

Lecture 17
 Frequency Response, $H(z)$,
 Poles and Zeros
 for IIR and FIR Systems
 26-Oct-09

Info: Lab, HW, etc.

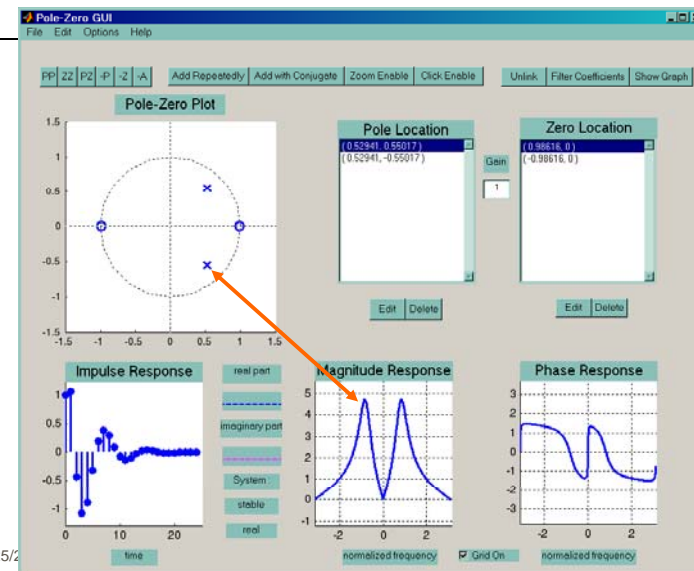
- Labs:
 - Lab #8: both partners do the hearing test.
 - Lab #9 starts on Tues, 27-Oct
 - Lab #10 will be Formal Report
- Homework:
 - #9 due NEXT week
- Quiz #3 will be 23-Nov (Monday)

Z-TRANSFORM TABLES

SHORT TABLE OF z -TRANSFORMS

	$x[n]$	\iff	$X(z)$
1.	$ax_1[n] + bx_2[n]$	\iff	$aX_1(z) + bX_2(z)$
2.	$x[n - n_0]$	\iff	$z^{-n_0} X(z)$
3.	$y[n] = x[n] * h[n]$	\iff	$Y(z) = H(z)X(z)$
4.	$\delta[n]$	\iff	1
5.	$\delta[n - n_0]$	\iff	z^{-n_0}
6.	$a^n u[n]$	\iff	$\frac{1}{1 - az^{-1}}$

PeZ Demo: Pole-Zero Placing



Lecture

READING ASSIGNMENTS

- This Lecture:
 - Chapter 7, Sections 7-6 to end
 - Chapter 8, Sections 8-4, 8-5, 8-6, 8-9, and 8-10
- Other Reading:
 - Recitation & Lab: same as above
 - ZEROS (and POLES)

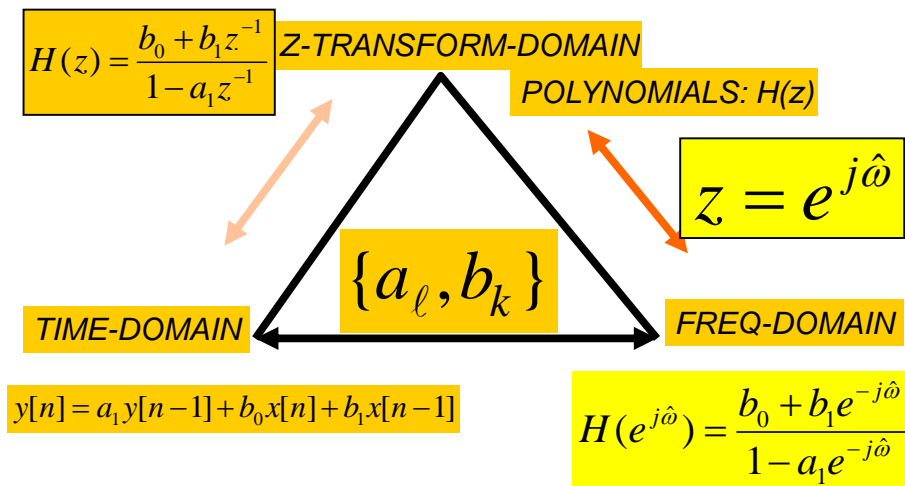
LECTURE OBJECTIVES

- ZEROS and POLES
- Relate $H(z)$ to FREQUENCY RESPONSE

$$H(e^{j\hat{\omega}}) = H(z) \Big|_{z=e^{j\hat{\omega}}}$$

- Four demos: PeZ, movies
 - Placing Poles and Zeros
- Bandpass Filters: IIR
- Nulling Filters: FIR

THREE DOMAINS: $H(e^{j\hat{\omega}})$



$H(z)$ = Rational Function

- First Order:

$$H(z) = \frac{b_0 + b_1 z^{-1}}{1 - a_1 z^{-1}}$$
- We can also study Second-Order Systems:

$$H(z) = \frac{b_0 + b_1 z^{-1} + b_2 z^{-2}}{1 - a_1 z^{-1} - a_2 z^{-2}} = \frac{B(z)}{A(z)}$$

- Numerator & Denominator Polynomials

POLES & ZEROS of H(z)

- Zeros of H(z), i.e., where is H(z)=0?

- Look for Roots of Numerator Polynomial

$$H(z) = \frac{B(z)}{A(z)}, \text{ so } B(z_0) = 0 \Rightarrow H(z_0) = 0$$

if $A(z_0) \neq 0$

- Poles of H(z), i.e., where is H(z)=∞?

- Look for Roots of Denominator Polynomial

$$H(z) = \frac{B(z)}{A(z)}, \text{ so } A(z_0) = 0 \Rightarrow H(z_0) \rightarrow \infty$$

if $B(z_0) \neq 0$

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Poles/Zeros of 1st-order H(z)

- Roots of Numerator & Denominator Polys:

$$H(z) = \frac{1 + b_1 z^{-1}}{1 - 0.8 z^{-1}}$$

$$H(z) = \frac{z(1 + b_1 z^{-1})}{z(1 - 0.8 z^{-1})} = \frac{z + b_1}{z - 0.8}$$

Pole at : $z = 0.8$

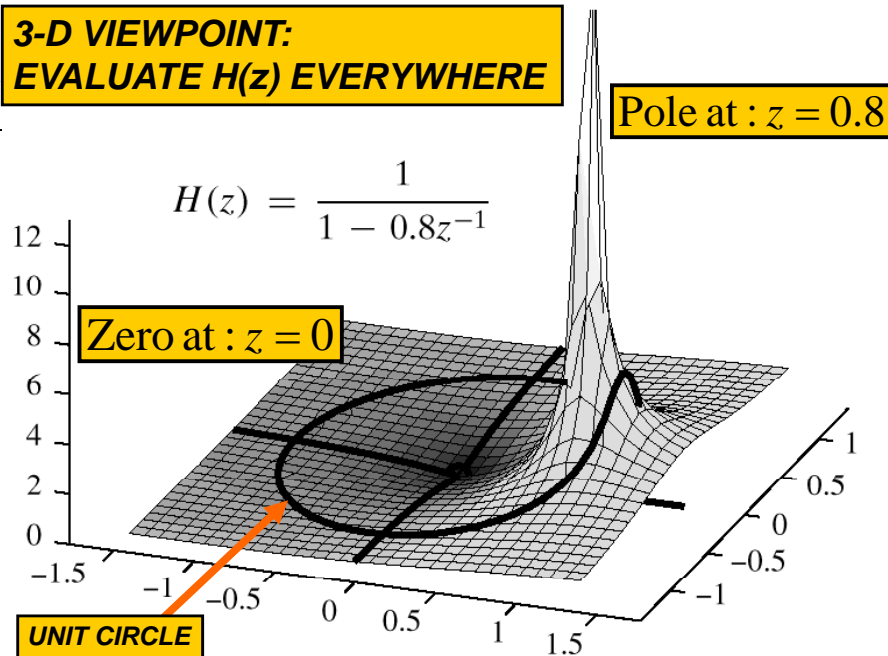
Zero at : $z = -b_1$

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**3-D VIEWPOINT:
EVALUATE H(z) EVERYWHERE**



FREQ. RESPONSE from H(z)

$$H(e^{j\hat{\omega}}) = H(z) \Big|_{z=e^{j\hat{\omega}}}$$

- Relate H(z) to FREQUENCY RESPONSE
- EVALUATE H(z) on the **UNIT CIRCLE**
 - ANGLE is same as FREQUENCY

$z = e^{j\hat{\omega}}$ (as $\hat{\omega}$ varies)
defines a **CIRCLE**, radius = 1

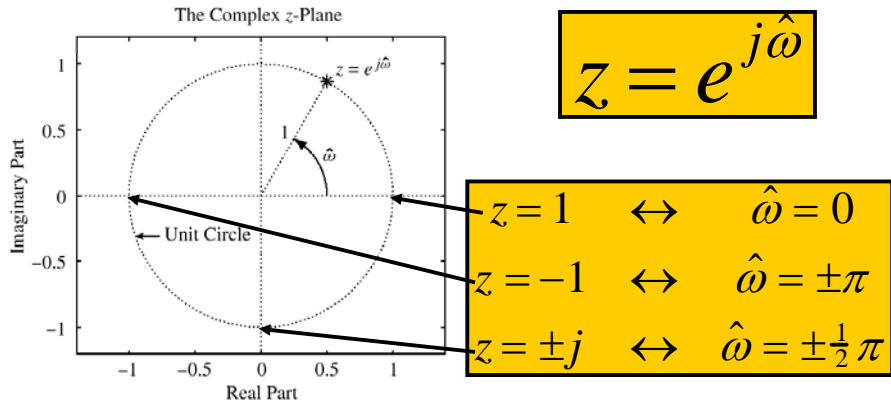
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UNIT CIRCLE: RECAP

- MAPPING BETWEEN z and $\hat{\omega}$



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IIR $H(z)$ example: two poles

- Poles just inside the unit circle (for stability)

$$H(z) = \frac{1}{1 + 0.97z^{-1} + 0.9409z^{-2}}$$

2 Poles: $z = 0.97e^{\pm j2\pi/3}$

2 Zeros: $z = 0, 0$

- MATLAB: `roots()` and `poly()`
 - `roots([1, 0.97, 0.9409])`
 - `poly(0.97*exp(j*2*pi*[1,-1]/3))`

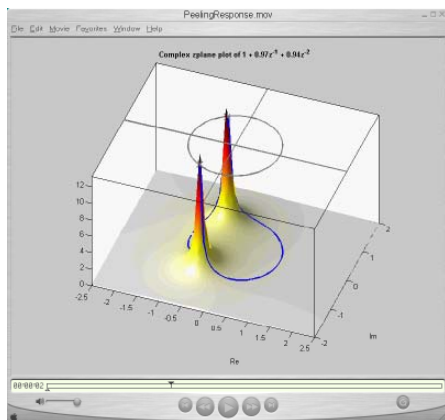
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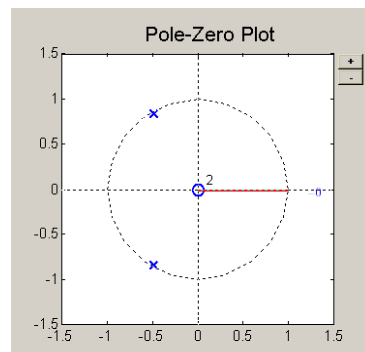
MOVIE for $H(z)$ in 3-D

- POLES to $H(z)$ to Frequency Reponse
 - TWO POLES SHOWN



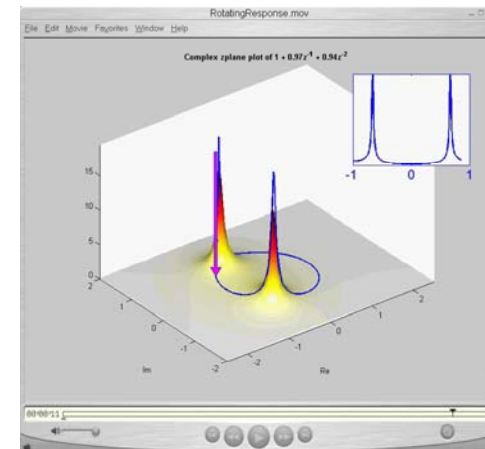
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Frequency Response from $H(z)$

Walking around the Unit Circle



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FIR H(z) has "ONLY ZEROS"

- FIR H(z) has NO DENOMINATOR, e.g.

$$H(z) = 1 - 2z^{-1} + 2z^{-2} - z^{-3}$$

$$H(z) = (1 - z^{-1})(1 - z^{-1} + z^{-2}) = \frac{(z-1)(z^2 - z + 1)}{z^3}$$

$$3 \text{ Zeros: } z = 1, \frac{1}{2} \pm j\frac{\sqrt{3}}{2} \quad e^{\pm j\pi/3}$$

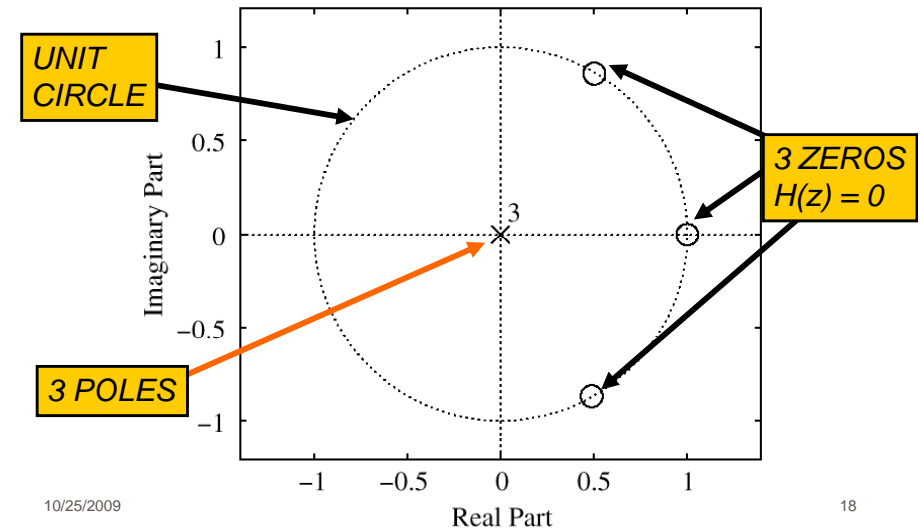
$$3 \text{ Poles: } z = 0, 0, 0$$

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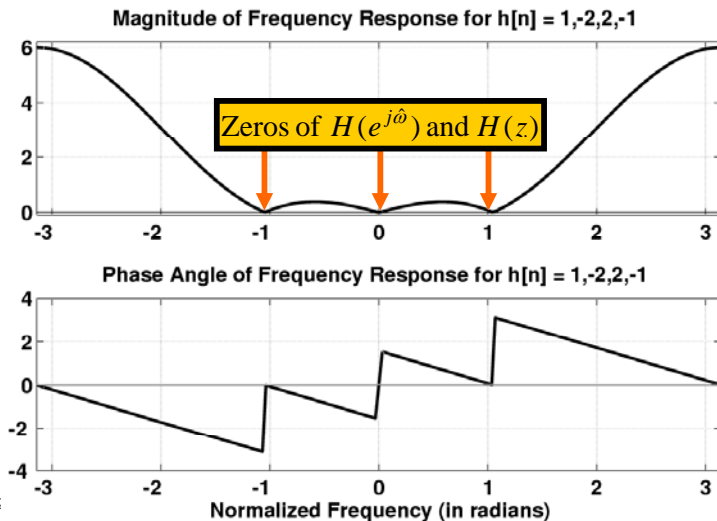
Pole-Zero Plot in z-PLANE



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FIR Frequency Response



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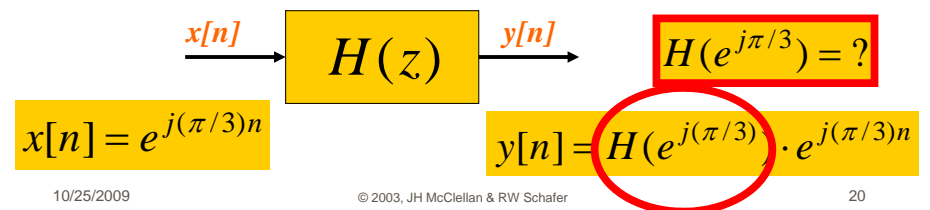
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NULLING PROPERTY of H(z)

- When $H(z)=0$ on the unit circle.
 - Find various inputs $x[n]$ that give zero output

$$H(z) = 1 - 2z^{-1} + 2z^{-2} - z^{-3}$$

$$H(e^{j\hat{\omega}}) = 1 - 2e^{-j\hat{\omega}} + 2e^{-j2\hat{\omega}} - e^{-j3\hat{\omega}}$$

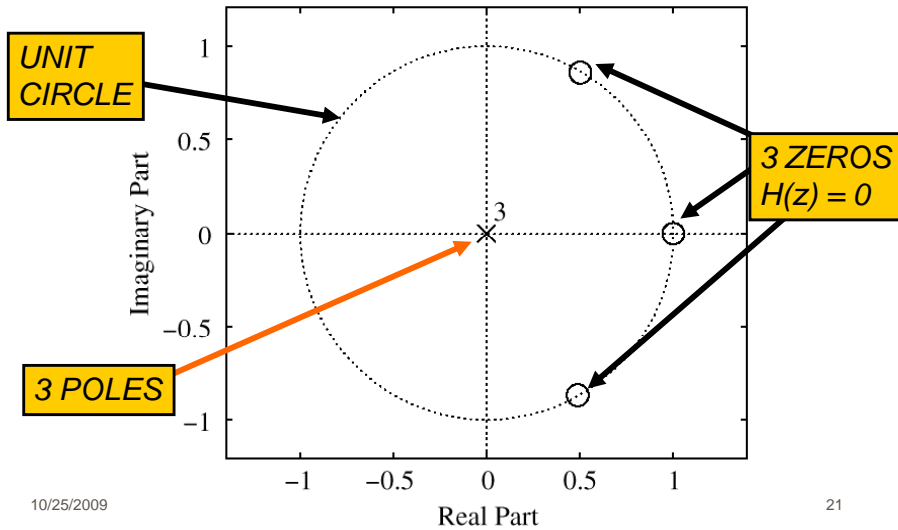


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PLOT ZEROS in z-DOMAIN



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NULLING PROPERTY of H(z)

- Evaluate H(z) at the input "frequency"

$$H(e^{j\hat{\omega}}) = 1 - 2e^{-j\hat{\omega}} + 2e^{-j2\hat{\omega}} - e^{-j3\hat{\omega}}$$

$$y[n] = H(e^{j\pi/3}) \cdot e^{j(\pi/3)n}$$

$$y[n] = (1 - 2e^{-j\pi/3} + 2e^{-j2\pi/3} - e^{-j3\pi/3}) \cdot e^{j(\pi/3)n}$$

$$(1 - 2(\frac{1}{2} - j\frac{\sqrt{3}}{2}) + 2(-\frac{1}{2} - j\frac{\sqrt{3}}{2}) - (-1))$$

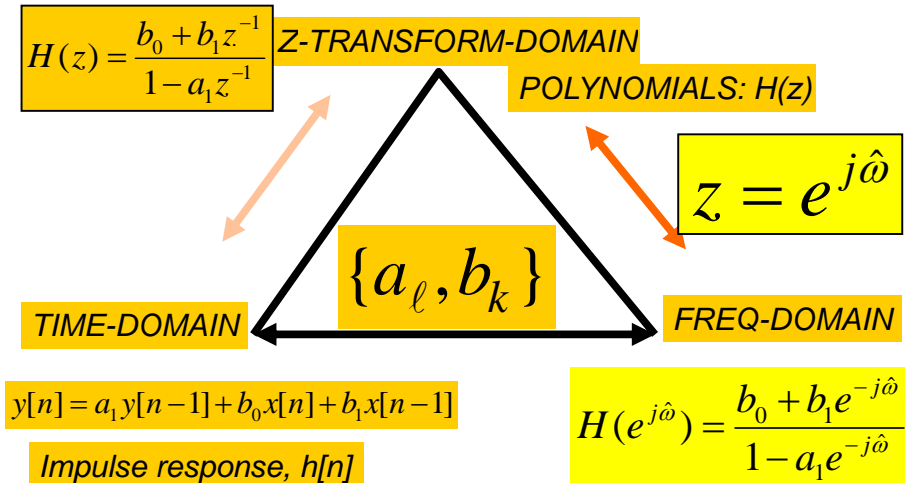
$$y[n] = (1 - 1 + j\sqrt{3} - 1 - j\sqrt{3} + 1) \cdot e^{j(\pi/3)n} = 0$$

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THREE DOMAINS: $H(e^{j\hat{\omega}})$

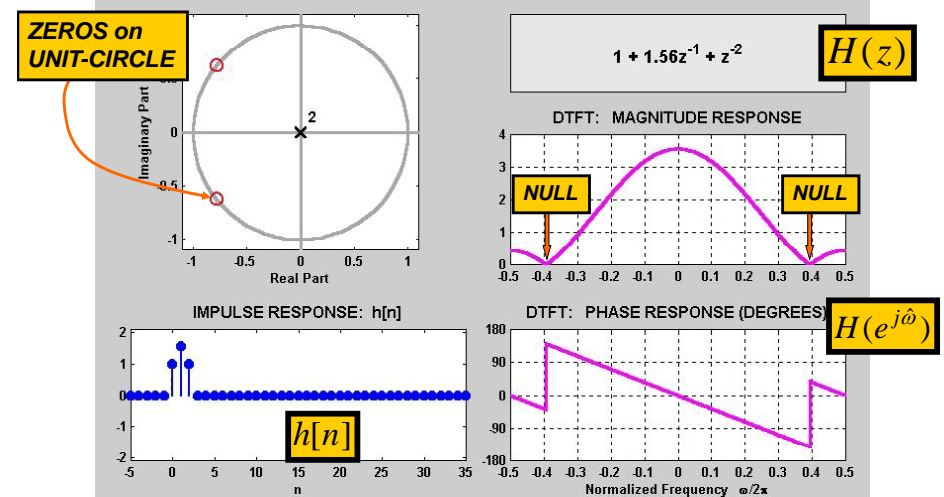


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3 DOMAINS MOVIE: FIR

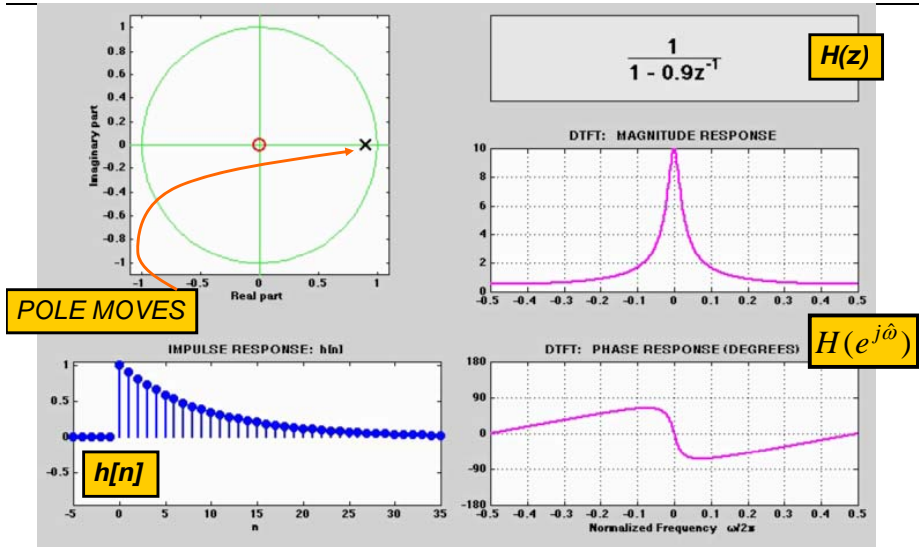


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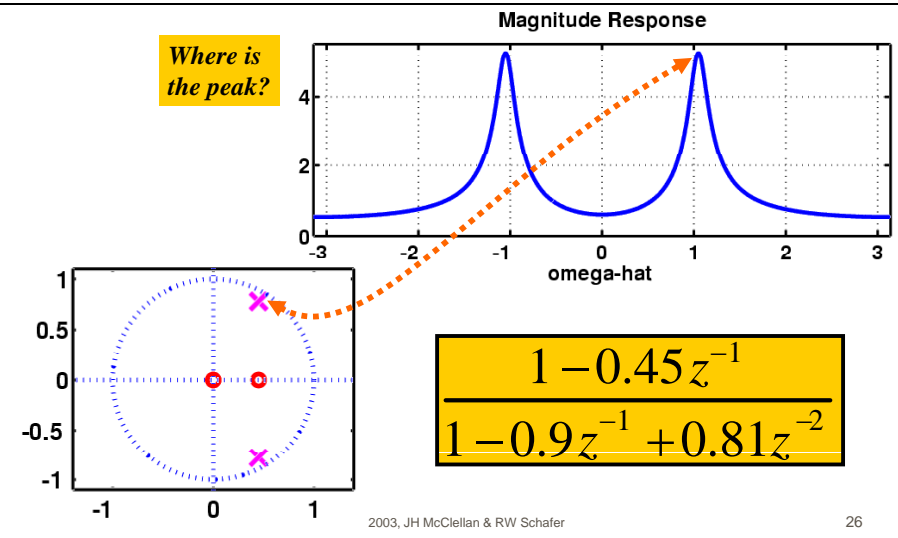
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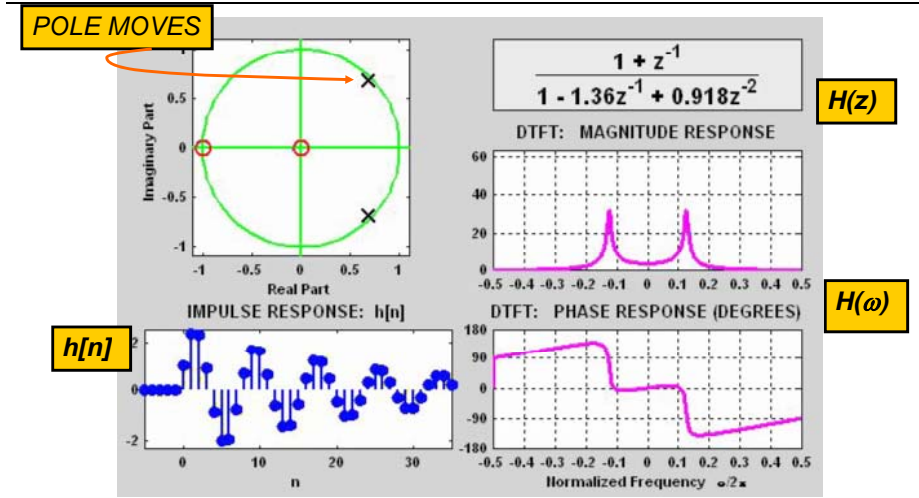
3 DOMAINS MOVIE: IIR



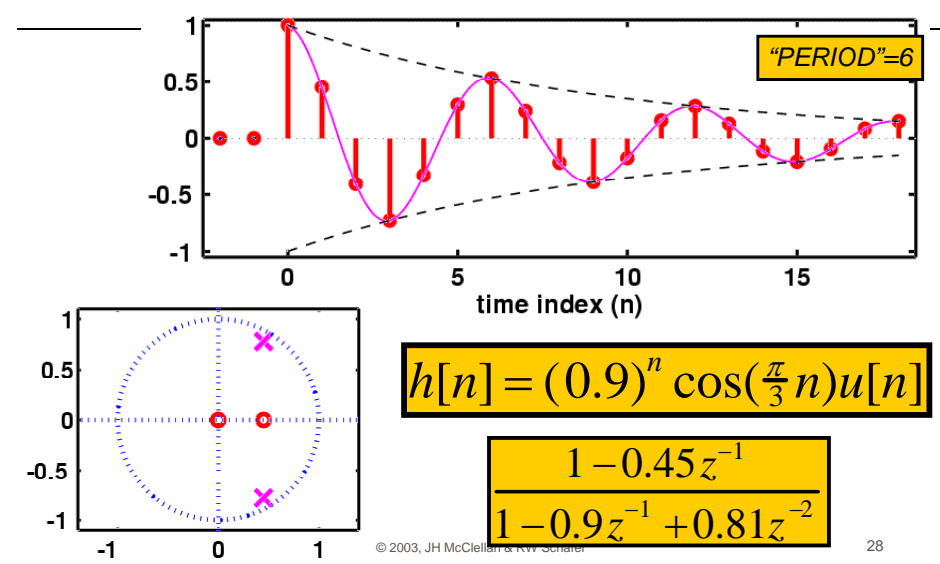
Complex POLE-ZERO PLOT



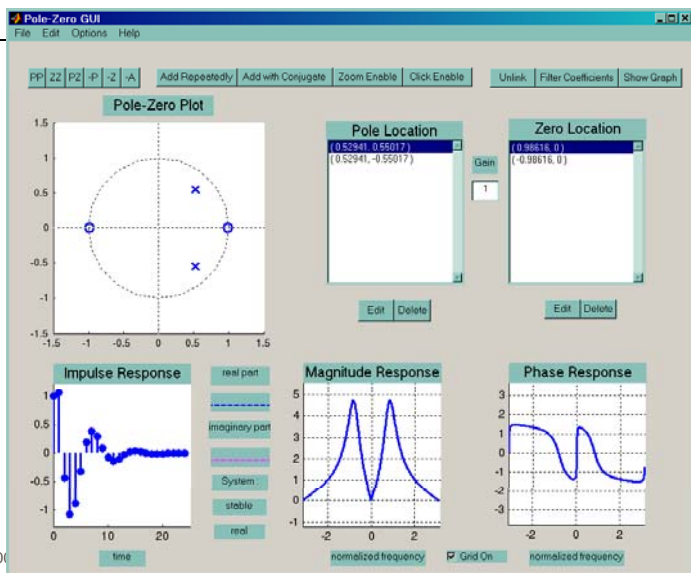
3 DOMAINS MOVIE: IIR



$h[n]$: Decays & Oscillates



PeZ Demo: Pole-Zero Placing



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DESIGN PROBLEM

- Example:
 - Design a Lowpass FIR filter (Find b_k)
 - Reject completely 0.7π , 0.8π , and 0.9π
 - This is NULLING
 - Estimate the filter length needed to accomplish this task. How many b_k ?
- Z POLYNOMIALS provide the TOOLS

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NULLING FILTER

- PLACE ZEROS to make $y[n] = 0$

$$H(z) = b_0 + b_1z^{-1} + b_2z^{-2} + b_3z^{-3} + b_4z^{-4} + b_5z^{-5} + b_6z^{-6}$$

Needs 6 ZEROS
where $H(z) = 0$

$$H(z_k) = 0, \text{ for } z_k = e^{\pm j0.7\pi}, e^{\pm j0.8\pi}, e^{\pm j0.9\pi}$$

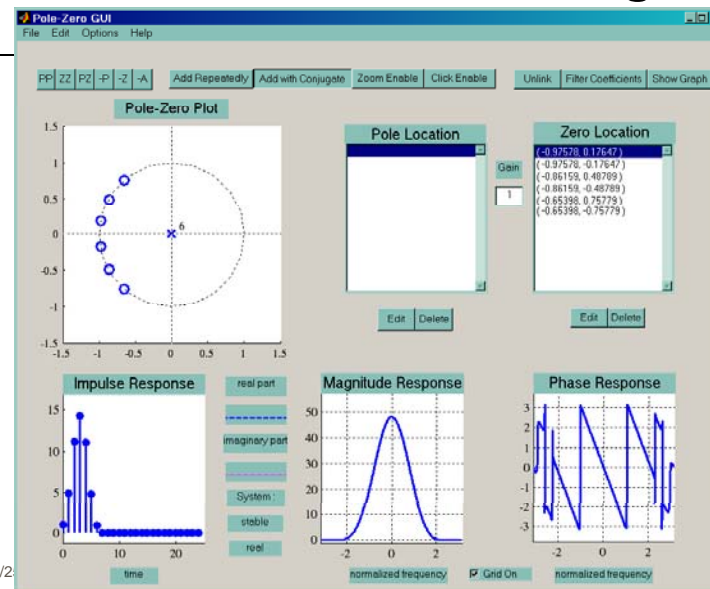
$$x[n] = e^{j0.8\pi n} \Rightarrow y[n] = H(e^{j0.8\pi})e^{j0.8\pi n}$$

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PeZ Demo: Zero Placing



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