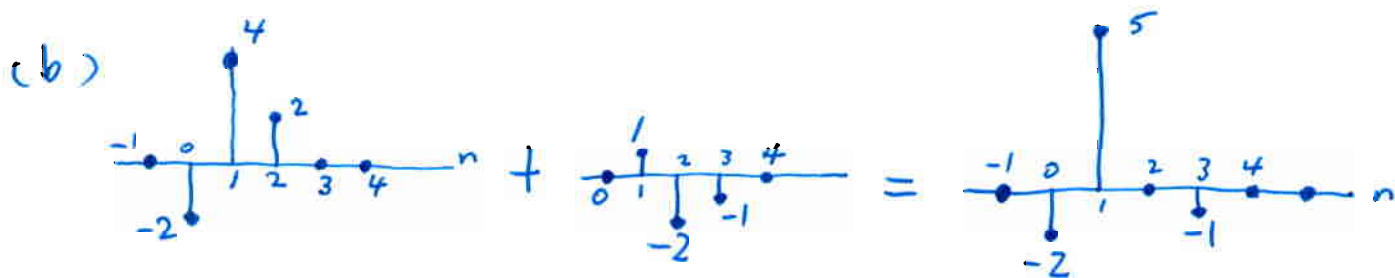
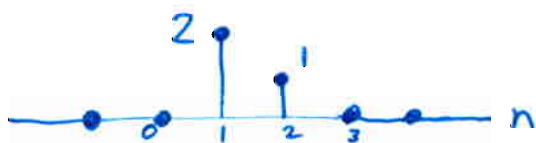
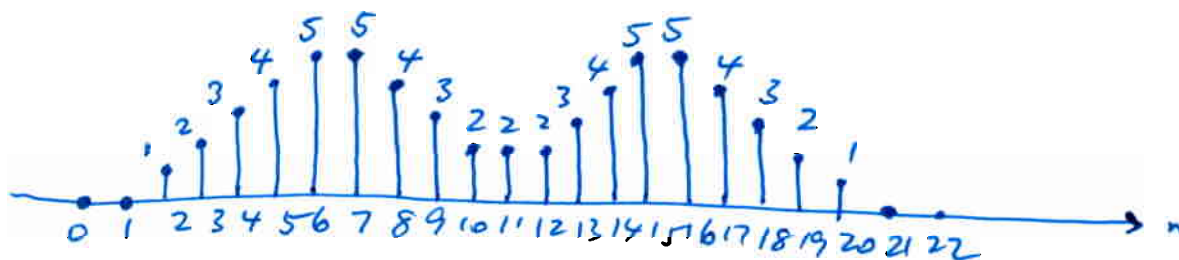


2.22

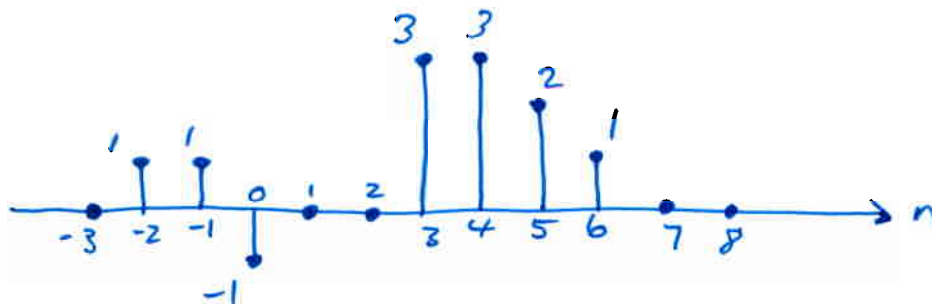
(a)  $y[n] = \delta[n-1] * h[n] = h[n-1]$



(c)



(d)



2.25

The output signal will be defined in 5 different regions:

- For  $n < 0$ :

$$y[n] = \sum_{k=-\infty}^0 x[k] \cdot h[n-k]$$

Since  $x[n] = 0$  for  $n < 0$ , the output will be equal to 0

$$y[n] = 0$$

- For  $0 \leq n \leq N_1$ :

$$y[n] = \sum_{k=0}^n a^k . \text{ Making use of the geometric series rule, we obtain}$$

the following result

$$y[n] = \frac{1 - a^{(n+1)}}{1 - a}$$

- For  $N_1 < n < N_2$ :

$$y[n] = \sum_{k=0}^{N_1} a^k$$

$$y[n] = \frac{1 - a^{(N_1+1)}}{1 - a}$$

- For  $N_2 \leq n \leq N_2 + N_1$

$$y[n] = \frac{1 - a^{(N_1+1)}}{1 - a} + \sum_{k=N_2}^n a^{(k-N_2)}$$

$$y[n] = \frac{1 - a^{(N_1+1)}}{1 - a} + a^{-N_2} \cdot \frac{a^{N_2} - a^{(n+1)}}{1 - a}$$

$$y[n] = \frac{1 - a^{(N_1+1)}}{1 - a} + \frac{1 - a^{(n-N_2+1)}}{1 - a}$$

- For  $N_2 + N_1 < n$

$$y[n] = \frac{1 - a^{(N_1+1)}}{1 - a} + \sum_{k=N_2}^{N_2+N_1} a^{(k-N_2)}$$

$$y[n] = \frac{1 - a^{(N_1+1)}}{1 - a} + a^{-N_2} \cdot \frac{a^{N_2} - a^{(N_2+N_1+1)}}{1 - a}$$

$$y[n] = \frac{1 - a^{(N_1+1)}}{1 - a} + \frac{1 - a^{(N_1+1)}}{1 - a}$$

$$y[n] = 2 \cdot \frac{1 - a^{(N_1+1)}}{1 - a}$$

## PROBLEM

2.30

For each of the following systems, determine whether the system is 1) stable, 2) causal, 3.) linear, and 4.) time invariant.

a.)  $T(x[n]) = \cos(\pi n) x[n]$

b.)  $T(x[n]) = x[n^2]$

c.)  $T(x[n]) = x[n] \sum_{k=0}^{\infty} \delta(n-k)$

d.)  $T(x[n]) = \sum_{k=n-1}^{\infty} x[k]$

The process of determining stability & causality is simplified if the system is linear & time invariant, (LTI). For example, a general system is stable if every bounded input produces a bounded output. If the system is LTI, the test for stability simplifies, such that

$$\sum_{k=-\infty}^{\infty} |h(k)| < \infty$$

is the qualifying criteria. The test for causality simplifies from checking that

$$y(n_0) \text{ depends only on } x(n), n \leq n_0, \forall n_0,$$

to checking that

$$h(n) = 0, n < 0,$$

if the system is LTI.

To take advantage of these properties, linearity and time invariance are determined first.

Linear systems are defined by superposition. If the system is linear, then

$$a_1 y_1[n] + a_2 y_2[n] = T(a_1 x_1[n] + a_2 x_2[n]),$$

where  $x_1[n]$  and  $x_2[n]$  are input functions that produce  $y_1[n]$  and  $y_2[n]$ . Note that  $a_1$  and  $a_2$  are constants.

Systems are considered time invariant if

$$y(n) = T(x(n)),$$

and

$$y(n-n_0) = T(x[n-n_0]).$$

This is represented by the following figure.

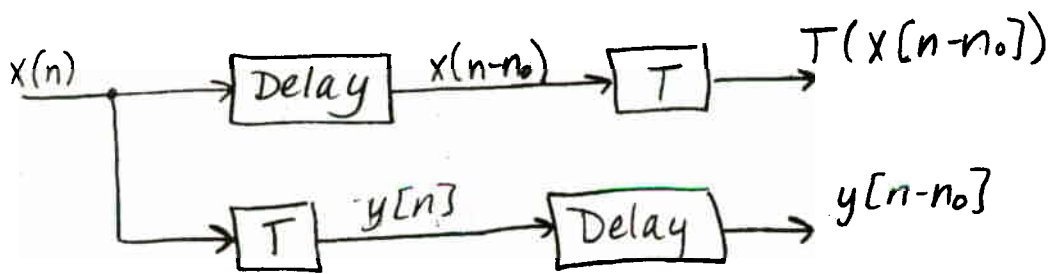


Figure 1: Flow Chart for Time Invariance

Note that  $T(x(n-n_0))$  is created when  $x(n)$  is delayed, and then transformed, while  $y(n-n_0)$  is created when  $x(n)$  is transformed and then shifted.

a.)  $T(x[n]) = \cos(\pi n) x[n]$

LINEARITY:

$y(n) = T(x(n))$

$$\begin{aligned}
T(a_1 x_1(n) + a_2 x_2(n)) &= [a_1 x_1(n) + a_2 x_2(n)] \cos(\pi n) \\
&= a_1 x_1(n) \cos(\pi n) + a_2 x_2(n) \cos(\pi n) \\
&= a_1 T(x_1(n)) + a_2 T(x_2(n)) \\
&= a_1 y_1(n) + a_2 y_2(n)
\end{aligned}$$

Since  $a_1 y_1(n) + a_2 y_2(n) = T(a_1 x_1(n) + a_2 x_2(n))$ , the system is linear.

TIME INVARIANCE:

$T(x(n))$  simply multiplies  $x(n)$  by  $\cos(\pi n)$ . The multiplier,  $\cos(\pi n)$ , is shown in Figure 2.

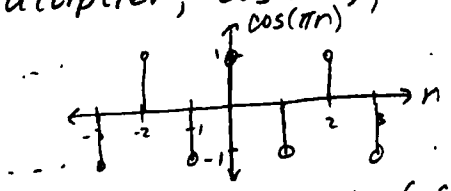


Figure 2: Plot of  $\cos(\pi n)$

Thus,  $\cos(\pi n) = \begin{cases} 1, & n \text{ is even} \\ -1, & n \text{ is odd} \end{cases}$ .

This means that

$T(x(n)) = \begin{cases} x(n), & n \text{ even} \\ -x(n), & n \text{ odd} \end{cases}$ .

It follows that

$T(x(n-m)) = \begin{cases} x(n-m), & n-m = \text{even} \\ -x(n-m), & n-m = \text{odd} \end{cases}$ .

Using the method described in Figure 1,

$y(n) = \begin{cases} x(n), & n \text{ even} \\ -x(n), & n \text{ odd} \end{cases}$ ,

and

$$y(n-n_0) = \begin{cases} x(n-n_0), & n-n_0 = \text{even} \\ -x(n-n_0), & n-n_0 = \text{odd} \end{cases}$$

Since  $y(n-n_0) = T(x(n-n_0))$ , the system is time invariant. This system is LTI, which is intuitive since  $T$  performs amplitude modulation on  $x(n)$ .

### STABILITY:

Since the system is LTI, it is stable if

$$\sum_{k=-\infty}^{\infty} |h(k)| < \infty$$

The impulse response is found when  $x(n) = \delta(n)$ . Thus,

$$h(n) = \begin{cases} \delta(n), & n \text{ even} \\ -\delta(n), & n \text{ odd} \end{cases}$$

but since  $\delta(n)$  is only non-zero at  $n=0$ , it simplifies to

$$h(n) = \delta(n)$$

The system is stable since

$$\sum_{k=-\infty}^{\infty} \delta(k) < \infty$$

### CAUSAL:

Since the system is LTI, it is causal if

$$h(n) = 0, \quad n < 0$$

The impulse response,  $h(n)$ , was determined to be

$$h(n) = \delta(n),$$

which only has a non-zero value at  $n=0$ .

Thus  $h(n) = 0$  for  $n < 0$

holds, & the system is causal.

b.)  $T(x(n)) = x(n^2)$

LINEARITY:

$$y(n) = T(x(n))$$

$$T(a_1 x_1(n) + a_2 x_2(n)) = a_1 x_1(n^2) + a_2 x_2(n^2) \\ = a_1 y_1(n) + a_2 y_2(n)$$

Since  $a_1 y_1(n) + a_2 y_2(n) = T(a_1 x_1(n) + a_2 x_2(n))$ , the system is Linear.

TIME INVARIANCE:

$$y(n) = T(x(n))$$

Using the method described in Figure 1,

$$T(x(n-n_0)) = x((n-n_0)^2) = x(n^2 - 2nn_0 + n_0^2)$$

while

$$y(n-n_0) = x(n^2 - n_0)$$

Since

$$y(n-n_0) \neq T(x(n-n_0)),$$

the system is NOT time invariant. Thus, the system is NOT LTI.

This answer is intuitive, since  $T$  impacts  $n$ , rather than  $x(n)$ .

STABILITY:

The system is stable if bounded inputs, denoted by

$$|x(n)| \leq B_x < \infty, \forall n,$$

produce bounded outputs, i.e.

$$|y(n)| \leq B_y < \infty, \forall n.$$

Since the transform  $T$  only acts on  $n$ , not  $x(n)$ , all values of  $y(n)$  are also values of  $x(n)$ . If all  $x(n)$  are bounded, it follows that  $y(n)$  is bounded for all  $n$ . Thus, it is stable.



CAUSAL:

The system is causal if  $y(n)$  only depends on  $x(n)$ ,  $n \leq n_0$ . A simple example shows that this system is not causal.

$$\text{EX: } y(n) = x(n^2)$$

$$n = 2$$

$$y(2) = x(4)$$

Thus,  $y(n_0)$  depends on future values of  $x(n)$ . The system is NOT causal.

$$c. T(x(n)) = x(n) \sum_{k=0}^{\infty} \delta(n-k)$$

P2.7

The equation for  $T(x(n))$  can be simplified.

$$\sum_{k=0}^{\infty} \delta(n-k) \Rightarrow \begin{array}{c} \sum_{k=0}^{\infty} \delta(n-k) \\ \uparrow \\ \text{---} \circ \text{---} \rightarrow k \end{array}$$

The plot of  $\sum_{k=0}^{\infty} \delta(n-k)$  reveals that it's equivalent to  $u(n)$ . Thus  $T(x(n)) = x(n)$ ,  $n \geq 0$ , or  $x(n)u(n)$ .

### LINEARITY:

$$y(n) = T(x(n))$$

$$\begin{aligned} T(a_1 x_1(n) + a_2 x_2(n)) &= a_1 x_1(n) + a_2 x_2(n), \quad n \geq 0. \\ &= a_1 y_1(n) + a_2 y_2(n), \quad n \geq 0. \end{aligned}$$

Since  $T(a_1 x_1(n) + a_2 x_2(n)) = a_1 y_1(n) + a_2 y_2(n)$ , the system is Linear.

### TIME INVARIANCE:

$$y(n) = T(x(n))$$

Using the method described in Figure 1,

$$T(x(n-n_0)) = \begin{cases} x(n-n_0), & n-n_0 \geq 0 \\ 0, & n-n_0 < 0 \end{cases}$$

which simplifies to

$$T(x(n-n_0)) = \begin{cases} x(n-n_0), & n \geq n_0 \\ 0, & n < n_0 \end{cases}$$

Similarly,

$$y(n) = \begin{cases} x(n), & n \geq 0 \\ 0, & n < 0 \end{cases}$$

and

$$y(n-n_0) = \begin{cases} x(n-n_0), & n \geq n_0 \\ 0, & n < n_0 \end{cases}$$

Pg. 8

Since  $y(n-n_0) = T(x(n-n_0))$ , the system is time invariant. The system is LTI. This is intuitive, since  $T(x(n))$  simply samples  $x(n)$  for all  $n \geq 0$ .

### STABILITY:

Since the system is LTI, stability is determined with the test

$$\sum_{k=-\infty}^{\infty} |h(k)| < \infty.$$

In this case,

$$h(n) = \delta(n) \sum_{k=-\infty}^{\infty} \delta(n-k) = \delta(n) u(n) = \delta(n).$$

Testing  $h(n)$  reveals that,

$$\sum_{k=-\infty}^{\infty} |\delta(k)| < \infty.$$

Thus, the system is stable.

### CAUSAL:

The impulse response,  $h(n)$ , was found to equal  $\delta(n)$ . Since  $\delta(n) = 0$  for  $n < 0$ , the system is CAUSAL.

$$d.) T(x(n)) = \sum_{k=-\infty}^{\infty} x(k)$$

LINEARITY:

$$y(n) = T(x(n))$$

$$\begin{aligned} T(a_1 x_1(n) + a_2 x_2(n)) &= \sum_{k=-\infty}^{\infty} (a_1 x_1(k) + a_2 x_2(k)) \\ &= \sum_{k=-\infty}^{\infty} a_1 x_1(k) + \sum_{k=-\infty}^{\infty} a_2 x_2(k) \\ &= a_1 \sum_{k=-\infty}^{\infty} x_1(k) + a_2 \sum_{k=-\infty}^{\infty} x_2(k) \\ &= a_1 y_1(n) + a_2 y_2(n) \end{aligned}$$

Since  $T(a_1 x_1(n) + a_2 x_2(n)) = a_1 y_1(n) + a_2 y_2(n)$ , the system is Linear.

TIME INVARIANCE:

$$y(n) = T(x(n))$$

Using the method portrayed in Figure 1,

$$T(x(n-n_0)) = \sum_{k=-\infty}^{\infty} x(k)$$

and

$$y(n) = \sum_{k=-\infty}^{\infty} x(k)$$

It follows that

$$y(n-n_0) = \sum_{k=-\infty}^{\infty} x(k)$$

Since  $y(n-n_0) = T(x(n-n_0))$ , the system is time invariant.

Thus, it is also LTI.

STABILITY:

Since the system is LTI, stability is tested by

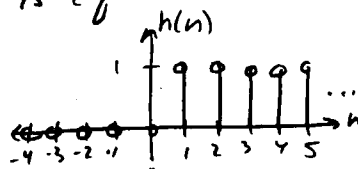
$$\sum_{k=-\infty}^{\infty} |h(k)| < \infty.$$

In this case, the impulse response,  $h(n)$ , is given by

$$h(n) = \sum_{k=n-1}^{\infty} \delta(k).$$

A plot of  $h(n)$  reveals that it is equal to  $u(n-1)$ .

$$h(n) = u(n-1)$$



Testing for stability shows that

$$\sum_{k=-\infty}^{\infty} |u(k-1)| = \infty.$$

Thus, the system is NOT stable.

CAUSALITY:

Since the system is LTI, causality is determined

if

$$h(n) = 0, \quad n < 0.$$

The impulse response was previously determined to be:

$$h(n) = u(n-1),$$

which is 0 for all  $n < 0$ . Thus, the system

IS CAUSAL.

SUMMARY

	<u>LINEAR</u>	<u>TIME INVARIANT</u>	<u>STABLE</u>	<u>CAUSAL</u>
A.)	✓	✓	✓	✓
B.)	✓		✓	
C.)	✓	✓	✓	✓
D.)	✓	✓		✓

OBSERVATIONS

- 1) of the 4 systems, the only one that is unstable has an impulse response that is a shifted unit step function.
- 2) of the 4 systems, the only one that is not time invariant is the system whose transform affects  $n$ , rather than  $(x(n))$ . Since this transform enlarges  $n$ , (i.e.  $n^2$ ), it makes  $y(n)$  anticipative. This forces  $y(n)$  to be noncausal.

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 gt4565a  
 ECE 4270

Consider the difference equation:

$$y[n] + \frac{1}{15} y[n-1] - \frac{2}{5} y[n-2] = x[n]$$

a.) general homogeneous solution

homogeneous form of difference equation

$$y[n] + \frac{1}{15} y[n-1] - \frac{2}{5} y[n-2] = 0$$

characteristic equation:

$$x^2 + \frac{1}{15}x - \frac{2}{5} = 0 \quad \text{roots: } = -\frac{2}{3} \text{ and } \frac{3}{5}$$

∴ the general solution for the homogeneous difference equation is

$A, B = \text{constants}$

$$A_h = A \left(-\frac{2}{3}\right)^n + B \left(\frac{3}{5}\right)^n$$

b.) impulse response of causal and anti causal versions of difference equation

Z-transforms used

$$y[n] + \frac{1}{15} y[n-1] - \frac{2}{5} y[n-2] = x[n]$$

$$Y(z) + \frac{1}{15} Y(z) z^{-1} - \frac{2}{5} Y(z) z^{-2} = X(z)$$

$$Y(z) \left[ 1 + \frac{1}{15} z^{-1} - \frac{2}{5} z^{-2} \right] = X(z)$$

$$H(z) = \frac{Y(z)}{X(z)} = \frac{1}{\left(1 + \frac{1}{15} z^{-1} - \frac{2}{5} z^{-2}\right)} = \frac{1}{(1 + \frac{2}{3} z^{-1})(1 - \frac{3}{5} z^{-1})}$$

$$= \frac{0.78947}{(1.5 + z^{-1})} - \frac{0.78947}{(-\frac{5}{3} + z^{-1})} = \frac{(\frac{2}{3}) \cdot 0.78947}{(1 + \frac{2}{3} z^{-1})} - \frac{(\frac{3}{5}) \cdot 0.78947}{(1 + \frac{3}{5} z^{-1})}$$

$$\therefore h_{\text{causal}} = \left(-\frac{2}{3}\right)^n u[n] (0.52632) + (0.4737) \left(\frac{3}{5}\right)^n u[n]$$

Z-transform rule

$$\frac{1}{1 - az^{-1}} \rightarrow a^n u[n]$$

$$h_{\text{causal}}[n] = (0.52632) \left(-\frac{2}{3}\right)^n u[n] + (0.4737) \left(\frac{3}{5}\right)^n u[n]$$

next page

b) Anti-causal:

$$\rightarrow h[n] = 0 \quad n \geq 0$$

$$h_{\text{anti-causal}}[n] = \sum_{k=1}^{\infty} (p_k)^n u[-n-1]$$

general form

Using homogeneous solution

$$\textcircled{1} h[n] = [A(-2/3)^n + B(3/5)^n] u[-n-1]$$

$$x[n] = \delta[n]$$

$$h[n] + \frac{1}{15} h[n-1] - \frac{2}{5} h[n-2] = x[n] \quad \text{from difference equation}$$

Use formula  $\textcircled{1}$  for  $h[-1], h[-2]$

$$n=0 \quad h[0] + \frac{1}{15} h[-1] - \frac{2}{5} h[-2] = \delta[0] = 1$$

$$\frac{1}{15} [A(-3/2) + B(5/3)] - \frac{2}{5} [A(3/2)^2 + B(5/3)^2] = 1$$

$$A(-3/2) + B(5/3) = 0 \rightarrow -2/5 h[-1] = 0$$

multiply both sides by 15

$$A(-3/2) + (5/3)B - 6 \cdot A(3/2)^2 - 6B(5/3)^2 = 15$$

$$(-3/2 - 6 \cdot 9/4)A + (5/3 - 6 \cdot 25/9)B = 15$$

$$-15A + (-15)B = 15$$

$$-A - B = 1$$

$$A + B = -1 \rightarrow A = -1 - B$$

$$-3/2 A + 5/3 B = 0 \rightarrow -3/2(-1-B) + 5/3 B = 0$$

$$3/2 + 3/2 B + 5/3 B = 0$$

$$B = -9/19$$

$$\therefore A = -10/19$$

Anti-causal

$$h_{\text{ac}}[n] = [(-10/19)(-2/3)^n + (-9/19)(3/5)^n] u[-n-1]$$



c.) prove  
causal h is stable  
and anti causal is unstable

$$\text{causal: } h_c[n] = \left[ \frac{10}{19} \left(-\frac{2}{3}\right)^n + \frac{9}{19} \left(\frac{3}{5}\right)^n \right] u[n]$$

$$(a)^\infty \rightarrow 0 \text{ when } |a| < 1$$

$$\sum_{k=-\infty}^{\infty} |h[k]| < \infty$$

$\left(-\frac{2}{3}\right)^n$  and  $\left(\frac{3}{5}\right)^n$  will decay as  $n \rightarrow \infty$   
 $\therefore h_c[n]$  represents a stable system.

$$\text{anti-causal: } h_{ac}[n] = \left[ -\frac{10}{19} \left(-\frac{2}{3}\right)^n - \frac{9}{19} \left(\frac{3}{5}\right)^n \right] u[-n-1]$$

$$(a)^\infty \rightarrow \infty \text{ when } |a| > 1$$

$$\sum_{k=-\infty}^{\infty} |h_{ac}[k]| = \sum_{k=-\infty}^{-1} \left( \frac{10}{19} \left(\frac{3}{2}\right)^{-k} + \frac{9}{19} \left(\frac{3}{5}\right)^{-k} \right)$$

behaves exactly like  $\sum_{k=1}^{\infty} \left( \frac{10}{19} \left(\frac{2}{3}\right)^k + \frac{9}{19} \left(\frac{5}{3}\right)^k \right) \rightarrow \infty$

$\therefore$  unstable

d.) Find a particular solution  
 when  $x[n] = \left(\frac{3}{5}\right)^n u[n]$

$$h_{\text{conv}}[n] = (0.52632) \left(-\frac{2}{3}\right)^n u[n] + (0.4737) \left(\frac{3}{5}\right)^n u[n]$$

$$x[n] \rightarrow X(z) = \frac{1}{1 - \frac{3}{5}z^{-1}}$$

$$H(z) = \frac{(0.52632)}{(1 + \frac{2}{3}z^{-1})} + \frac{(0.4737)}{(1 - \frac{3}{5}z^{-1})}$$

$$Y(z) = X(z)H(z) = \frac{(0.52632)}{(1 + \frac{2}{3}z^{-1})(1 - \frac{3}{5}z^{-1})} + \frac{(0.4737)}{(1 - \frac{3}{5}z^{-1})(1 - \frac{3}{5}z^{-1})}$$

$$Y(z) = \frac{0.41552}{1.5 + z^{-1}} - \frac{0.41552}{(-\frac{5}{3} + z^{-1})} + \frac{(0.4737)}{(1 - \frac{3}{5}z^{-1})^2}$$

$$Y(z) = \frac{\frac{2}{3}(0.4155)}{\frac{2}{3}(1.5 + z^{-1})} - \frac{(\frac{2}{3})(0.41552)}{(\frac{-2}{3})(-\frac{5}{3} + z^{-1})} + \frac{(0.4737)}{(1 - \frac{3}{5}z^{-1})^2}$$

$$Y(z) = \frac{0.277}{1 + \frac{2}{3}z^{-1}} + \frac{0.249312}{(1 - \frac{3}{5}z^{-1})} + \frac{(0.4737)}{(1 - \frac{3}{5}z^{-1})^2}$$

$$Y(z) \xrightarrow{z\text{-transform}} y[n] = 0.277 \cdot \left(-\frac{2}{3}\right)^n u[n] + (0.249312) \cdot \left(\frac{3}{5}\right)^n u[n] + (0.4737)(n) \left(\frac{3}{5}\right)^n u[n]$$

$$y[n] = \left[ (0.277) \left(-\frac{2}{3}\right)^n + (0.249) \left(\frac{3}{5}\right)^n + (0.4737)(n) \left(\frac{3}{5}\right)^n \right] \cdot u[n]$$

Solution: 
$$y[n] = x[n] * h[n] = \sum_{k=-\infty}^{\infty} h[k] x[n-k]$$

$$= \sum_{k=-\infty}^{\infty} \left(\frac{j}{2}\right)^k u[k] \cos((n-k)\pi) u[n-k]$$

clearly  $y[n] = 0$  for all  $n < 0$ .

for  $n \geq 0$ , 
$$y[n] = \sum_{k=0}^n \left(\frac{j}{2}\right)^k \cos[(n-k)\pi]$$

$$= \sum_{k=0}^n \left(\frac{j}{2}\right)^k (-1)^{n-k} = (-1)^n \sum_{k=0}^n \left(-\frac{j}{2}\right)^k$$

$$= (-1)^n \cdot \frac{1 - \left(-\frac{j}{2}\right)^{n+1}}{1 + \frac{j}{2}}$$

Hence, the steady-state response is

if  $n$  is even, 
$$y[n] = \frac{1 - \left(-\frac{j}{2}\right)^{n+1}}{1 + \frac{j}{2}} \rightarrow \frac{1}{1 + \frac{j}{2}} = 0.8 - 0.4j$$

if  $n$  is odd, 
$$y[n] = -\frac{1 - \left(-\frac{j}{2}\right)^{n+1}}{1 + \frac{j}{2}} \rightarrow -\frac{1}{1 + \frac{j}{2}} = -0.8 + 0.4j.$$

Also, I used program to compute the steady-state response. and I got the same answer. Here I prefer<sup>to</sup> theoretical derivation.

2.42

## Problem Statement

Consider the system in Figure P2.42-1.

- (a) Find the impulse response  $h[n]$  of the overall system.
- (b) Find the frequency response of the overall system.
- (c) Specify a difference equation that relates the output  $y[n]$  to the input  $x[n]$ .
- (d) Is this system causal? Under what condition would the system be stable?

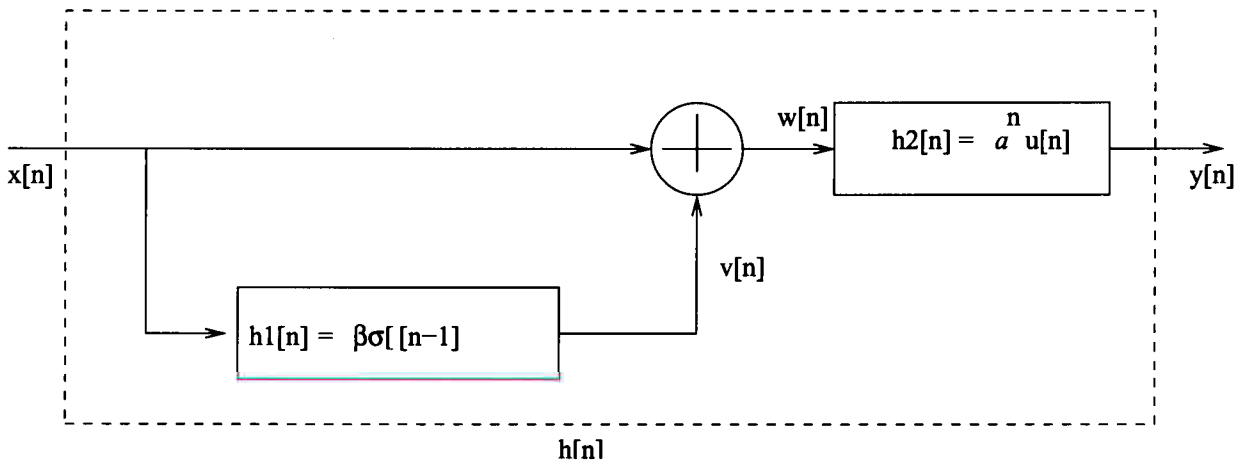


Figure 1: Figure P2.42-1

## Detailed Solution

$v[n]$  and  $w[n]$  are shown in Figure 1. We have:

$$\begin{aligned}v[n] &= x[n] * h_1[n] \\v[n] &= \sum_{k=-\infty}^{k=+\infty} \beta \delta[k-1] x[n-k] \\v[n] &= \beta x[n-1]\end{aligned}$$

and for  $w[n]$  we have:

$$w[n] = x[n] + \beta x[n-1]$$

It's easier to find the frequency response of the system first and then use DTFT<sup>-1</sup> to find the impulse response.

part(b):

$$\begin{aligned}y[n] &= w[n] * h_2[n] \\Y(e^{j\omega}) &= W(e^{j\omega}) H_2(e^{j\omega}) \\Y(e^{j\omega}) &= (X(e^{j\omega}) + \beta e^{-j\omega} X(e^{j\omega})) H_2(e^{j\omega}) \\Y(e^{j\omega}) &= X(e^{j\omega}) (1 + \beta e^{-j\omega}) H_2(e^{j\omega})\end{aligned}$$

Therefore the frequency response ( $H(e^{j\omega})$ ) of the system is:

$$H(e^{j\omega}) = \frac{1 + \beta e^{-j\omega}}{1 - \alpha e^{-j\omega}}$$

part(a):

$$H(e^{j\omega}) = \frac{1}{1 - \alpha e^{-j\omega}} + \beta \frac{e^{-j\omega}}{1 - \alpha e^{-j\omega}}$$

using the inverse of F.T. we have :

$$h[n] = \alpha^n u[n] + \beta \alpha^{n-1} u[n-1]$$

part(c):

$$\begin{aligned}y[n] &= h[n] * x[n] \\y[n] &= \sum_{k=-\infty}^{k=+\infty} h[k] x[n-k] \\y[n] &= \sum_{k=-\infty}^{k=+\infty} x[n-k] (\alpha^k u[k] + \beta \alpha^{k-1} u[k-1]) \\y[n] &= \sum_{k=0}^{k=+\infty} \alpha^k x[n-k] + \sum_{k=1}^{k=+\infty} \beta \alpha^{k-1} x[n-k]\end{aligned}$$

$$y[n-1] = \sum_{k=0}^{k=+\infty} \alpha^k x[n-1-k] + \sum_{k=1}^{k=+\infty} \beta \alpha^{k-1} x[n-1-k]$$

Substitute  $l+k = k'$ :

$$y[n-1] = \sum_{k'=1}^{k'=+\infty} \alpha^{k'-1} x[n-k'] + \sum_{k'=2}^{k'=+\infty} \beta \alpha^{k'-2} x[n-k']$$

We can rewrite  $y[n]$  as follows:

$$y[n] = x[n] + \sum_{k=1}^{k=+\infty} \alpha^k x[n-k] + \beta x[n-1] + \sum_{k=2}^{k=+\infty} \beta \alpha^{k-1} x[n-k]$$

$$y[n] = x[n] + \beta x[n-1] + \alpha \left( \sum_{k=1}^{k=+\infty} \alpha^{k-1} x[n-k] + \sum_{k=2}^{k=+\infty} \beta \alpha^{k-2} x[n-k] \right)$$

$$y[n] = x[n] + \beta x[n-1] + \alpha y[n-1]$$

$$y[n] - \alpha y[n-1] = x[n] + \beta x[n-1]$$

part(d): Causality requires that  $h[n] = 0$  for  $n < 0$ . It is clear that the condition holds for  $h[n]$  of the overall system obtained in part a.

Stability requires that  $\sum_{k=-\infty}^{+\infty} |h[k]| < \infty$ . To satisfy the stability condition for  $h[n]$  (from part (a)), the condition  $|\alpha| < 1$  must be true.