

Nyquist Pulses

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ECE4823

Receiver Filter

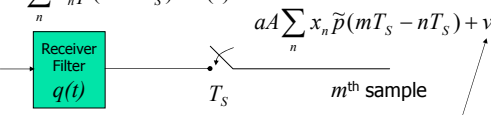
- Baseband output is a superposition of filtered pulses plus filtered noise

$$aA \sum_n x_n p(t - nT_s) + n(t) \xrightarrow{\text{Receiver Filter } q(t)} aA \sum_n x_n \tilde{p}(t - nT_s) + v(t)$$

$$\tilde{p}(t) = \frac{1}{2} \int_{-\infty}^{+\infty} p(t - \tau) q(\tau) d\tau$$

Sampling in the Receiver

- The baseband representation of the receiver in additive Gaussian white noise (AWGN) :

$$aA \sum_n x_n \tilde{p}(t - nT_s) + v(t)$$


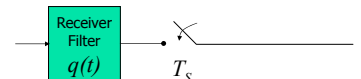
$$aA \sum_n x_n \tilde{p}(mT_s - nT_s) + v_m$$

m^{th} sample

zero-mean Gaussian RVs

Ideal Situation

- Ideally, the m^{th} sample depends on only the m^{th} symbol and the noise

$$aA \sum_n x_n \tilde{p}(mT_s - nT_s) + v_m = aAx_m + v_m$$


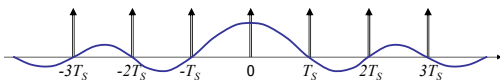
Necessary Condition

- To have this ideal situation, we must have

$$\tilde{p}(mT_s - nT_s) = \begin{cases} 1 & m = n \\ 0 & m \neq n \end{cases}$$

- or, alternatively,

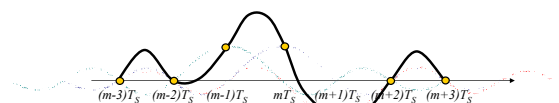
$$\tilde{p}(t) \sum_{n=-\infty}^{+\infty} \delta(t - nT_s) = \delta(t)$$



Received Signal Example

- Suppose the $m-1^{\text{st}}$, m^{th} , and $m+1^{\text{st}}$ symbols were $+1$, $+1$, -1 , respectively

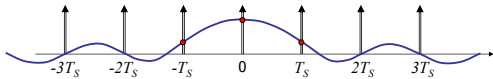
The signal is constrained only at the sample points



Non-Ideal Situation

- Suppose the received pulse did not satisfy the condition, but did this instead:

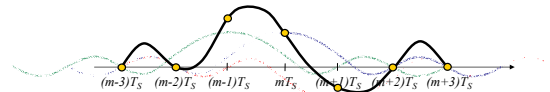
$$\tilde{p}(t) \sum_{n=-\infty}^{+\infty} \delta(t - nT_s) = \delta(t) + 0.3\delta(t - T_s) + 0.3\delta(t + T_s)$$



Intersymbol Interference (ISI)

- The $m-1^{\text{st}}$, m^{th} , and $m+1^{\text{st}}$ received samples become

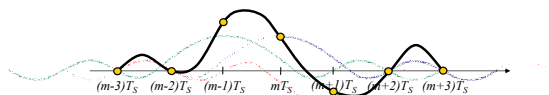
$$1.3 + v_{m-1}, 1 + v_m, -0.7 + v_{m+1}$$



The Negative Effect of ISI

- The worst case dominates the probability of bit error

$$1.3 + v_{m-1}, 1 + v_m, 0.7 + v_{m+1}$$



Nyquist Pulses

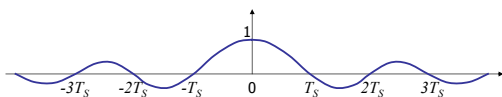
- Pulses that satisfy the condition for no ISI are called Nyquist Pulses

$$\tilde{p}(t) \sum_{n=-\infty}^{+\infty} \delta(t - nT_s) = \delta(t)$$

Sinc Pulse

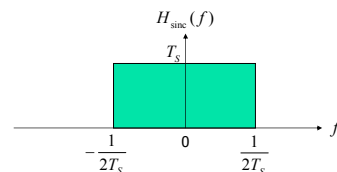
- The sinc pulse is a Nyquist pulse

$$\text{sinc}(t/T_s) = \frac{\sin(\pi t/T_s)}{\pi t/T_s}$$



Fourier Transform of Sinc Pulse

- The F.T. of the sinc pulse is the "brick wall" characteristic

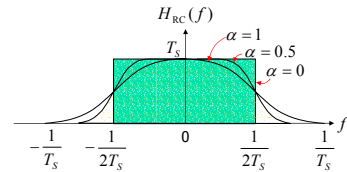


Pros and Cons of Sinc

- The F.T. has the narrowest possible bandwidth of all Nyquist pulses
- Noncausal
- Roll-off too gradual

Popular Alternative: Raised Cosine

- Single parameter α



Raised Cosine Formulas

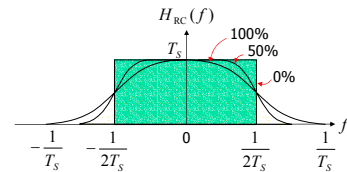
[Pahlavan '95]

$$H_{RC}(f) = \begin{cases} \frac{T_s}{2} \left[1 - \sin \frac{\pi T_s}{\alpha} \left(\left| f \right| - \frac{1}{2T_s} \right) \right] & 0 \leq |f| \leq (1-\alpha)/2T_s \\ 0 & (1-\alpha)/2T_s \leq |f| \leq (1+\alpha)/2T_s \\ 0 & (1+\alpha)/2T_s \leq |f| \end{cases}$$

$$p_{RC}(t) = F^{-1}\{H_{RC}(f)\} = \frac{\sin \pi t / T_s}{\pi t / T_s} \frac{\cos \alpha \pi t / T_s}{1 - 4\alpha^2 t^2 / T_s^2}$$

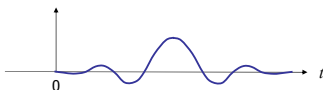
Excess Bandwidth

- The bandwidth of any pulse will be X% greater than that of the sinc Nyquist pulse
- This percentage is the excess bandwidth



Pros of Raised Cosine

- Smooth F.T.
 - Easy to build filters that approximate it
- Falls off as $1/t^3$
- Can approximate a delayed pulse with a causal filter response



Summary

- Intersymbol interference (ISI) can dominate BER
- Nyquist Pulses are pulses that do not cause ISI in an AWGN channel
- Sinc is the Nyquist Pulse with the least bandwidth, but it is impractical
- Raised Cosine is a popular Nyquist Pulse
- Excess Bandwidth indicates how much broader is the bandwidth than that of the sinc Nyquist Pulse



References

- [Rapp, '02] T.S. Rappaport, *Wireless Communications*, Prentice Hall, 2002
- [Stuber, '01] Gordon Stuber, *Principles of Mobile Communication, 2nd ed*, Kluwer Academic Publishers, 2001
- [Pahlavan, '95] Pahlavan and Levesque, *Wireless Information Networks*, Wiley Interscience, 1995