

Sustainable Systems: Power Electronics Issues

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Background

Power electronics is not only an enabling technology for sustainable energy systems, but it also motivates a re-examination of the fundamental analytical framework - **circuitry and theory**.

A dynamic interplay among theory, laboratory experiment and industrial practice.

Emphasis on faster phenomena - harmonics, EMI, **transients**.

Merging of established fields - instrumentation, protection, control.

Functions in a power electronic converter

Main converter functions: **1.** Switching - abrupt changes in the energy flow,
2. Conduction - enables the energy flow,,
3. Storage - compensates for variations in the energy flow,
4. Control - oversees the first three functions and interfaces with the environment.

A technological invariant - product of power and the switching frequency is (approximately) constant for a switch technology (Ge, Si, SiC)

Power Switches Today

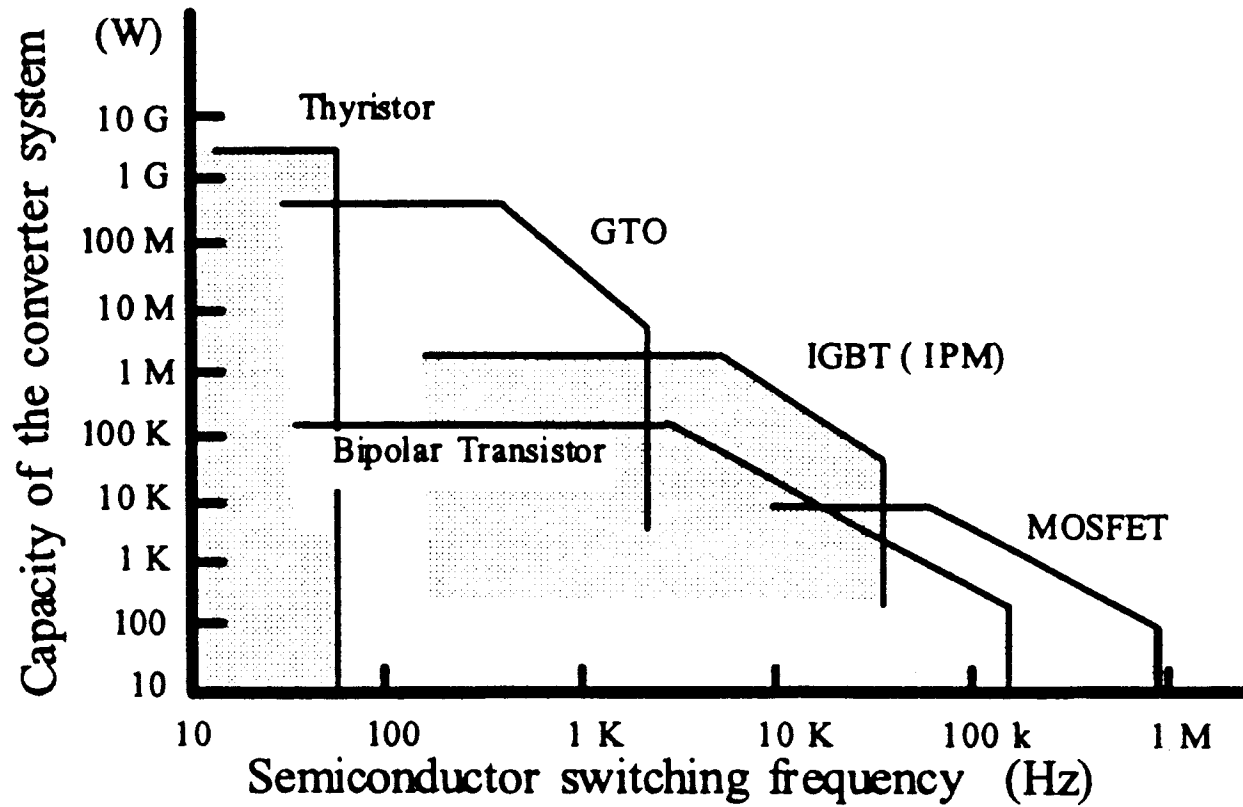


Figure 1: Semiconductor switches today [Uchida, Yamada 2000].

Basic Inverter Topology

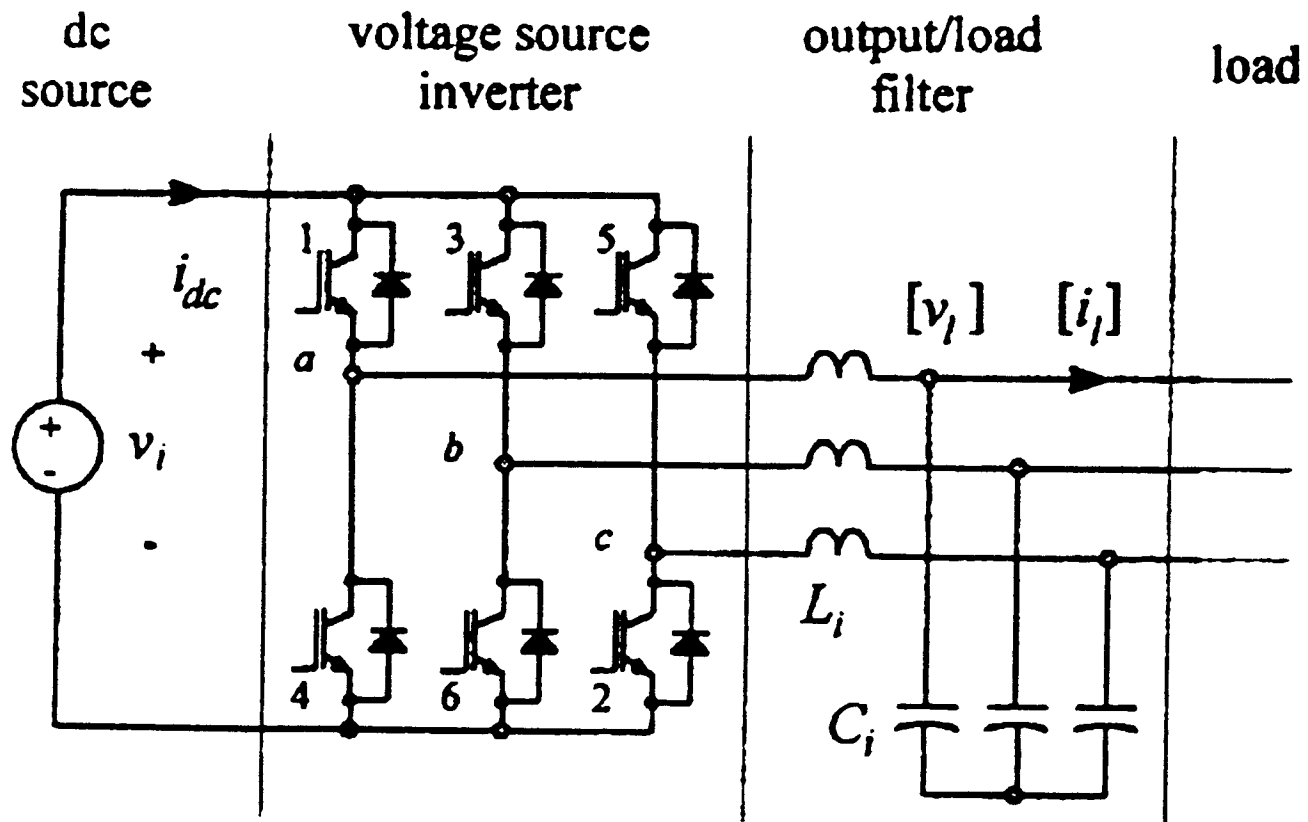


Figure 2: Basic inverter topology - voltage sourced inverter [Espinoza et al, 1999].

Industrial Concerns

Packaging is essential for cost reduction.

Simplified topologies and increased product integration at the low end.

Emerging concerns:

1. Emissions: electromagnetic, acoustic, thermal;
2. Material take-back.

System Integration

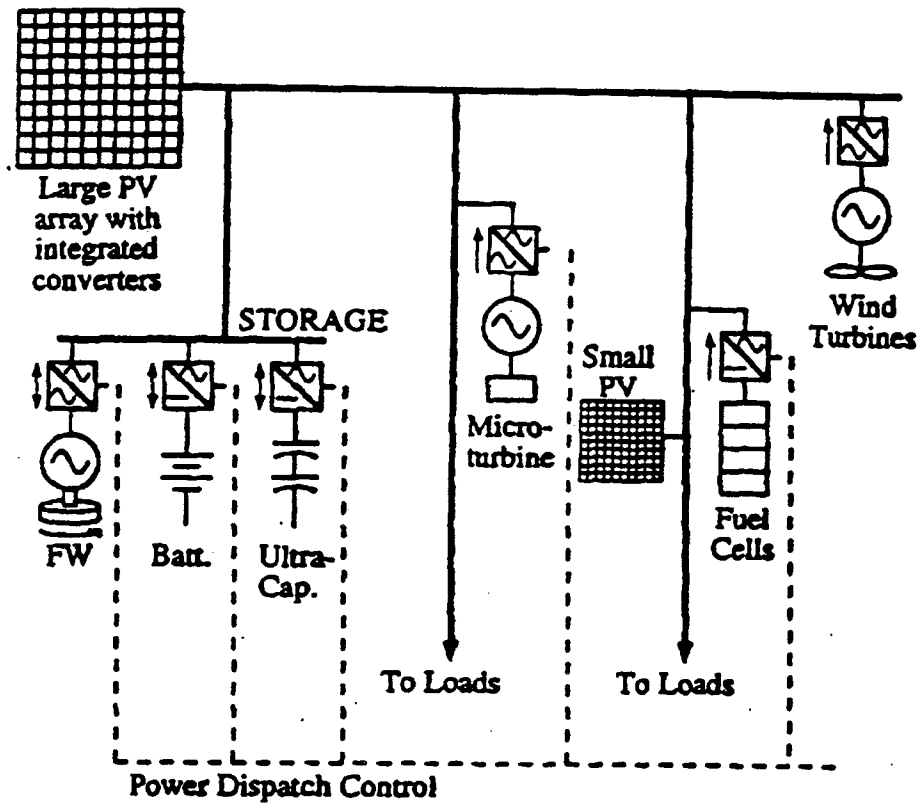


Figure 3: Distributed generation - a new energy supply framework [Divan, Brumsickle 1999]

Power Electronics and Wind Energy

Various degrees of utilization of power electronics are possible in sustainable energy sources - e.g., wind:

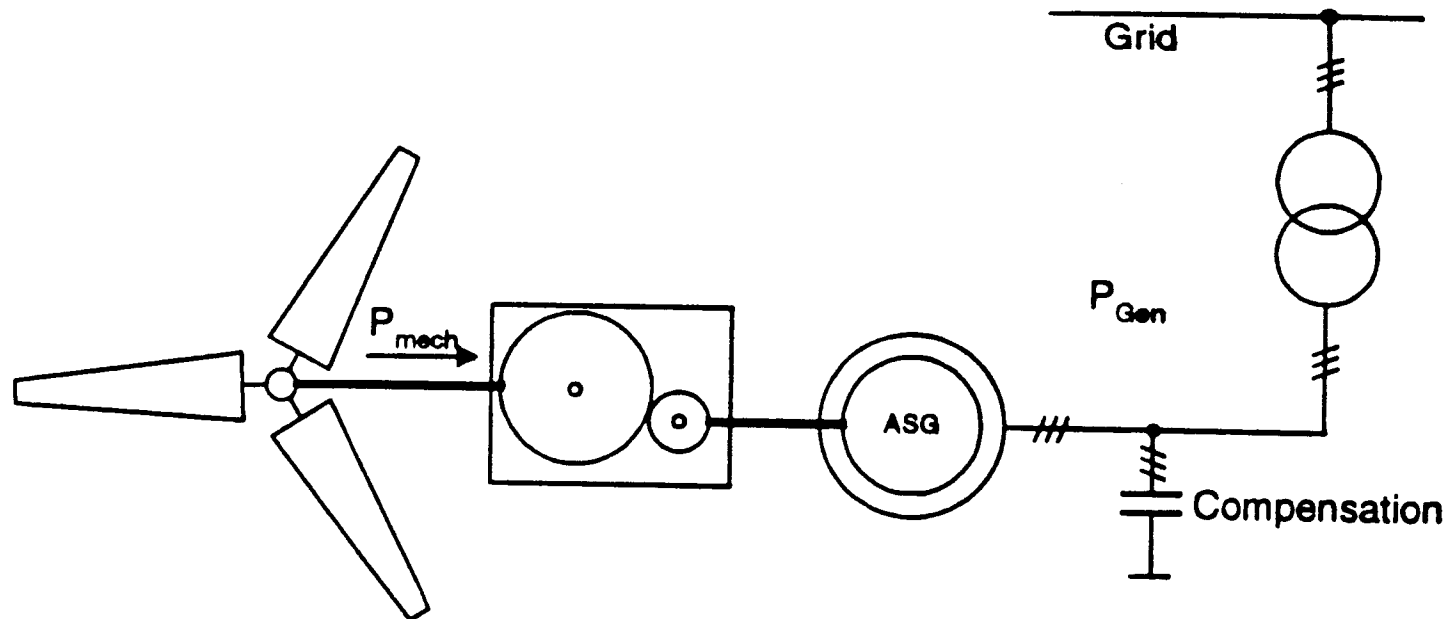


Figure 4: Danish concept [Muller et al., 2000].

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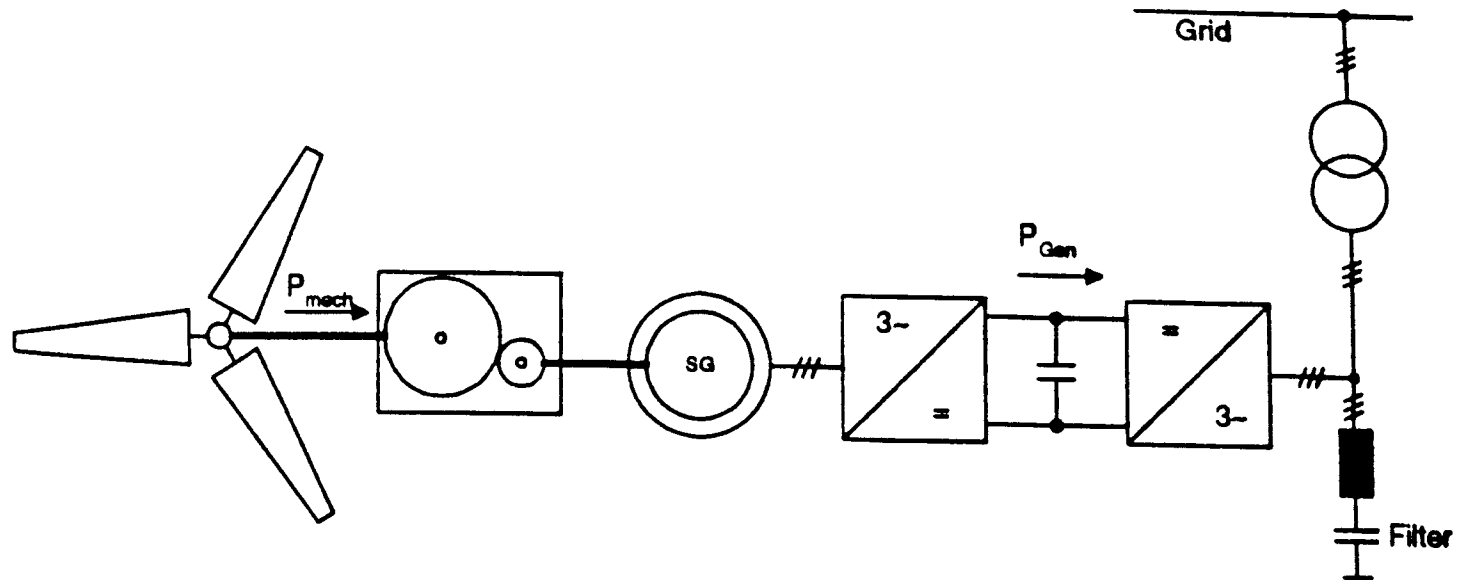


Figure 5: In-line solution.

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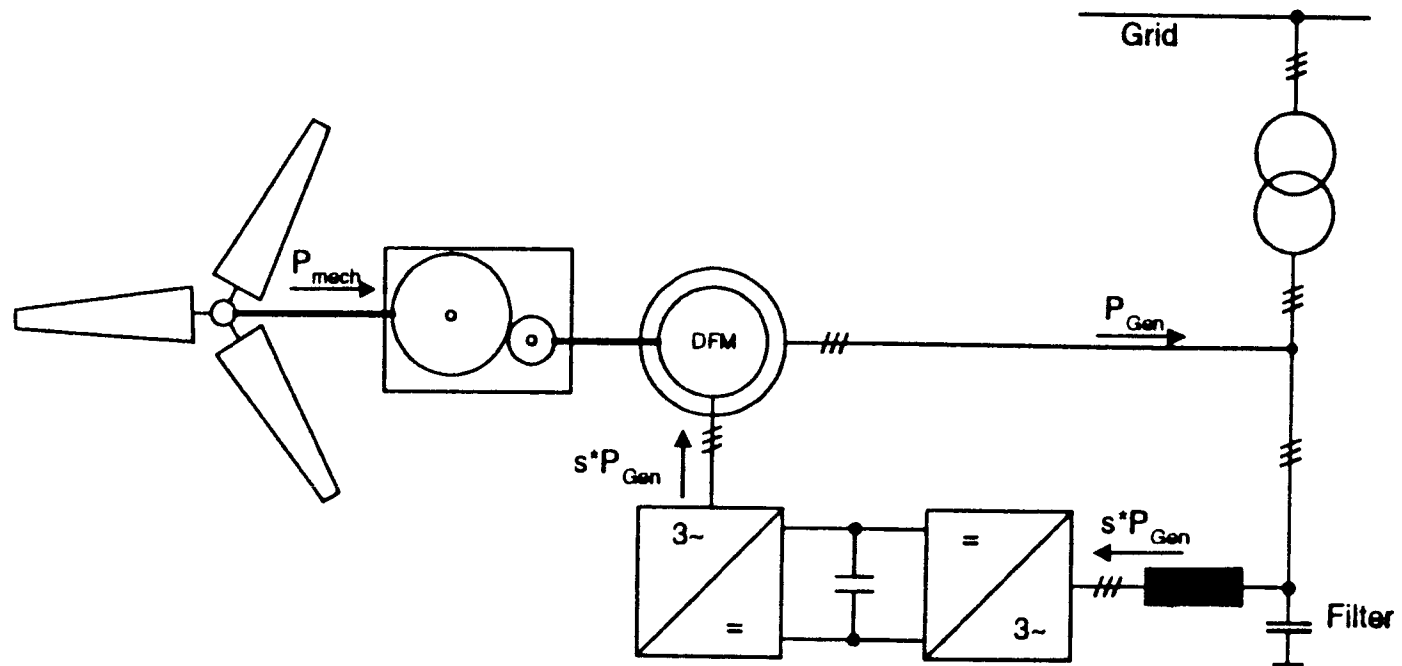
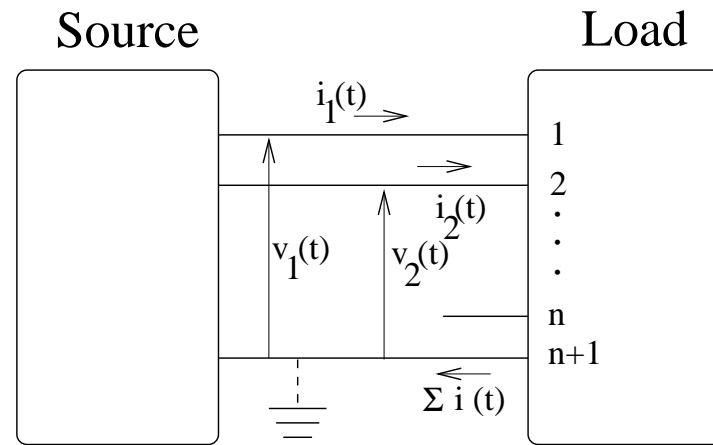


Figure 6: Doubly-fed induction machine.

Theory - compensation of inactive power

Vector waveforms (e.g. voltages, currents) $x(t)$ are T -periodic; fundamental frequency is $\omega = 2\pi/T$; **no** limit on the number of harmonics.



The ℓ -th harmonic (vector) is

$$\langle x \rangle_\ell = \frac{1}{T} \int_T x(t) e^{-j\ell\omega t} dt \quad (1)$$

Collective rms quantities $\|x\|^2 = \langle x^\top x \rangle_0$.

Time-domain definitions, ctd. 1

The *instantaneous power* is

$$p(t) = v(t)^\top i(t) \quad (2)$$

The *instantaneous* projection of $i(t)$ onto the subspace spanned by $v(t)$ is defined to be the *active current* $i_a(t)$ [Depenbrock(1964), Akagi(1983), Willems(1992)]:

$$i_a(t) = \frac{i(t)^\top v(t)}{v(t)^\top v(t)} v(t) = \frac{p(t)}{v(t)^\top v(t)} v(t) \quad (3)$$

The *inactive current* $i_x(t)$

$$i_x(t) = i(t) - i_a(t) \quad (4)$$

i_x is orthogonal to i_a (and v) at any time instant; consequently $p(t) = v(t)^\top i(t) = v(t)^\top i_a(t)$.

Time-domain definitions, ctd. 2

A different notion of active current is due to Fryze (1931):

$$i_{act}(t) = \frac{\langle p(t) \rangle_0}{\|v\|^2} v(t) = \frac{\langle p(t) \rangle_0}{\langle v(t)^\top v(t) \rangle_0} v(t) \quad (5)$$

In circuit terms:

- $i_a(t)$ is the smallest current (by rms) that supplies the same $p(t)$ as the load current $i(t)$,
- $i_{act}(t)$ is the smallest current (by rms) that supplies the same **dc component** $\langle p(t) \rangle_0$ of $p(t)$ as the load current $i(t)$.

Example

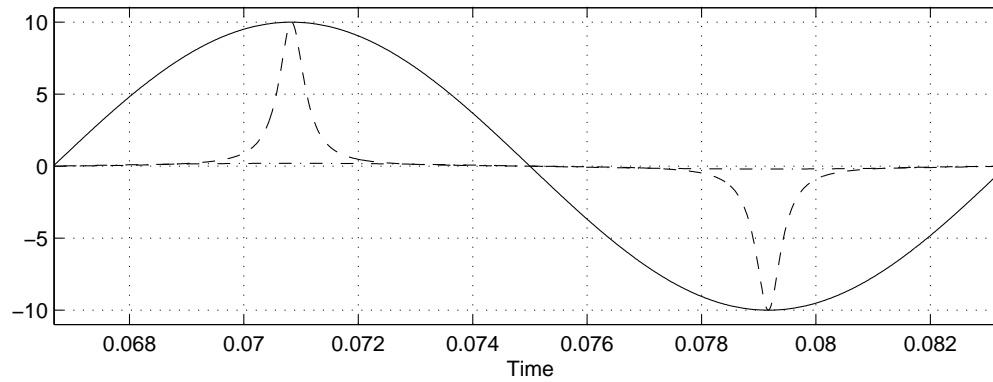
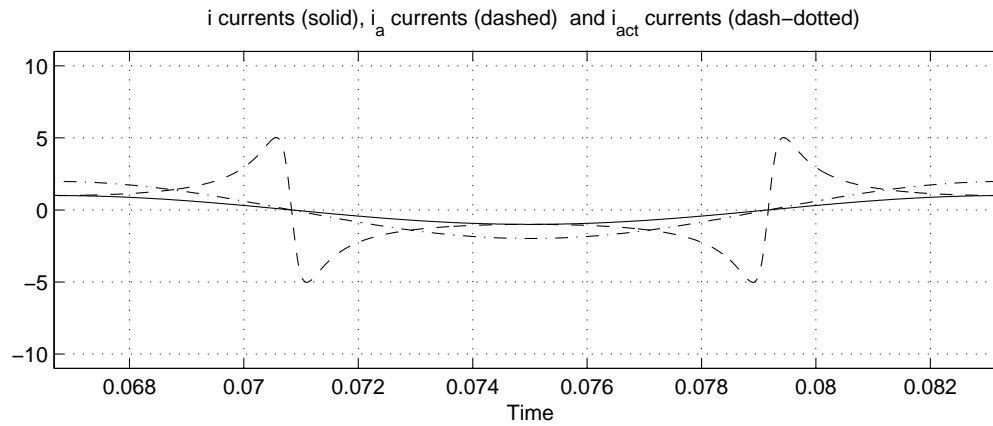
Consider a *resistive* two phase circuit with

$$v(t) = \begin{bmatrix} V_1 \cos(\omega t) \\ V_2 \sin(\omega t) \end{bmatrix}, \quad i(t) = \begin{bmatrix} VI/V_1 \cos(\omega t) \\ VI/V_2 \sin(\omega t) \end{bmatrix}$$

We pick $V_1 = 1$, $V_2 = 0.1$, $VI = 1$

Note that both voltages and currents contain **only** the fundamental frequency component.

Example, ctd. 1



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Example, ctd. 2

In this example $\|v\|^2 = (v_1^2 + v_2^2)/2 = 0.505$, $\|i\|^2 = (VI)^2(1/2V_1^2 + 1/2V_2^2) = 50.5$; also

$$i_a(t) = \frac{2VI}{V_1 + V_2} \sum_{\ell=0}^{\infty} \left(-\frac{V_1 - V_2}{V_1 + V_2}\right)^\ell \begin{bmatrix} \cos(2\ell + 1)\omega t \\ \sin(2\ell + 1)\omega t \end{bmatrix}$$

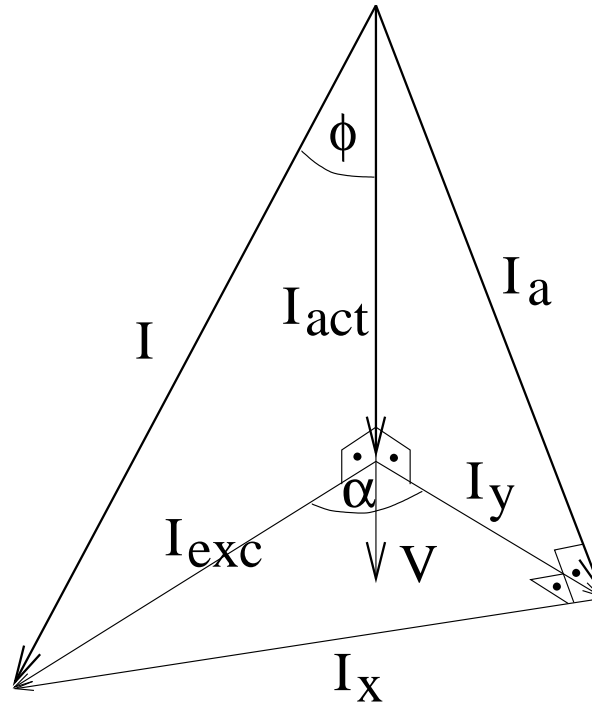
for our numerical values $\|i_a\|^2 = (VI)^2/V_1V_2 = 10$;

$$i_{act}(t) = \frac{2VI}{V_1^2 + V_2^2} \begin{bmatrix} V_1 \cos(\omega t) \\ V_2 \sin(\omega t) \end{bmatrix}$$

and $\|i_{act}\|^2 = 2(VI)^2/(V_1^2 + V_2^2) = 1.98$.

Frequency domain results

In the (potentially infinite dimensional Hilbert) space of Fourier coefficients:



We define the inner product of vector signals x and y , with corresponding matrices X and Y , as

$$\langle x^\top y \rangle_0 = \frac{1}{T} \int_0^T x(t)^\top y(t) dt = \text{tr}(X^H Y) \quad (6)$$

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Dynamic Extensions

A (possibly complex) waveform $x(\cdot)$ can be represented on the interval $(t - T, t]$ using (short-time) Fourier series:

$$x(\tau) = \sum_{k=-\infty}^{\infty} X_k(t) e^{jk\omega_s\tau}$$

where $X_k(t)$ are the complex, slowly time-varying Fourier coefficients, or *dynamic phasors*.

$$X_k(t) = \frac{1}{T} \int_{t-T}^t x(\tau) e^{-jk\omega_s\tau} d\tau = \langle x \rangle_k (t)$$

Our dynamical models describe evolution of $X_k(t)$; for real $x(\cdot)$ we have $X_{-k} = X_k^*$

A useful facts:

$$\frac{dX_k}{dt} = \left\langle \frac{d}{dt} x \right\rangle_k - j k \omega_s X_k$$

Polyphase Systems in Transients

Dynamic symmetric components - recall $\alpha = e^{j\frac{2\pi}{3}}$,

$$\begin{bmatrix} x_a \\ x_b \\ x_c \end{bmatrix} (\tau) = \sum_{l=-\infty}^{\infty} e^{jl\omega_s\tau} \underbrace{\frac{1}{\sqrt{3}} \begin{bmatrix} 1 & 1 & 1 \\ \alpha^* & \alpha & 1 \\ \alpha & \alpha^* & 1 \end{bmatrix}}_{\mathcal{M}} \begin{bmatrix} X_{p,l} \\ X_{n,l} \\ X_{z,l} \end{bmatrix} (t)$$

$$\begin{bmatrix} X_{p,l} \\ X_{n,l} \\ X_{z,l} \end{bmatrix} (t) = \frac{1}{T} \int_{t-T}^t e^{-jl\omega_s\tau} \mathcal{M}^H \begin{bmatrix} x_a \\ x_b \\ x_c \end{bmatrix} (\tau) d\tau = \begin{bmatrix} \langle x \rangle_{p,l} \\ \langle x \rangle_{n,l} \\ \langle x \rangle_{z,l} \end{bmatrix} (t).$$

$$\frac{d}{dt} \begin{bmatrix} X_{p,l} \\ X_{n,l} \\ X_{z,l} \end{bmatrix} (t) = \mathcal{M}^H \begin{bmatrix} \langle d/d\tau x_a(\tau) \rangle_l \\ \langle d/d\tau x_b(\tau) \rangle_l \\ \langle d/d\tau x_c(\tau) \rangle_l \end{bmatrix} (t) - jl\omega_s \begin{bmatrix} X_{p,l} \\ X_{n,l} \\ X_{z,l} \end{bmatrix} (t)$$

Long Term Vision

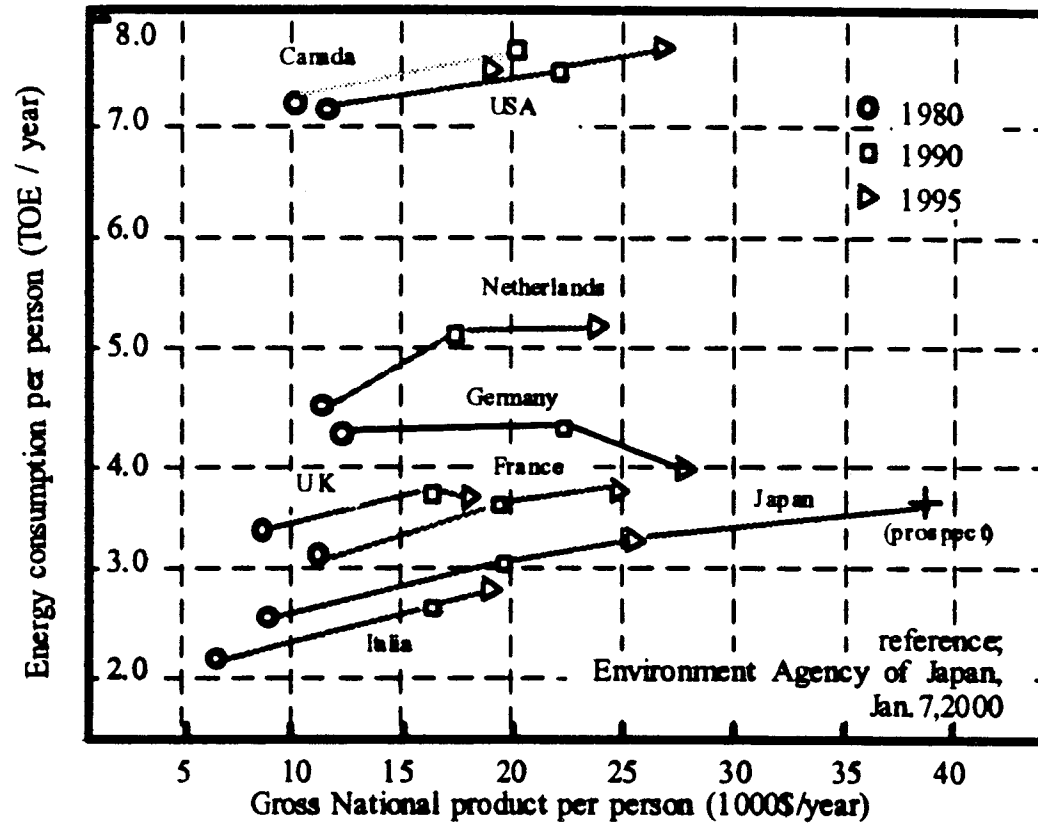


Figure 7: GNP and Energy Consumption [Uchida.Yamada 2000].