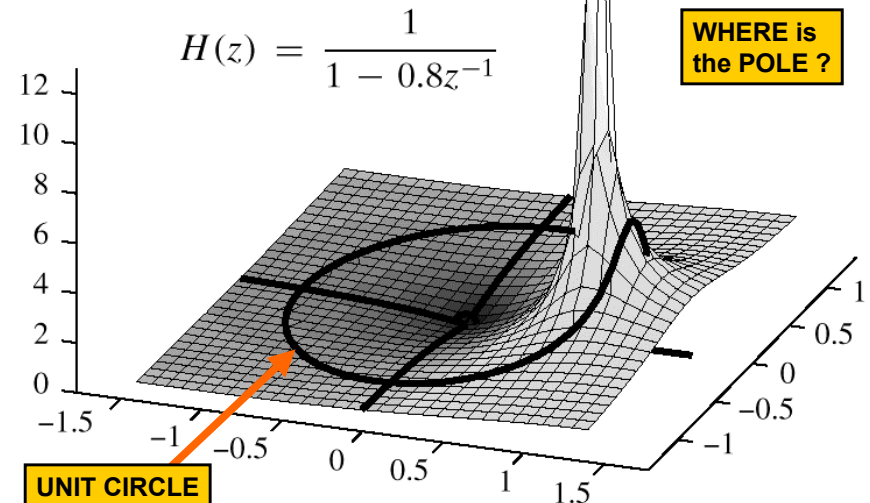
$$\begin{aligned} z = 1 &\leftrightarrow \hat{\omega} = 0 \\ z = -1 &\leftrightarrow \hat{\omega} = \pm\pi \\ z = \pm j &\leftrightarrow \hat{\omega} = \pm\frac{1}{2}\pi \end{aligned}$$

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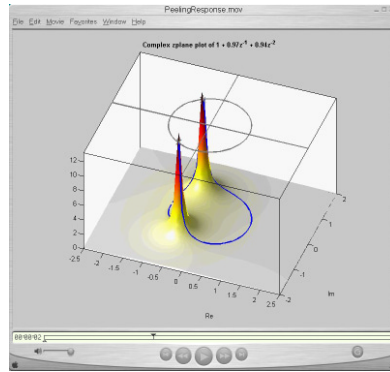
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## 3-D VIEWPOINT: EVALUTE H(z) EVERYWHERE



# MOVIE for H(z) in 3-D

- POLES to H(z) to Frequency Reponse
  - TWO POLES SHOWN

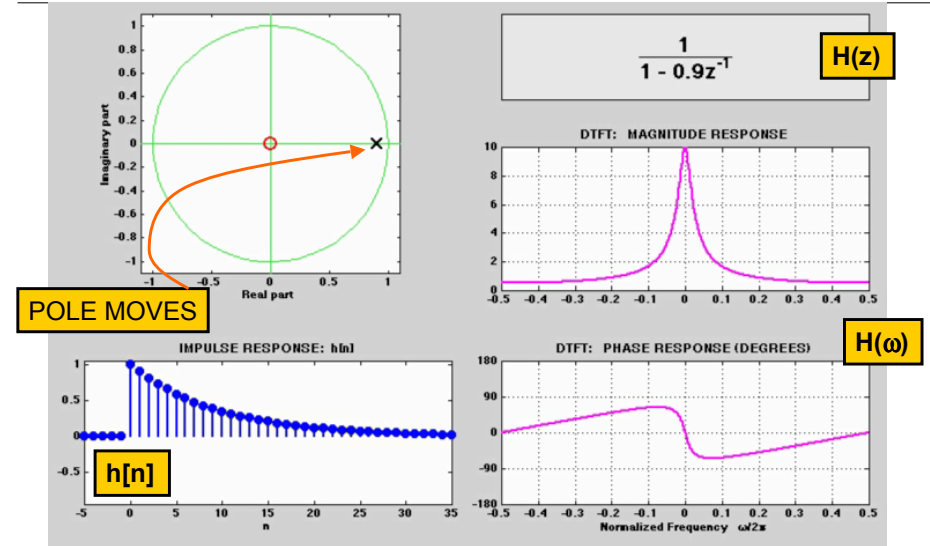


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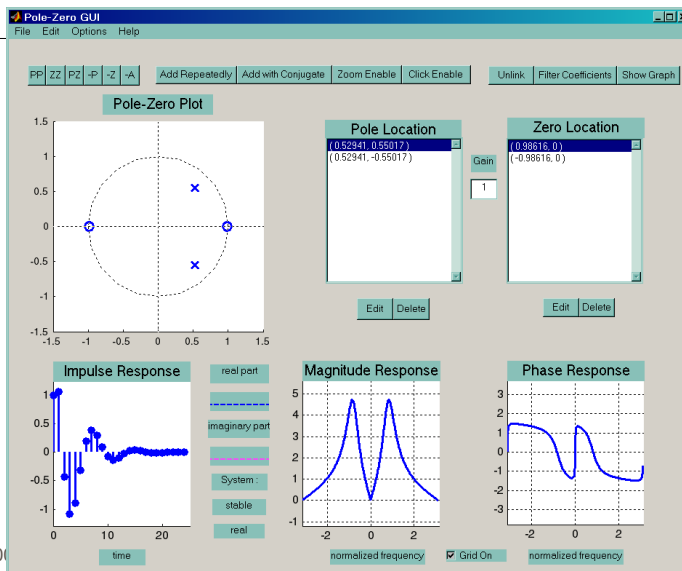
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# 3 DOMAINS MOVIE: IIR



# PeZ Demo: Pole-Zero Placing



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