Modeling and Analysis for Failure Amplification Method

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ABSTRACT

The failure amplification method (FAMe) is an experimental technique that uses a special type of factor known as amplification factor to amplify the failure probability so as to maximize the information in the experiment. A general strategy for model building is proposed by utilizing the information in the amplification factor. Generalized linear models (GLM) for binary response provide the flexibility of choosing proper links. The best design settings for improving the process capability are determined through carefully selected GLMs and loss functions. Two experiments for improving the quality of printed circuit boards are used to illustrate the proposed strategy.

KEY WORDS: Design of Experiments; Generalized Linear Models; Process Capability; Robust Parameter Design; Variation Reduction.

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Introduction

Designed experiments are widely used for reducing failure or defect rate in manufacturing processes. The experiment is carried out by changing the factor settings according to the levels specified by the investigator and observing the number of failures. When the probability of failures is small, it can happen that very few failures occur in the experiment. With such an outcome, it is difficult or even impossible to build an adequate model and obtain optimum process settings. To overcome this difficulty, Joseph and Wu (2004) proposed a novel experimentation strategy known as failure amplification method (FAMe). In FAMe, an amplification factor is selected based on the physical knowledge of the process and is used to amplify the failures. The experiment is then performed at the amplified conditions to ensure that an adequate number of failures are observed, which will provide sufficient information for modeling, analysis, and optimization. It is assumed that if the process is improved at the amplified conditions, then it is also improved at the normal conditions. One natural question would be: if the failure rate in the normal conditions is small, then why should we even perform an experiment? The answer is to improve the process capability. If the process capability is not improved, then even with a slight shift in the process conditions (due to some special causes) the failure rate can shoot up. This is pictorially depicted in Figure 1. Thus it is essential to reduce the failure rate as much as we can.

The models proposed in Joseph and Wu can be simplistic and may not be adequate to deal with more complex processes. In this article we propose new modeling strategies and provide a general framework for the analysis of experiments using FAMe. Since the experiment is performed at the amplified conditions, we need to *extrapolate* to the normal conditions for optimization. Because of this, carefully selected models should be used for the analysis. The performance of these models at the tails, i.e., regions with low failure rate, is critical for obtaining accurate results during extrapolation. Generalized linear models (GLMs) are suitable for the analysis of failure data (see McCullagh and Nelder (1989), Hamada and Nelder (1997)). Our proposed strategy builds appropriate nonlinear models but avoids the complexity of using generalized nonlinear models by utilizing the availability of GLMs in



Figure 1: Importance of using FAMe to improve process capability.

standard software. It is illustrated with the analysis of two real experiments on printed circuit boards (PCBs).

Motivating Examples

We use two examples to illustrate the proposed modeling strategy. The first is an experiment on the inner layer (IL) manufacturing process of PCBs, which was analyzed by Joseph and Wu (2004). It is included here as a comparison of our proposed methods with theirs. The second example is an experiment on the outer layer (OL) manufacturing process of PCBs. The new feature of this experiment is the presence of a noise factor.

In PCBs, circuits can be laid out in different layers. A double-sided (DS) PCB has two layers of circuits (top and bottom), whereas a multi-layer (ML) PCB has more than two layers of circuits (4, 6, 8, \cdots). To manufacture an ML PCB, first inner layers are manufactured and sandwiched between two copper layers using a pressing operation. Each inner layer has circuits on two sides. Thus, a 4-layer PCB contains one inner layer, a 6-layer PCB contains two inner layers, and so on. An overall view of the process is shown in Figure 2. Numerous defects are generated during the manufacturing of PCBs, of which shorts and opens in the circuits are the major ones. The first experiment was conducted to reduce shorts and opens in the inner layers and the second one in the outer layers.



Figure 2: Inner layer (IL) and outer layer (OL) processes in PCB manufacturing.

The rejection due to inner layer shorts and opens is of the order of 1-2%. If we conduct an experiment with such a low failure rate, then we may observe only a few defects in the experiment, which cannot be used for estimation of the model parameters. We can overcome the problem using a large number of PCBs for each experiment, but the cost will be prohibitive. Therefore, a cost-effective strategy is to run the experiment after amplifying the failures. It is known that the shorts will increase if the spacing between the conductors is reduced and opens will increase if the line width of the conductors is reduced. The industry was mainly engaged in the production of circuits with line width/spacing 5 mils (1 mil = 0.001 inch) or higher (up to 15 mils). Therefore to amplify the failures, the experiments were performed by creating a special circuit pattern with 3 and 4 mils. Joseph and Wu (2004) classified this type of amplification as complexity factor method, because the line width and spacing here are the complexity factors of the product.

Example 1: Eight factors were selected from the inner layer PCB process for experimentation. The factors and levels are shown in Table 1. An 18-run orthogonal array given in Table 2 was used for the experiment. The experiment was conducted by processing one inner layer for each run. A special inner layer circuit pattern was designed only for the purpose of the experiment. The inner layer contains conductors with 3, 4, 5, 6, and 7 mils of line width and spacing. See Maruthi and Joseph (1999) for details of the experiment. There were 80 pairs of conductors on each inner layer. Since a pair of conductors give rise to two opportunities for

Control factors	Notation		Levels	
		1	2	3
Preheat	x_1	No*	-	Yes
Surface preparation	x_2	$Scrub^*$	Pumice	Chemical
Lamination speed	x_3	1.2 mpm	1.5 mpm^*	$1.8 \mathrm{mpm}$
Lamination pressure	x_4	20 psi	40 psi^*	60 psi
Lamination temperature	x_5	$95~^{0}\mathrm{C}$	$105 \ ^{0}C^{*}$	$115 \ ^{0}C$
Exposure energy	$x_6(m)$	14	17^{*}	20
Developer speed	x_7	3 fpm	4 fpm^*	$5~{\rm fpm}$
ORP	x_8	500	530^{*}	560

Table 1: Factors and levels for the IL PCB experiment in example 1.

* Operating levels of the factors in production.

opens and one opportunity for shorts, there are a total of 160 opportunities for opens under each line width and 80 opportunities for shorts under each spacing. The data on shorts and opens from the experiment are shown in Table 2.

Example 2: About 5% of the PCBs have shorts in the outer layers and about 1% have opens. Most of the outer layer shorts can be reworked, whereas very few of the opens can be corrected. Thus in outer layers, an open is more serious than a short. Again eight factors were selected from the image transfer stage of the outer layer manufacturing process for experimentation. The factors and levels are shown in Table 3. The same 18-run orthogonal array in Table 2 was used for the experiment. The factors are assigned to the columns in the same order given in Table 3. Outer layer process can be used for DS boards or ML boards. Because the manufacturer had no control over the type of the board, it is treated as a noise factor, and was included in the experiment. There was one more important noise factor in the process. The solution in the developer was changed after processing several boards. Therefore, its performance towards the end was not as efficient as when it was new. When a cross array is used for experimentation, the number of runs will double if two noise factors are used. To reduce the run size, the method of noise factor compounding (Taguchi 1986) was employed. It is known that more shorts will occur with older developer bath. Moreover,

									Opens					Sł	nort	\mathbf{s}		
Run	x_1	x_2	x_3	x_4	x_5	x_6	x_7	x_8	3	4	5	6	7	3	4	5	6	7
1	1	1	1	1	1	1	1	1	33	7	4	0	1	1	0	0	0	0
2	1	1	2	2	2	2	2	2	7	9	1	0	0	4	1	0	0	0
3	1	1	3	3	3	3	3	3	14	3	1	0	0	19	2	0	0	0
4	1	2	1	1	2	2	3	3	2	0	2	0	0	9	0	0	0	0
5	1	2	2	2	3	3	1	1	7	1	2	1	0	22	1	1	1	0
6	1	2	3	3	1	1	2	2	78	30	7	1	1	8	0	0	0	0
7	1	3	1	2	1	3	2	3	9	1	3	0	0	19	1	0	0	0
8	1	3	2	3	2	1	3	1	7	0	1	0	1	4	0	1	0	0
9	1	3	3	1	3	2	1	2	4	3	0	0	0	7	0	0	0	0
10	2	1	1	3	3	2	2	1	6	0	0	0	0	22	1	0	0	1
11	2	1	2	1	1	3	3	2	13	2	0	0	0	34	2	2	0	0
12	2	1	3	2	2	1	1	3	34	5	0	1	3	13	4	1	0	0
13	2	2	1	2	3	1	3	2	8	3	0	0	0	7	0	1	0	0
14	2	2	2	3	1	2	1	3	25	8	0	2	1	25	1	0	0	0
15	2	2	3	1	2	3	2	1	7	0	0	0	0	41	1	0	0	1
16	2	3	1	3	2	3	1	2	10	6	0	0	0	45	9	5	0	1
17	2	3	2	1	3	1	2	3	8	0	0	0	0	3	0	0	0	0
18	2	3	3	2	1	2	3	1	12	2	0	0	1	7	2	0	0	0

Table 2: $OA(18, 2^1 \times 3^7)$ and data from the IL PCB experiment in example 1

ML boards have greater number of shorts than DS boards. Therefore the combination of ML boards with old bath will lead to more shorts than the combination of DS boards with fresh bath. Thus the two levels of the compounded noise factor are defined as N_1 : DS/Fresh bath and N_2 : ML/Old bath. As in the inner layer experiments, a special test pattern was designed for the outer layers. In the case of outer layer board, there were 88 pairs of conductors. Thus, there is a total of 88 opportunities for shorts and 176 opportunities for opens for each line width and spacing. The data are shown in Table 4. More details about this experiment can be found in Maruthi et al. (1998).

Factors	Notation		Levels	
		1	2	3
Vacuum delay time	x_1	$2 \mathrm{sec}^*$	-	$10 \sec$
Developer concentration	x_2	$.85 \ \%$	$1.0\%^{*}$	1.15~%
Preheat temperature	x_3	$150 \ ^{0}C^{*}$	$250~^0\mathrm{C}$	$400~^{0}{\rm C}$
Developer break point	x_4	30~%	$50\%^*$	70~%
Exposure step	x_5	13	16^{*}	19
Develop pressure	x_6	$1.7 \mathrm{\ bar}$	2.0 bar^*	2.3 bar
Lamination temperature	x_7	$90~^0\mathrm{C}$	$105 \ ^{0}C^{*}$	$115 \ ^{0}C$
Lamination speed	x_8	$1.5 \mathrm{m/min}$	1.8 m/min^*	$2.1 \mathrm{m/min}$
Noise factor	N	DS/Fresh bath	_	ML/Old bath

Table 3: Factors and levels for the OL PCB experiment in example 2.

* Operating levels of the factors in production.

		DS/Fresh Bath										ML/Old Bath								
		0	pens				Shorts				Opens				Shorts					
Run	3	4	5	6	7	3	4	5	6	7	3	4	5	6	7	3	4	5	6	7
1	1	0	0	0	0	30	12	3	1	2	1	1	0	0	0	42	30	6	3	7
2	15	1	0	0	0	6	5	1	0	1	1	1	0	1	1	38	12	5	3	0
3	43	3	1	0	0	2	4	2	0	0	32	2	0	0	1	18	8	6	1	1
4	2	1	0	0	0	16	8	1	0	2	1	0	0	0	1	27	13	5	3	2
5	114	4	0	0	0	2	2	0	0	0	6	1	0	0	0	17	15	9	3	3
6	4	0	0	0	0	23	8	0	0	1	1	1	1	1	0	47	20	5	3	2
7	4	0	0	0	0	32	13	4	3	4	0	1	0	0	0	41	22	5	2	4
8	20	1	0	0	0	10	1	2	1	0	3	1	0	0	0	21	10	2	1	1
9	112	4	0	0	0	16	13	1	0	0	17	1	0	0	0	43	30	7	6	1
10	115	13	1	0	0	3	1	1	1	0	12	2	0	0	0	27	9	0	3	1
11	1	0	0	0	0	18	5	2	0	5	1	0	0	0	0	46	20	9	4	5
12	23	11	12	10	9	45	15	9	2	2	2	3	2	1	1	62	37	26	16	8
13	68	2	0	0	0	8	5	1	0	0	3	1	0	0	0	18	10	2	5	1
14	5	1	0	0	0	22	4	0	0	0	1	1	0	0	1	22	19	15	7	1
15	3	0	0	0	0	33	9	4	1	1	2	2	0	0	0	42	23	11	6	5
16	7	1	0	0	0	29	7	1	1	0	5	1	0	1	1	40	27	9	4	3
17	73	1	0	0	0	13	1	3	1	0	6	2	1	1	2	40	26	7	5	3
18	1	0	0	0	0	28	12	4	4	2	1	0	1	0	0	37	16	8	5	3

Table 4: Data from the OL PCB experiment in example 2.

Model Building

Statistical-Physical Based Modeling Strategies

When the responses are recorded as failure or success and are independent under some control (**X**) and noise (**N**) factors, binomial model with failure probability $p(\mathbf{X}, \mathbf{N}, M)$ is a natural candidate for describing the outcome, where M denotes the amplification factor. A general model relating failure probability and the variables can be described by

$$f(p) = \lambda(\mathbf{X}, \mathbf{N}, M), \tag{1}$$

where f is the link function in the GLM and λ is an appropriate function of the variables. Now we propose a modeling strategy for choosing f and λ .

For binomial family, the typical link functions are complimentary log-log (cloglog), logit, and probit:

$$\log \log \frac{1}{1-p}, \ \log \frac{p}{1-p}, \ \text{and} \ \Phi^{-1}(p).$$
 (2)

The canonical link of binomial case is logit, which leads to logistic regression and is often used because the minimum sufficient statistics exist for regression parameters (Lindsey 1997). The complimentary log-log link has its own glamor through the connection with Poisson distribution. The probit link assumes the underlying random effect follows a normal distribution. However none of the above guarantees the best fitting for real data. One may use other reasonable cumulative distribution functions or Box-Cox transformations to obtain links for better fitting. To keep the modeling strategy simple, we stay with the popular link functions in (2). Other flexible links can be considered only when the model fit is not satisfactory. Another advantage of using these links is their availability in most statistical software.

The functional form of $\lambda(\mathbf{X}, \mathbf{N}, M)$ can be obtained by statistical modeling on the observed data and physical knowledge of the factors. The amplification factor is deliberately chosen because of its dominant effect on failure. The physical knowledge of the factor effects, especially on the amplification factor, will help to simplify the possible model forms. We consider the following simpler model:

$$f(p) = \lambda(\mathbf{X}, \mathbf{N}) + g(\mathbf{X}, \mathbf{N})h(M), \qquad (3)$$



Figure 3: The probability values of cloglog, logit and probit links. Left panel shows the middle portion and the right panel shows the tail part.

where $\lambda(\mathbf{X}, \mathbf{N})$ and $g(\mathbf{X}, \mathbf{N})$ are linear models in \mathbf{X} and \mathbf{N} . The function h(M) is a monotonic nonlinear function in M. The choice of h is very critical, particulary because we need to extrapolate outside the experimental range of M. Because the failure rate is likely to plateau in normal production levels of the amplification factor, simple polynomial functions will not be appropriate for h. Figure 4 shows the failure probability in Example 1 under the complimentary log-log link. Clearly there is a nonliner trend. Low order polynomial functions of amplification factor do not fit well with any of the three links. Three candidates,

$$\log M, \ (M-a)^{-b}, \ \text{and} \ \exp(-a(M-b)^{c})$$
 (4)

are thus considered for the h function, where a, b, and c are constants that will be determined through an iterative estimation procedure. More complicated functions can be considered but should only be used when the simpler ones fail.

Model Selection Procedure

There are three parts of the model that need to be chosen, namely, a link function, a nonlinear function h, and the significant control and noise factor effects in λ and g. Figure 3 shows the differences of the three link functions. For failure amplification problem, lower tail should be the primary concern. Interestingly the lower tail of cloglog and logit behave almost the same, whereas the probit link has smaller probabilities compared to the other two.

These properties indicate that these three links have different scales of model parameters and that the probit link will have more distinct estimates than the other two links. We use the Akaike information criterion (AIC) for model selection because of its ability to compare nested and non-nested models and its popularity in applications. Other criteria like BIC or C_p can also be used and the ensuing steps will be similar to that of using AIC.

Choosing/estimating functions f, λ, g , and h simultaneously is cumbersome and is not supported by most commercial statistical software. Therefore we provide an iterative strategy that foregoes the burden of programming and utilizes a standard GLM software (for example, the *glm* command in Splus and R).

Modeling Strategy:

- Choose a function h with parameters a and a link function f. Fit the model f(p) = h(M, a), which can be done by minimizing the objective function of AIC with respect to a. Denote the estimate by â.
- Use forward selection procedure (Neter et al., 1996) to select important factors and interactions based on the effect heredity principle (Hamada and Wu, 1992) to obtain λ and g in f(p) = λ(**X**, **N**) + g(**X**, **N**)h(M, â).
- Use the functions λ and g obtained in step 2 to minimize the objective function of AIC with respect to a.
- 4. Repeat steps 2 and 3 until the model functions λ , g, and the parameter estimate $\hat{\mathbf{a}}$ converge.
- 5. Repeat steps 1 through 4 for different choices of h and f. Select the functions which gives the lowest AIC.

Note that step 2 can be easily carried out using a standard GLM software, thus simplifying the estimation procedure. This step can be further simplified by using plots of data to select the important factors in g. Then the variable selection needs to be applied only for λ . We now illustrate this simplified strategy using the two PCB examples.



Figure 4: Example 1. Failure probability with a complimentary log-log transformation for opens. Numbers correspond to the average of factor levels. Lines are the fitted values by the model with cloglog link (9).

Example 1. IL PCB Experiment

The data was analyzed by Joseph and Wu (2004) assuming the relationship:

$$\log \log \frac{1}{1-p} = \lambda(\mathbf{X}_{(-6)}) + \gamma \log m + \alpha \log M,$$
(5)

where the exposure energy (x_6) is treated as an adjustment factor denoted by m. Note that this is a special case of (3) with f as the cloglog link, g equal to a constant α , and $h(M) = \log M$. The $\lambda(\mathbf{X}_{(-6)})$ is a second-order linear model (main effects and two-factor interactions) of the control factors, excluding x_6 . The two degrees of freedom of three-level factors are split into linear and quadratic components with contrasts $x_l = (-1, 0, 1)$ and $x_q = (1, -2, 1)$. The two-level factor x_1 is coded with $x_{1l} = (-1, 1)$.

Denote the two amplification factors, line width and spacing, by M_1 and M_2 . Note that M_1 is an amplification factor for opens and M_2 is an amplification factor for shorts. The following two models for opens (6) and shorts (7) were selected by forward selection based

Data	Link (Eq.#)	AIC	Size	Deviance
Opens	Cloglog-JW (6)	147.91	7	133.91
	Cloglog (9)	121.64	10	101.64
	Logit	122.62	11	100.62
	Probit	137.48	12	113.48
Shorts	Cloglog-JW (7)	101.08	6	89.08
	Cloglog	84.24	11	62.24
	Logit (10)	81.71	10	61.71
	Probit	85.53	12	61.53

Table 5: Analysis of example 1.

on AIC:

$$\log \log \frac{1}{1-p} = 10.27 - .73x_{5l} + 0.57x_{4l} - .33x_{2l} - .27x_{1l}x_{5q} + 2.77\log m + 5.06\log M_1, \quad (6)$$
$$\log \log \frac{1}{1-p} = -6.66 + .48x_{1l} + .20x_{4l} - .15x_{1l}x_{5q} + 4.70\log m + 7.66\log M_2. \quad (7)$$

Two concerns are raised here. Are the amplification factors which dominate the failure effect adequately modeled? How do other links fit the data?

Plots of data can shed light on the first question. The numbers in Figure 4 represent the effect of the amplification factor averaged over the other factors for the opens data. There is no indication of interaction between control and amplification factors. So the term g in (3) is set to be a constant. Same conclusion is obtained for the shorts data also. Thus, a simple model is chosen for both shorts and opens:

$$f(p) = \lambda(\mathbf{X}, M) = \lambda(\mathbf{X}) + \beta h(M).$$
(8)

For consistency of notation with the rest of the paper, the variable, exposure energy, is denoted by x_6 instead of m.

Table 5 gives the AIC, number of parameters and deviances of models with differnt link functions. For numerical stability, the parameters a, b, and c in the function $h(M) = \exp(-a(M-b)^c)$ are restricted to lie in the range [-100, 0], [0, 3] and [-20, 0] respectively. The model Clolog-JW is suggested by Joseph and Wu (2004). Other models are built by using the proposed strategy. Clearly models (6) and (7) are not good enough because of the high AIC (and hence high deviance). The proposed procedure provides better models in all three links. The cloglog link for opens and the logit link for shorts have the lowest AIC. There is not much difference in terms of AIC between cloglog and logit links. Thus, if the same link should be used for both opens and shorts data, one may pick either cloglog or logit.

The following model is selected for opens:

$$\log \log \frac{1}{1-p} = -2.63 - .82x_{5l} - .34x_{6l} + .28x_{8l} - .27x_{8q} + .15x_{6q} + .24x_{1l} + 1.06x_{5l}x_{6l} - .38x_{1l}x_{6l} - 5.27e^{-1.36(M_1-3)^{-1.16}}.$$
(9)

The order of the terms in the model is the order in which they are selected by the regression procedure. The fitted values are plotted in Figure 4, which shows a reasonably good fit to the data. Comparing models (6) and (9), the surface preparation (x_2) is selected in model (6) but not in model (9). Model (9) also selects ORP (x_8) as an important factor with linear and quadratic effects which were absent in model (6).

The following model is selected for shorts:

$$\log \frac{p}{1-p} = -1.86 + .96x_{6l} + .47x_{1l} - .18x_{7l} + .29x_{6q} + .25x_{4l} - .24x_{1l}x_{7l} - .45x_{4l}x_{6l} - .31x_{1l}x_{4l} - .715.18e^{-5.52(M_2-3)^{-.09}}.$$
 (10)

Comparing models (7) and (10), developer speed (x_7) is selected in (10) but not in (7). Moreover, the interaction terms are completely different. This indicates that the proposed procedure provides more information about the factor effects which was missing in Joseph and Wu (2004).

Example 2. OL PCB Experiment

The plot of opens data shown in Figure 5 suggests that after the logit transformation only factors x_5 and N have interaction with the amplification factor. The cloglog and probit transformation plots, which are not shown here, lead to similar results. Thus, the function



Figure 5: Example 2. Failure probability with a logit transformation for opens. Numbers correspond to the average of factor levels. Lines are the fitted values by the logit link model including the noise factor.

g in (3) is chosen to be $\beta_1 X_5 + \beta_2 N + \beta_3$. Then the model for opens becomes

$$f(p) = \lambda_1(\mathbf{X}, N) + (\beta_1 X_5 + \beta_2 N + \beta_3) h_1(M_1).$$

Similar analysis done on shorts data shows that the amplification factor has no interaction with the control factors but has strong interaction with noise factor. Hence the function gis chosen to be $\beta_1 N + \beta_2$. Then the model for shorts becomes

$$f(p) = \lambda_2(\mathbf{X}, N) + (\beta_1 N + \beta_2)h_2(M_2).$$

The following cloglog and logit models ((11) and (12)) are selected for opens and shorts, respectively:

$$\log \log \frac{1}{1-p} = -3.71 + 2.19x_{5l} - .74N - .48x_{7l} - .32x_{6l} + .55x_{4l} + .14x_{3l} - .07x_{4q} + .25x_{7l}N + .26x_{6l}N - .43x_{5l}N + .21x_{3l}N - .29x_{4l}x_{6l} + .22x_{3l}x_{5l} + .15Nx_{4q} + .21x_{5l}x_{4q} + .06x_{7l}x_{4q} + (-2.09x_{5l} + .66N - 2.56)e^{-39.54(M_1 - 1.85)^{-5.69}},$$
(11)

Data	Link (Eq. $\#$)	AIC	Size	Deviance
Opens	Cloglog(11)	266.33	20	220.33
	Logit	288.28	19	250.28
	Probit	311.18	22	267.18
Shorts	Cloglog	243.34	16	211.34
	Logit (12)	238.29	17	204.29
	Probit	249.61	17	215.61

Table 6: Analysis of example 2

$$\log \frac{p}{1-p} = -.92 + .54N - .47x_{5l} - .27x_{7l} - .28x_{4l} + .19x_{3l} + .17x_{3q} + .19x_{1l} + .30x_{8l} -.16x_{7l}x_{3q} + .19x_{5l}N - .10Nx_{3q} + .09x_{1l}x_{3q} - .08x_{5l}x_{3q} - 07x_{7l}N + (.31N - 3.77)e^{-100.00(M_2 - .59)^{-3.58}}.$$
(12)

For comparison, we repeated the analysis using the models recommended in Joseph and Wu (2004). In their models a cloglog link was used for both opens and shorts with $h(M) = \log M$ and g equal to a constant. The resulting AIC was 486.7 for opens and 272.9 for shorts, both much larger than the AIC of our models (see Table 6). This clearly shows the superiority of the proposed modeling strategy.

Optimum Settings

Optimizing on shorts and opens separately may lead to conflicting levels for the factors. One approach to overcome this problem is to use a loss function to combine the two responses. We propose two loss functions:

$$c_1\lambda_1(\mathbf{X}, N, M_1) + c_2\lambda_2(\mathbf{X}, N, M_2) \tag{13}$$

and

$$c_1 p_1(\mathbf{X}, N, M_1) + c_2 p_2(\mathbf{X}, N, M_2) = c_1 f_1^{-1}(\lambda_1(\mathbf{X}, N, M_1)) + c_2 f_2^{-1}(\lambda_2(\mathbf{X}, N, M_2)).$$
(14)

The first one is a weighted sum of link functions and the second one is a weighted sum of probabilities. The former is appropriate if the same link function is used for both models, whereas the latter is appropriate if the link functions are different. The weights c_1 and c_2 can be chosen depending on the importance of the two responses. It is easier to determine the weights in (14), because the weights are directly related to the costs of rejection and rework. In the case of inner layers (example 1), both shorts and opens are equally bad, and thus we take $c_1 = c_2 = .5$. In the case of outer layers (example 2), most of the shorts can be reworked whereas very few of the opens can be reworked. Therefore PCBs with opens will be rejected leading to greater loss. Engineers suggested that the loss due to five shorts can be considered as equivalent to the loss with one open. Therefore, we take $c_1 = 5/6$ and $c_2 = 1/6$.

Optimum settings of the control factors can be obtained by averaging the loss over the distribution of the noise factors and amplification factors. Because the link functions for shorts and opens in (9) and (10) are different, we use the loss function based on probabilities. Table 7 gives the optimum settings for Example 1 by taking expectation over the amplification factors in the range 5 to 7 mils assuming a uniform distribution. Note that for optimization we adopt only the levels used in production; amplification is applied only to facilitate estimation. In the table, we have also shown the optimum levels obtained by Joseph and Wu (2004). They are clearly different from the levels obtained from our models. Because our models have better fit to the data than theirs, the optimum levels obtained using our models are expected to be closer to the true optimum.

The optimum settings for Example 2 based on the models in (11) and (12) are shown in Table 8. It is obtained by averaging over the distribution of noise factor and the amplification factors, and using the loss function in (14). Again, because the models selected based on the proposed modeling strategy are better than those selected using the strategy in Joseph and Wu (2004), the optimum settings obtained here should be better.

Figures 6 and 7 show the estimated failure probabilities. We can see that the failure probabilities at the optimum settings are substantially lower than that at the existing settings. This indicates that the process capability has been greatly improved through the

Table 7: Example 1: Optimum settings.

Model	x1	x2	x3	x4	x5	x6	x7	x8
Cloglog-JW	-1	1	х	-1	0.342	-0.399	х	х
Cloglog + Logit	-1	х	х	-1	1	-1	-1	-1

Table 8: Example 2: Optimum settings.

Model	x1	x2	x3	x4	$\mathbf{x5}$	x6	$\mathbf{x7}$	x8
Cloglog + Logit	-1	х	-1	.325	1	1	1	-1



Figure 6: Example 1, failure probabilities. 1=production setting, 2=optimum setting. Left panel gives cloglog model for opens. Right panel gives logit model for shorts.



Figure 7: Example 2, failure probabilities. 1=production setting, 2=optimum setting. Left panel gives cloglog model for opens. Right panel gives logit model for shorts.

experimentation.

Conclusion

FAMe points the engineers and statisticians to an avenue for resolving estimation problem in situations with low failure rate. In this article we propose a general strategy for building models with amplification factors. The proposed strategy is illustrated using two real experiments in PCB manufacturing.

Model selection is done by combining the physical knowledge of the process and the information from the experimental data. The procedure takes advantage of widely available GLM software and tunes it to find a nonlinear function of amplification factor through a model selection criterion, such as the AIC. This strategy can be implemented easily without heavy burden of programming.

Two types of loss functions, link and probability, are discussed. Link loss function possesses additive property of factor effects in models from two failures modes. It is easier to measure the total impact of the factors on the failure rate and thus easier to optimize. On the other hand, probability loss function provides the flexibility of combining models with different link functions.

Accurate failure prediction is essential for proper production planning and control. In PCB manufacturing, the failure prediction is difficult because the design of PCB (line width,

spacing, etc.) varies with customer requirements (see Joseph and Adya 2001). The models obtained from the experiment, which are functions of line width and spacing, can be used for failure prediction. However, because there are various other failure modes, these models need to be calibrated using actual production data for better prediction.

Acknowledgements

Dr. Jeng's research was partly supported by Taiwan National Science Council grant NSC 43140F. Dr. Joseph's research was supported by US National Science Foundation CAREER grant DMI-0448774. Dr. Wu's research was supported in part by a grant from the U.S. Army Research Laboratory and the U.S. Army Research Office under contract number W911NF-05-1-0264.

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